## ST52F510/F513/F514

## 8-BIT INTELLIGENT CONTROLLER UNIT (ICU) Two Timer/PWMs, ADC, $I^{2} \mathrm{C}, \mathrm{SPI}, \mathrm{SCI}$

## Memories

■ Up to 8 Kbytes Single Voltage Flash Memory

- 256 bytes of Register File
- 256 bytes of RAM
- Up to 4 Kbytes Data EEPROM
- In Situ Programming in Flash devices (ISP)
- Single byte and Page modes and In Application Programming for writing data in Flash memory
- Readout protection and flexible write protection


## Core

■ Register File based architecture

- 107 basic instructions
- Hardware multiplication and division
- Decision Processor for the implementation of Fuzzy Logic algorithms
■ Deep System and User Stacks


## Clock and Power Supply

■ Up to 24 MHz clock frequency

- Programmable Oscillator modes:
- 10 MHz Internal Oscillator
- External Clock/ Oscillator
- External RC Oscillator

■ Power-On Reset (POR)
■ Programmable Low Voltage Detector (PLVD) with 3 configurable thresholds

- Power Saving features


## Interrupts

- 8 interrupt vectors with one SW Trap

■ Non-Maskable Interrupt (NMI)

- Two Port Interrupts with up to 16 sources


## I/O Ports

- From 10 up to 22 I/O PINs configurable in pullup, push-pull, weak pull-up, open-drain and high-impedance
■ High current sink/source in all pins

TARGET SPECIFICATION


## Peripherals

- On-chip 10-bit A/D Converter with 8 channel analog multiplexer and Autocalibration.
- 2 Programmable 16 bit Timer/PWMs with internal 16-bit Prescaler featuring:
- PWM output
- Input capture
- Output compare
- Pulse generator mode
- Watchdog timer
- Serial Communication Interface (SCI) with asynchronous protocol (UART).
- $I^{2} C^{\top M}$ Peripheral with master and slave mode
- 3-wire SPITM Peripheral supporting Single Master and Multi Master SPI modes


## Development tools

- High level Software tools
- 'C' Compiler
- Emulator
- Low cost Programmer
- Gang Programmer


## TABLE OF CONTENTS

1 GENERAL DESCRIPTION ..... 7
1.1 Introduction ..... 7
1.2 Functional Description ..... 8
1.2.1 Memory Programming Mode ..... 8
1.2.2 Working Mode ..... 8
1.3 Pin Description ..... 16
2 INTERNAL ARCHITECTURE. ..... 17
2.1 Control Unit and Data Processing Unit ..... 17
2.1.1 Program Counter ..... 18
2.1.2 Flags ..... 18
2.2 Arithmetic Logic Unit ..... 19
2.3 Register Description ..... 20
3 ADDRESSING SPACES. ..... 21
3.1 Memory Interface .....  21
3.2 Register File .....  21
3.3 Program/Data Memory ..... 21
3.4 System and User Stacks ..... 23
3.5 Input Registers .....  24
3.6 Output registers ..... 24
3.7 Configuration Registers \& Option Bytes ..... 25
4 MEMORY PROGRAMMING ..... 31
4.1 Program/Data Memory Organization ..... 31
4.2 Memory Programming ..... 32
4.2.1 Programming Mode start ..... 32
4.2.2 Fast Programming procedure ..... 33
4.2.3 Random data writing ..... 33
4.2.4 Option Bytes Programming. ..... 34
4.3 Memory Verify. ..... 35
4.3.1 Fast read procedure ..... 35
4.3.2 Random data reading ..... 36
4.4 Memory Lock ..... 36
4.5 ID Code. ..... 37
4.6 Error cases ..... 37
4.7 In -Situ Programming (ISP) ..... 38
4.8 In -Application Programming (IAP) ..... 38
4.8.1 Single byte write ..... 38
4.8.2 Block write ..... 38
4.8.3 Memory Corruption Prevention ..... 38
4.8.4 Option Bytes ..... 39
4.8.5 Input Register ..... 39
5 INTERRUPTS ..... 40
5.1 Interrupt Processing ..... 40
5.2 Global Interrupt Request Enabling ..... 40
5.3 Interrupt Sources ..... 41
5.4 Interrupt Maskability and Priority Levels ..... 41
5.5 Interrupt RESET ..... 41
5.6 Register Description ..... 42
6 CLOCK, RESET \& POWER SAVING MODES ..... 44
6.1 Clock ..... 44
6.2 Reset. ..... 45
6.2.1 External Reset ..... 45
6.2.2 Reset Procedures ..... 45
6.3 Programmable Low Voltage Detector ..... 46
6.4 Power Saving modes ..... 46
6.4.1 Wait Mode ..... 46
6.4.2 Halt Mode ..... 46
6.5 Register Description ..... 48
6.5.1 Configuration Register ..... 48
6.5.2 Option Bytes ..... 48
7 I/O PORTS ..... 50
7.1 Introduction ..... 50
7.2 Input Mode ..... 50
7.3 Output Mode ..... 50
7.4 Interrupt Mode ..... 50
7.5 Alternate Functions ..... 51
7.6 Register Description ..... 51
7.6.1 Configuration Registers ..... 52
7.6.2 Input Registers ..... 54
7.6.3 Output Registers ..... 55
8 FUZZY COMPUTATION (DP). ..... 56
8.1 Fuzzy Inference ..... 56
8.2 Fuzzyfication Phase ..... 56
8.3 Inference Phase ..... 56
8.4 Defuzzyfication ..... 57
8.5 Input Membership Function ..... 57
8.6 Output Singleton ..... 58
8.7 Fuzzy Rules ..... 58
9 INSTRUCTION SET ..... 60
9.1 Addressing Modes ..... 60
9.2 Instruction Types. ..... 60
10 10-bit A/D CONVERTER ..... 65
10.1 Introduction ..... 65
10.2 Functional Description ..... 66
10.3 Operating Modes ..... 66
10.3.1 One Channel Single Mode ..... 66
10.3.2 Multiple Channels Single Mode ..... 66
10.3.3 One Channel Continuous Mode ..... 66
10.3.4 Multiple Channels Continuous Mode ..... 67
10.4 Power Down Mode ..... 67
10.5 A/D Converter Register Description ..... 67
10.5.1 A/D Converter Configuration Registers ..... 67
10.5.2 Input Registers ..... 68
11 WATCHDOG TIMER. ..... 69
11.1 Functional Description ..... 69
11.2 Register Description ..... 69
12 PWM/TIMERS ..... 71
12.1 Introduction ..... 71
12.2 Timer Mode ..... 71
12.3 PWM Mode ..... 73
12.3.1 Simultaneous Start ..... 74
12.4 Timer Interrupts. ..... 74
12.5 PWM/Timer 0 Register Description. ..... 74
12.5.1 PWM/Timer 0 Configuration Registers ..... 74
12.5.2 PWM/Timer 0 Input Registers ..... 75
12.5.3 PWM/Timer 0 Output Registers ..... 76
12.6 PWM/Timer 1 Register Description. ..... 77
12.6.1 PWM/Timer 1 Configuration Registers ..... 77
12.6.2 PWM/Timer 1 Input Registers ..... 78
12.6.3 PWM/Timer 1 Output Registers ..... 79
13 SERIAL COMMUNICATION INTERFACE ..... 80
13.1 SCI Receiver block ..... 80
13.1.1 Recovery Buffer Block ..... 81
13.1.2 SCDR_RX Block ..... 81
13.2 SCI Transmitter Block ..... 82
13.3 Baud Rate Generator Block ..... 82
13.4 SCI Register Description. ..... 84
13.4.1 SCI Configuration Registers ..... 84
13.4.2 SCI Input Registers. ..... 85
13.4.3 SCI Output Register ..... 85
14 I2C BUS INTERFACE (I2C) ..... 86
14.1 Introduction ..... 86
14.2 Main Features ..... 86
14.3 General Description ..... 86
14.3.1 Mode Selection ..... 86
14.3.2 Communication Flow ..... 86
14.3.3 SDA/SCL Line Control ..... 87
14.4 Functional Description ..... 87
14.4.1 Slave Mode ..... 87
14.4.2 Master Mode ..... 88
14.5 Register Description ..... 92
14.5.1 I2C Interface Configuration Registers ..... 92
14.5.2 I2C Interface Input Registers ..... 93
14.5.3 I2C Interface Output Registers ..... 95
15 SERIAL PERIPHERAL INTERFACE (SPI) ..... 96
15.1 Introduction ..... 96
15.2 Main Features ..... 96
15.3 General description ..... 96
15.4 Functional Description ..... 96
15.4.1 Master Configuration ..... 96
15.4.2 Slave Configuration ..... 98
15.4.3 Data Transfer Format ..... 98
15.4.4 Write Collision Error ..... 98
15.4.5 Master Mode Fault ..... 99
15.4.6 Overrun Condition ..... 101
15.4.7 Single Master and Multimaster Configurations ..... 101
15.4.8 Interrupts ..... 102
15.5 SPI Register Description ..... 103
15.5.1 SPI Configuration Registers ..... 103
15.5.2 SPI Input Register ..... 104
15.5.3 SPI Output Register ..... 105

## 1 GENERAL DESCRIPTION

### 1.1 Introduction

ST52F510/F513/F514 are devices of ST FIVE family of 8 -bit Intelligent Controller Units (ICU), which can perform, both boolean and Fuzzy algorithms in an efficient manner, in order to reach the best performances that the two methodologies allow.
Produced by STMicroelectronics using the reliable high performance CMOS process for Single Voltage Flash versions, ST52F510/F513/F514 include integrated on-chip peripherals that allow maximization of system reliability, and decreased system costs in order to minimize the number of external components.
The flexible I/O configuration of ST52F510/F513/ F514 allow one to interface with a wide range of external devices (for example D/A converters or power control devices), and to communicate with the most common serial standards.
ST52F510/F513/F514 pins are configurable. The user can set input or output signals on each single pin in 8 different modes, reducing the need for external components in order to supply a suitable interface with the port pins.
A hardware multiplier and divider, together with a wide instruction set, allow the implementation of complex functions by using a single instruction. Therefore, program memory utilization and computational speed is optimized.
Fuzzy Logic dedicated structures in ST52F510/ F513/F514 ICU's can be exploited to model complex system with high accuracy in a useful and simple manner.
Fuzzy Expert Systems for overall system management and Fuzzy Real time Controls can be designed to increase performance at competitive costs.
The linguistic approach characterizing Fuzzy Logic is based on a set of IF-THEN rules, which describe the control behavior and on Membership Functions associated with input and output variables.
Up to 340 Membership Functions, with triangular and trapezoidal shapes, or singleton values are available to describe fuzzy variables.
The Timer/PWM peripheral allows one to manage power devices and timing signals, by implementing different operating modes and high frequency PWM (Pulse Width Modulation) controls. Input Capture and Output Compare functions are available on the Timers.
The Timer has a 16-bit programmable internal Prescaler and a 16-bit Counter, which can use internal or external START/STOP signals and clock.

An internal programmable WATCHDOG is available to avoid loop errors and reset the ICU.
ST52F510/F513/F514 includes a 10-bit, selfcalibrating, Analog to Digital Converter with an 8analog channel Multiplexer. Single/Multiple channels and Single/Sequence conversion modes are supported. External reference can be supplied to obtain more stability and precision in the conversion.
ST52F510/F513/F514 supply different peripherals to implement the most common serial communication protocols. SCl allows the performance of serial asynchronous communication (UART). $I^{2} \mathrm{C}$ and SPI peripherals allow the implementation of synchronous serial protocols. $I^{2} \mathrm{C}$ peripherals can work both in master and slave mode. SPI implements Single and Multi Master modes using 3 -wire.
Up to 8 interrupt vectors are available, which allow synchronization with peripherals and external devices. Non-Maskable Interrupt and S/W TRAP are available. All interrupts have configurable priority levels and are maskable excluding the Non-Maskable Interrupt, which has fixed top level priority. Two versatile Port Interrupts are available for synchronization with external sources.
The ST52F510/F513/F514 also include an on-chip Power-on-Reset (POR), which provides an internal chip reset during power up situation and a Programmable Low Voltage Detector (PLVD), which causes the ICU to reset if the voltage source $V_{D D}$ dips below a threshold. Three programmable thresholds are available, allowing to work with different supply voltages (from 2.7 to 5.5 V ).
In order to optimize energy consumption, two different power saving modes are available: Wait mode and Halt mode.
Internal Oscillator at $10 \mathrm{MHz} \pm 1 \%$ is available. External clock, quartz oscillator or RC oscillator are also applicable. The device always starts with the Internal Oscillator, then it reads an Option Byte where the clock mode to be used is programmed.
Program Memory addressing capability addresses up to 8 Kbytes of memory location to store both program instructions and data.
Memory can be locked by the user in order to prevent external undesired operations.
Operations may be performed on data stored in RAM, allowing direct combination of new inputs and feedback data. All RAM bytes are used like Register File.
An additional RAM bench is added to the Program Memory addressing space in order to allow the management of the System/User Stacks and user data storage.

ST52F510/F513/F514 supply the system stack and the user stack located in the additional RAM bench. The user stack can be located anywhere in the additional RAM by writing the top address in the configuration registers, in order to avoid overlap with other data.
Single Voltage Flash allows the user to reprogram the devices on-board by means of the In Situ Programming (ISP) feature. It is possible to store in safe way up to 4 K of data in the available EEPROM memory benches. Permanent data, both in Flash and EEPROM can be managed by means of the In-Application-Programming (IAP) feature. Single byte and Page write modes are supported. Flexible write protection, of permanent data or program instructions, is also available.
The Instruction Set composed of up to 107 instructions allows code compression and high speed in the program implementation.
A powerful development environment consisting of a board and software tools allows an easy configuration and use of ST52F510/F513/F514.
The Visual FIVE software tool allows the development and debugging of projects via a userfriendly graphical interface and optimization of generated microcode.
Third-party Hardware Emulators and 'C' Compiler are available to speed-up the application implementation and time-to-market.

### 1.2 Functional Description

ST52F510/F513/F514 ICU's can work in two modes according to the Vpp signal levels:

- Memory Programming Mode
- Working Mode

During Working Mode Vpp must be tied to Vss. To enter the Memory Programming Mode, the Vpp pin must be tied to Vdd.
A RESET signal must be applied to the device to switch from one mode to the other.

### 1.2.1 Memory Programming Mode.

The ST52F510/F513/F514 memory is loaded in the Memory Programming Mode. All instructions and data are written inside the memory during this phase.

The Option Bytes are loaded during this phase by using the programming tools. The Option Bytes can only be loaded in this phase and cannot be modified run-time.
Data and commands are transmitted by using the $I^{2} \mathrm{C}$ protocol, implemented using the internal $I^{2} \mathrm{C}$ peripheral. The In-Situ Programming protocol (ISP) uses the following pins:

- SDA and SCL for transmission
- Vpp for entering in the mode
- RESET for starting the protocol in a stable status
- Vdd and Vss for the power supply.

The Internal clock is used in this phase.

### 1.2.2 Working Mode.

The processor starts the working phase following the instructions, which have been previously loaded in the first locations of the memory. The first instruction must be a jump to the first program instruction, skipping the data (interrupt vectors, Membership Functions, user data) stored in the first memory page.
ST52F510/F513/F514's internal structure includes two computational blocks, the CONTROL UNIT (CU) and the DATA PROCESSING UNIT (DPU), which performs boolean functions. The DECISION PROCESSOR (DP) block cooperates with these blocks to perform Fuzzy algorithms.
The DP can manage up to 340 different Membership Functions for the antecedent part of fuzzy rules. The consequent terms of the rules are "crisp" values (real numbers). The maximum number of rules that can be defined is limited by the dimensions of the standard algorithm implemented.
The Program/Data Memory is shared between Fuzzy and standard algorithms. Within this memory, the user data can be stored both in non volatile memory as well as in the RAM locations.
The Control Unit (CU) reads information and the status of the peripherals.
Arithmetic calculus can be performed on these values by using the internal CU and Register File, which supports all computations. The peripheral inputs can be Fuzzy and/or arithmetic output values contained in the Register File or Program/ Data Memory.

Table 1.1 ST52F510/F513/F514 Devices Summary

| Device | NVM | RF | RAM | EEPROM | Timers | ADC | Comms | I/O | Package |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ST52F510YmM6 | 4/8 K FLASH | 256 | 256 | - | 2X16-bit | $\begin{aligned} & \text { 10-bit } \\ & 2 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCII} \mathrm{I}^{2} \mathrm{C}$ | 10 | So 16 |
| ST52F510FmM6 | 4/8 K FLASH | 256 | 256 | - | 2X16-bit | $\begin{aligned} & \text { 10-bit } \\ & 6 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCII} \mathrm{I}^{2} \mathrm{C}$ | 14 | So 20 |
| ST52F510FmB6 | 4/8 K FLASH | 256 | 256 | - | 2X16-bit | $\begin{aligned} & \text { 10-bit } \\ & 6 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCII} \mathrm{I}^{2} \mathrm{C}$ | 14 | Dip 20 |
| ST52F510GmM6 | 4/8 K FLASH | 256 | 256 | - | 2X16-bit | $\begin{aligned} & \text { 10-bit } \\ & 8 \mathrm{Ch} \end{aligned}$ | SCI I2C SPI | 22 | So 28 |
| ST52F510GmB6 | 4/8 K FLASH | 256 | 256 | - | 2X16-bit | $\begin{aligned} & 10-\text {-bit } \\ & 8 \mathrm{Ch} \\ & \hline \end{aligned}$ | $\mathrm{SCI} \mathrm{I}^{2} \mathrm{C}$ SPI | 22 | SDip 32 |
| ST52F513YmM6 | 4/8 K FLASH | 256 | 256 | 256 | 2X16-bit | $\begin{aligned} & \text { 10-bit } \\ & 2 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCII}{ }^{2} \mathrm{C}$ | 10 | So 16 |
| ST52F513FmM6 | 4/8 K FLASH | 256 | 256 | 256 | 2X16-bit | $\begin{aligned} & 10 \text {-bit } \\ & 6 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCII} \mathrm{I}^{2} \mathrm{C}$ | 14 | So 20 |
| ST52F513FmB6 | 4/8 K FLASH | 256 | 256 | 256 | 2X16-bit | $\begin{aligned} & 10 \text {-bit } \\ & 6 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCII}{ }^{2} \mathrm{C}$ | 14 | Dip 20 |
| ST52F513GmM6 | 4/8 K FLASH | 256 | 256 | 256 | 2X16-bit | $\begin{aligned} & \text { 10-bit } \\ & 8 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCI} \mathrm{I}^{2} \mathrm{C}$ SPI | 22 | So 28 |
| ST52F513GmB6 | 4/8 K FLASH | 256 | 256 | 256 | 2X16-bit | $\begin{aligned} & 10 \text {-bit } \\ & 8 \mathrm{Ch} \\ & \hline \end{aligned}$ | $\mathrm{SCI}{ }^{2} \mathrm{C}$ SPI | 22 | SDip 32 |
| ST52F514YmM6 | 4 K FLASH | 256 | 256 | 1024 / 4096 | 2X16-bit | $\begin{aligned} & \text { 10-bit } \\ & 2 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCII}{ }^{2} \mathrm{C}$ | 10 | So 16 |
| ST52F514FmM6 | 4 K FLASH | 256 | 256 | 1024 / 4096 | 2X16-bit | $\begin{aligned} & \text { 10-bit } \\ & 6 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCII}{ }^{2} \mathrm{C}$ | 14 | So 20 |
| ST52F514FmB6 | 4 K FLASH | 256 | 256 | 1024 / 4096 | 2X16-bit | $\begin{aligned} & 10 \text {-bit } \\ & 6 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCII} \mathrm{I}^{2} \mathrm{C}$ | 14 | Dip 20 |
| ST52F514GmM6 | 4 K FLASH | 256 | 256 | 1024 / 4096 | 2X16-bit | $\begin{aligned} & \text { 10-bit } \\ & 8 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCI}{ }^{2} \mathrm{C}$ SPI | 22 | So 28 |
| ST52F514GmB6 | 4 K FLASH | 256 | 256 | 1024 / 4096 | 2X16-bit | $\begin{aligned} & \text { 10-bit } \\ & 8 \mathrm{Ch} \end{aligned}$ | $\mathrm{SCI}{ }^{2} \mathrm{C}$ SPI | 22 | SDip 32 |


| COMMON FEATURES | ST52F510/F513/F514 |
| :---: | :---: |
| Watchdog | Yes |
| Other Features | NMI, PLVD, POR |
| Temperature Range | From $-40^{\circ}$ to $+85^{\circ}$ |
| Operating Supply | $2.7-5.5 \mathrm{~V}$ |
| CPU Frequency | from 1 to 24 MHz. |

## Legend:

Sales code:
Memory type (t):
Subfamily (nnn):
Pin Count (c):
Memory Size (m):
Packages (p):
Temperature (y):

## ST52tnnncmpy

F=FLASH
510, 513, 514
$Y=16$ pins, $F=20$ pins, $G=28$ pins, $K=32 / 34$ pins
$2=4 \mathrm{~Kb}, 3=8 \mathrm{~Kb}$ Flash (ST52F510 \& ST52F513)
1=1024, 3=4096 EEPROM (only ST52F514)
$\mathrm{B}=\mathrm{PDIP}, \mathrm{M}=\mathrm{PSO}, \mathrm{T}=$ TQFP
$0=+25,1=0+70,3=-40+125,5=-10+85,6=-40+85,7=-40+105$

Figure 1.1 ST52F510/F513/F514 Block Diagram


Figure 1.2 ST52F510/F513/F514 SO20/DIP20 Pin Configuration


Figure 1.3 ST52F510/F513/F514 SO16 Pin Configuration


Figure 1.4 ST52F510/F513/F514 SDIP32/DIP28 Pin Configuration


Table 1.2 ST52F510/F513/F514 SDIP32 Pin List

| SDIP32 | NAME | Programming Phase | Working Phase |
| :---: | :---: | :---: | :---: |
| 1 | Vdd | Digital Power Supply | Digital Power Supply |
| 2 | VddIO | Digital Power Supply | Digital I/O Ports Power Supply |
| 3 | OSCOUT |  | Oscillator Output |
| 4 | OSCIN |  | Oscillator Input |
| 5 | Vpp | Programming Mode Selector | Programming Mode Selector |
| 6 | PB0/VREF/AIN0 |  | Digital I/O, A/D Voltage Reference, Analog Input |
| 7 | PB1/AIN1 |  | Digital I/O, Analog Input |
| 8 | PB2/AIN2 |  | Digital I/O, Analog Input |
| 9 | PB3/AIN3 |  | Digital I/O, Analog Input |
| 10 | PB4/AIN4 |  | Digital I/O, Analog Input |
| 11 | PB5/AIN5 |  | Digital I/O, Analog Input |
| 12 | PB6/AIN6 |  | Digital I/O, Analog Input |
| 13 | PB7/AIN7 |  | Digital I/O, Analog Input |
| 14 | PC0/SCK |  | Digital I/O, SPI Serial Clock |
| 15 | PC1/MOSI |  | Digital I/O, SPI Master out Slave in |
| 16 | N.C |  | Not Connected |
| 17 | N.C |  | Not Connected |
| 18 | PC2/MISO |  | Digital I/O, SPI Master in Slave out |
| 19 | PC3/SS |  | Digital I/O, SPI Slave Select |
| 20 | PC4/TX |  | Digital I/O, SCI Transmission |
| 21 | PC5/TRES |  | Digital I/O, Timer/PWM 0 Reset |
| 22 | PA7/INT |  | Digital I/O, Non Maskable Interrupt |
| 23 | PA6/T0OUT |  | Digital I/O, Timer/PWM 0 output |
| 24 | PA5/TCLK/TX |  | Digital I/O, Timer/PWM 0 clock |
| 25 | PA4/TSTRT |  | Digital I/O, Timer/PWM 0 start/stop |
| 26 | PA3/RX |  | Digital I/O, SCI Reception |
| 27 | PA2/T1OUT |  | Digital I/O, Timer/PWM 1 output |
| 28 | PA1/SDA | Serial Data I/O | Digital I/O, I ${ }^{2} \mathrm{C}$ Serial Data I/O |
| 29 | PA0/SCL | Serial Clock | Digital I/O, $\mathrm{I}^{2} \mathrm{C}$ Serial Clock |
| 30 | RESET | General Reset | General Reset |
| 31 | VssiO | Digital Ground | Digital I/O Ports Ground |
| 32 | Vss | Digital Ground | Digital Ground |

Table 1.3 ST52F510/F513/F514 SO28 Pin List

| SO28 | NAME | Programming Phase | Working Phase |
| :---: | :---: | :---: | :---: |
| 1 | Vdd | Digital Power Supply | Digital Power Supply |
| 2 | oscout |  | Oscillator Output |
| 3 | OSCIN |  | Oscillator Input |
| 4 | Vpp | Programming Mode Selector | Programming Mode Selector |
| 5 | Pbo/VREF/AIN0 |  | Digital I/O, A/D Voltage Reference, Analog Input |
| 6 | PB1/AIN1 |  | Digital I/O, Analog Input |
| 7 | PB2/AIN2 |  | Digital I/O, Analog Input |
| 8 | PB3/AIN3 |  | Digital I/O, Analog Input |
| 9 | PB4/AIN4 |  | Digital I/O, Analog Input |
| 10 | PB5/AIN5 |  | Digital I/O, Analog Input |
| 11 | PB6/AIN6 |  | Digital I/O, Analog Input |
| 12 | PB7/AIN7 |  | Digital I/O, Analog Input |
| 13 | PCO/SCK |  | Digital I/O, SPI Serial Clock |
| 14 | PC1/MOSI |  | Digital I/O, SPI Master out Slave in |
| 15 | PC2/MISO |  | Digital I/O, SPI Master in Slave out |
| 16 | PC3/SS |  | Digital I/O, SPI Slave Select |
| 17 | PC4/TX |  | Digital I/O, SCI Transmission |
| 18 | PC5/TRES |  | Digital I/O, Timer/PWM 0 Reset |
| 19 | PA7/INT |  | Digital I/O, Non Maskable Interrupt |
| 20 | PA6/T00UT |  | Digital I/O, Timer/PWM 0 output |
| 21 | PA5/TCLK/TX |  | Digital I/O, Timer/PWM 0 clock |
| 22 | PA4/TSTRT |  | Digital I/O, Timer/PWM 0 start/stop |
| 23 | PA3/RX |  | Digital I/O, SCI Reception |
| 24 | PA2/T10UT |  | Digital I/O, Timer/PWM 1 output |
| 25 | PA1/SDA | Serial Data I/O | Digital I/O, $I^{2} \mathrm{C}$ Serial Data I/O |
| 26 | PAO/SCL | Serial Clock | Digital I/O, $I^{2} \mathrm{C}$ Serial Clock |
| 27 | RESET | General Reset | General Reset |
| 28 | Vss | Digital Ground | Digital Ground |

Table 1.4 ST52F510/F513/F514 SO20/DIP20/SO16 Pin List

| SO20 <br> DIP20 | SO16 | NAME | Programming Phase | Working Phase |
| :---: | :---: | :---: | :---: | :---: |
| 1 | 1 | Vdd | Digital Power Supply | Digital Power Supply |
| 2 | 2 | OSCOUT |  | Oscillator Output |
| 3 | 3 | OSCIN |  | Oscillator Input |
| 4 | 4 | Vpp | Programming Mode Selector | Programming Mode Selector |
| 5 | 5 | PB0/VREF/AINO |  | Digital I/O, A/D Voltage Reference, Analog Input |
| 6 | 6 | PB1/AIN1 |  | Digital I/O, Analog Input |
| 7 | - | PB2/AIN2 |  | Digital I/O, Analog Input |
| 8 | - | PB3/AIN3 |  | Digital I/O, Analog Input |
| 9 | - | PB4/AIN4 |  | Digital I/O, Analog Input |
| 10 | - | PB5/AIN5 |  | Digital I/O, Analog Input |
| 11 | 7 | PA7/INT |  | Digital I/O, Non Maskable Interrupt |
| 12 | 8 | PA6/T0OUT |  | Digital I/O, Timer/PWM 0 output |
| 13 | 9 | PA5/TCLK/TX |  | Digital I/O, Timer/PWM 0 clock, SCI transmission |
| 14 | 10 | PA4/TSTRT |  | Digital I/O, Timer/PWM 0 start/stop |
| 15 | 11 | PA3/RX |  | Digital I/O, SCI Reception |
| 16 | 12 | PA2/T1OUT |  | Digital I/O, Timer/PWM 1 output |
| 17 | 13 | PA1/SDA | Serial Data I/O | Digital I/O, I'C Serial Data I/O |
| 18 | 14 | PA0/SCL | Serial Clock | Digital I/O, IC Serial Clock |
| 19 | 15 | RESET | General Reset | General Reset |
| 20 | 16 | Vss | Digital Ground | Digital Ground |

### 1.3 Pin Description

ST52F510/F513/F514 pins can be set in digital input mode, digital output mode, interrupt mode or in Alternate Functions. Pin configuration is achieved by means of the configuration registers. The functions of the ST52F510/F513/F514 pins are described below:
$V_{D D}$. Main Power Supply Voltage.
$V_{\text {DdIo. }}$ I/O Ports Power Supply Voltage. It is reccomended to connect this pin with a supply voltage de-coupled with $\mathrm{V}_{\mathrm{DD}}$ in order to improve the immunity from the noise generated by the I/O switching.
$\mathbf{V}_{\mathbf{S S}}$. Digital circuit Ground.
$\mathrm{V}_{\text {SSIO }}$ I/O Ports Ground. See $\mathrm{V}_{\text {DDIO }}$
$\mathbf{V}_{\text {Pp. }}$. Programming/Working mode selector. During the Programming phase $V_{P P}$ must be set to $V_{D D}$. In Working phase $\mathrm{V}_{\mathrm{PP}}$ must be equal to $\mathrm{V}_{\mathrm{SS}}$.

OSCin and OSCout. These pins are internally connected to the on-chip oscillator circuit. A quartz crystal or a ceramic resonator can be connected between these two pins in order to allow correct use of ST52F510/F513/F514 with various stability/ cost trade-offs. An external clock signal can be applied to OSCin: in this case OSCout must be grounded. To reduce costs, an RC circuit can be applied to the OSCin pin to establish the internal clock frequency, instead of the quartz. Without any connection, the device can work with its internal clock generator ( 10 MHz )

RESET. This signal is used to reset the ST52F510/ F513/F514 and re-initialize the registers and control signals. It is also used when switching from the Programming Mode to Working Mode and vice versa.

PA0-PA7, PB0-PB7,PC0-PC5. These lines are organized as I/O ports. Each pin can be configured as an input, output (with pull-up, push-pull, weak-pull-up, open-drain, high-impedance), or as an interrupt source.

VREF, AINO-AIN7. These pins are used to input the analog signals into the A/D Converter. An analog multiplexer is available to switch these inputs to the A/D Converter. The pin VREF is used to input an external A/D Reference Voltage.

TOOUT, T1OUT. These pins output the signals generated by the PWM/Timer 0 and PWM/Timer 1 peripheral.

TRES, TSTRT, TCLK. These pins are related to the PWM/Timer 0 peripheral and are used for Input Capture and event counting. The TRES pin is used to set/reset the Timer; the TSTRT pin is used to start/stop the counter. The Timer can be driven by the internal clock or by an external signal connected to the TCLK pin.

INT. This pin is used as input for the Non-Maskable (top level) interrupt. The interrupt signal is detected only if the pin is configured in Alternate Function.

SCL, SDA. These pin are used respectively as Serial Clock and Serial Data I/O in I ${ }^{2}$ C peripheral protocol. They are used also in Programming Mode to receive and transmit data.

TX, RX. Serial data output of SCI Transmitter block (TX) and Serial data input of the SCI Receiver block (RX).

SCK, MISO, MOSI, SS. These pins are used by the Serial Peripheral Interface (SPI) peripheral. SCK is the serial clock line. MISO (Master In Slave Out) and MOSI (Master Out Slave In) are the serial data lines, which work in input or in output depending on if the device is working in slave or master mode. The SS pin allows the selection of the device master/slave mode.

## 2 INTERNAL ARCHITECTURE

ST52F510/F513/F514's architecture is Register File based and is composed of the following blocks and peripherals:

- Control Unit (CU)
- Data Processing Unit (DPU)
- Decision Processor (DP)
- ALU
- Memory Interface

■ up to 256 bytes Register File

- Program/Data Memory
- Data EEPROM
- Interrupts Controller
- Clock Oscillator
- PLVD and POR
- Digital I/O ports
- Analog Multiplexer and A/D Converter
- Timer/PWMs
- $\mathrm{I}^{2} \mathrm{C}$
- SPI
- SCI


### 2.1 Control Unit and Data Processing Unit

The Control Unit (CU) decodes the instructions stored in the Program Memory and generates the appropriate control signals. The main parts of the CU are illustrated in Figure 2.1.
The five different parts of the CU manage Loading, Logic/Arithmetic, Jump, Control and the Fuzzy instruction set.
The block called "Collector" manages the signals deriving from the different parts of the CU. The collector defines the signals for the Data Processing Unit (DPU) and Decision Processor (DP), as well as for the different peripherals of the ICU.
The block called "Arbiter" manages the different parts of the CU, so that only one part of the system is activated during working mode.
The CU structure is extremely flexible and was designed with the purpose of easily adapting the core of the microcontroller to market needs. New instruction sets or new peripherals can easily be included without changing the structure of the microcontroller, maintaining code compatibility.
A set of 107 different instructions is available. Each instruction requires a number of clock pulses to be performed that depends on the complexity of the instruction itself. The clock pulses to execute the instructions are driven directly by the masterclock, which has the same frequency of the oscillator signal supplied.

Figure 2.1 CU Block Diagram


Figure 2.2 Data Processing Unit (DPU)


The DPU receives, stores and sends the instructions deriving from the Program/Data Memory, Register File or from the peripherals. It is controlled by the CU on the basis of the decoded instruction. The Fuzzy registers store the partial results of the fuzzy computation. The accumulator register is used by the ALU and is not accessible directly: the instructions used by the ALU can address all the Register File locations as operands, allowing a more compact code and a faster execution.
The following addressing modes are available: inherent, immediate, direct, indirect, bit direct.

### 2.1.1 Program Counter.

The Program Counter (PC) is a 16-bit register that contains the address of the next memory location to be processed by the core. This memory location may be both an instruction or data address.
The Program Counter's 16-bit length allows the direct addressing of a maximum of 64 Kbytes in the Program/Data Memory space.

The PC can be changed in the following ways:

- JP (Jump)
PC = Jump Address
- Interrupt
- RETI
- RET
- CALL
- Reset
- Normal Instruction $P C=P C+1$


### 2.1.2 Flags.

The ST FIVE core includes different sets of flags that correspond to 2 different modes: normal mode and interrupt mode. Each set of flags consist of a CARRY flag (C), ZERO flag (Z) and SIGN flag (S). Each set is stacked: one set of flags is used during normal operation and other sets are used during each level of interrupt. Formally, the user has to manage only one set of flags: $\mathrm{C}, \mathrm{Z}$ and S since the flag stack operation is performed automatically.

Each interrupt level has its own set of flags, which is saved in the Flag Stack during interrupt servicing. These flags are restored from the Flag Stack automatically when a RETI instruction is executed.
If the ICU was in normal mode before an interrupt, after the RETI instruction is executed, the normal flags are restored.
Note: A subroutine CALL is a normal mode execution. For this reason a RET instruction, consequent to a CALL instruction, doesn't affect the normal mode set of flags.
Flags are not cleared during context switching and remain in the state they were in at the exit of the last interrupt routine switching.
The Carry flag is set when an overflow occurs during arithmetic operations, otherwise it is cleared. The Sign flag is set when an underflow occurs during arithmetic operations, otherwise it is cleared.
The flags, related to the current context, can be checked by reading the FLAGS Input Register 38 (026h).

Figure 2.3 Multiplication


### 2.2 Arithmetic Logic Unit

The 8-bit Arithmetic Logic Unit (ALU) performs arithmetic calculations and logic instructions such as: sum, subtraction, bitwise AND, OR, XOR, bit set and reset, bit test and branch, right/left shift and rotate (see the Chapter 9 Instruction Set for further details).
In addition, the ALU of ST52F510/F513/F514 can perform multiplication (MULT) and division (DIV). Multiplication is performed by using 8 bit operands storing the result in 2 registers ( 16 bit values); the division instruction addresses the MSB of the dividend (the LSB is stored in the next address): the result and remainder are stored in these source addresses (see Figure 2.3 and Figure 2.4).
In order to manage signed type values, the ALU also performs addition and subtraction with offset (ADDO and SUBO). These instructions respectively subtract and add 128 to the overall result, in order to manage values logically in the range between $-128,127$.

Figure 2.4 Division


### 2.3 Register Description

## Flags Register (FLAG)

Input Register 38 (026h) Read Only
Reset Value: 00000000 (00h)
7 0

| - | - | - | - | - | $z$ | $s$ | $c$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-3: Not Used

Bit 2: Z Zero flag
Bit 1: S Sign flag

Bit 0: C Carry flag

## 3 ADDRESSING SPACES

ST52F510/F513/F514 has six separate addressing spaces:

- Register File
- Program/Data Memory

■ Stacks

- Input Registers
- Output Registers


## - Configuration Registers

Each space is addressed by a load type instruction that indicates the source and the destination space in the mnemonic code (see Figure 3.1).

### 3.1 Memory Interface

The read/write operation in the space addresses are managed by the Memory Interface, which can recognize the type of memory addressed and set the appropriate access time and mode.
In addition, the Memory Interface manages the In Application Programming (IAP) functions in Flash devices like writing cycle and memory write protection.

### 3.2 Register File

The Register File consists of 256 general purpose 8 -bit RAM locations called "registers" in order to recall the functionality.
The Register File exchanges data with all the other addressing spaces and is used by the ALU to perform all the arithmetic and logic instructions. These instructions have any Register File address as operands.
Data can be moved from one location to another by using the LDRR instruction; see further ahead for information on the instruction used to move data between the Register File and the other addressing spaces.

### 3.3 Program/Data Memory

The Program/Data Memory consists of both nonvolatile memory (Flash, EEPROM) and RAM memory benches.
Non-volatile memory (NVM) is mainly used to store the user program and can also be used to store permanent data (constant, look-up tables).
Each RAM bench consists of 256 locations used to store run-time user data. At least one bench is present in the devices. RAM benches are also used to implement both System and User Stacks.

Figure 3.1 Addressing Spaces


NVM is always located beginning after the first locations of the addressing space. RAM banks are always located after NVM.
NVM is organized in accordance to the following blocks (see Figure 3.2):

- Reset Vector block (from address 0 to 2) contains an absolute jump instruction to the first user program instruction. The Assembler tool automatically fills these locations with correct data.
- Interrupt Vectors block (from location 3 up to 32) contains the interrupt vectors. Each address is composed of three bytes (the jump opcode and the 16 bit address). Interrupt vectors are set by using IRQ pseudo-instruction (see the Programming Manual).

■ Mbfs Setting block (just after the interrupt vectors) contains the coordinates of the vertexes of every Mbf defined in the program. The last address that can be assigned to this block is 1023. This area is dynamically assigned according to the size of the fuzzy routines. The memory area that remains unused, if any, is assigned to the Program Instructions block.
■ The Program Instructions block (just after the last Mbf data through the last NVM address) contains the instruction of the user program and the permanent data.
■ Option bytes block (from location 3000h to 307 Fh ) is the addressing space reserved for the option bytes. In ST52F510/F513/F514, only the location from 3000h to 3007 h are used.

Figure 3.2 Program/Data Memory Organization


Flash and EEPROM are programmed electrically just applying the supply voltage ( 2.7 V to 5.5 V ) and it is also erased electrically; this feature allows the user to easily reprogram the memory without taking the device off from the board (In Situ Programming ISP). Data and commands are transmitted through the $I^{2} \mathrm{C}$ serial communication protocol. Data can also be written run-time with the In Application Programming (IAP)
NVM can be locked by the user during the programming phase, in order to prevent external operation such as reading the program code and assuring protection of user intellectual property. Flash and EEPROM pages can be protected by unintentional writings.
The operations that can be performed on the NVM during the Programming Phase, ISP and IAP are described in detail in the Section 4.

Figure 3.3 System and User Stack

### 3.4 System and User Stacks

The System and User Stacks are located in the Program/Data memory in the RAM benches.
System Stacks are used to push the Program Counter (PC) after an Interrupt Request or a Subroutine Call. After a RET (Return from a subroutine) or a RETI (Return from an interrupt) the PC that is saved is popped from the stack and restored. After an interrupt request, the flags are also saved in a reserved stack inside the core, so each interrupt has its own flags.
The System Stack is located in the last RAM bench starting from the last address (255) inside the bench page. The System Stack Pointer (SSP) can be read and modified by the user. For each level of stack 2 bytes of the RAM are used. The SSP points to the first currently available stack position. When a subroutine call or interrupt request occurs, the content of the PC is stored in a couple of locations pointed to by the SSP that is decreased by 2.


When a return occurs (RET or RETI instruction), the SSP is increased by 2 and the data stored in the pointed locations couple is restored back into the PC.
The current SSP can be read and write in the couple of Configuration Registers 44 02Ch (MSB: page number, always 32020 h ) and 45 02Dh (LSB: location address) (see Figure 3.3). In ST52F510/ F513/F514 the user can only consider the LSB because the MSB is always the same.
The User Stack is used to store user data and is located beginning from a RAM bench location set by the user (USTP) by writing the couple of Configuration Registers 5 005h (MSB: page number) and 6005 h (LSB: location address) (see Figure 3.3). Register 5, which is the page number, must always be set to a value between 32 (020h) and 255 ( 0 FFh ): values higher than 32 always address RAM on page 32.

Note: In ST52F510/F513/F514 MSB doesn't have to be set or read because the RAM is only 256 bytes. The LSB of the user stack is equal to 0 at reset. The LSB of the system stack is equal to 255 at reset.

This feature allows a flexible use of the User Stack in terms of dimension and to avoid overlaps. The User Stack Pointer (USP) points to the first currently available stack location. When the user stores a byte value contained in the Register File by using the PUSH instruction, the value is stored in the position pointed to by the USP that is increased (the User Stack order is opposite to the System Stack one). When the user takes a value from the User Stack with the POP instruction, the USP is decreased and the value pointed to is copied in the specified Register File location.
By writing the USTP, the new address is automatically written in the USP. The current USP can be read from the Input Registers 11 OBh (MSB: page number, always 32 020h) and 120 Ch (LSB: location address) (see Figure 3.3). In ST52F510/F513/F514 the user can only consider the LSB because the MSB is always the same.
Note: The user must pay close attention to avoid overlapping user and Stacks data. The User Stack Top location and the System Stack Pointer should be configured with care in order to have enough space between the two stacks.

### 3.5 Input Registers

The ST52F510/F513/F514 Input Registers bench consists of a file of 8 -bit registers containing data or the status of the peripherals. For example, the

Input Registers contain data converted by the ADC, Ports, serial communication peripherals, Timers, etc.
The Input Registers can be accessed by using the LDRI instruction that loads the specified Register File address with the contents of the specified Input Register. See the Programming Manual for further details on this instruction. The Input Registers are read-only registers.
In order to simplify the concept, a mnemonic name is assigned to each register. The same name is used in Visual FIVE development tools. The list of the Input Registers is shown in Table 3.1.

### 3.6 Output registers

The ST52F510/F513/F514 Output Registers bench consists of a file of 8 -bit registers containing data sent to the Peripherals and the I/O Ports (for example: Timer Counters, data to be transmitted by the serial communication peripherals, data to be sent to the Port pins in output, etc.).
The registers are located inside the Peripherals and Ports, which allow flexibility and modularity in the design of new family devices.
The Output Registers are write only. In order to access the configuration Register the user can use the following instructions:

- LDPI: loads the immediate value in the specified Output Register.
■ LDPR: loads the contents of the specified Register File location into the output register specified. This instruction allows computed data to be sent to Peripherals and Ports.
- LDPE direct: loads the contents of the specified Program/Data Memory location into the output register specified. This instruction allows data to be sent to Peripherals and Ports from a table.
- LDPE indirect: loads the contents of the Program/Data Memory location whose address is contained in the specified Register File location into the output register specified. This instruction allows data to be sent to Peripherals and Ports from a table pointed to by a register.
See the Programming manual for further details about these instructions.
In order to simplify the concept, a mnemonic name is assigned to each register. The same name is used in Visual FIVE development tools. The list of the Output Registers is shown in Table 3.2.
3.7 Configuration Registers \& Option Bytes

The ST52F510/F513/F514 Configuration
Registers bench consists of a file of 8 -bit registers that allows the configuration of all the ICU blocks. The registers are located inside the block they configure in order to obtain greater flexibility and modularity in the design of new family devices. In the Configuration Registers, each bit has a peculiar use, so the logic level of each of them must be considered.
Some special configuration data, that needs to be load at the start-up and not further changed, are stored in Option Bytes. These are loaded only during the device programming phase. See Table 3.3 and Section 4 for a detailed description of the Option Bytes.
The Configuration Registers are readable and writable; the addresses refer to the same register both in read and in write. In order to access the Configuration Register the user can work in several modes by utilizing the following instructions:

■ LDCI: loads the immediate value in the Configuration Register specified and is the most commonly used to write configuration data.
■ LDCR: loads the Configuration Register specified with the contents of the specified Register File location, allowing a parametric configuration.
■ LDCE: loads the Configuration Register specified with the contents of the specified Program/Data Memory location, allowing the configuration data to be taken from a table.

- LDCNF: loads the Register File location specified with the contents of the Configuration Register indicated, allowing for the inspection of the configuration of the device (permitting safe run-time modifications).
In order to simplify the concept, a mnemonic name is assigned to each register. The same name is used in Visual FIVE development tools. The list of the Configuration Registers is shown in Table 3.4.

Table 3.1 Input Registers

| Mnemonic | Description | Address |  |
| :---: | :---: | :---: | :---: |
| PORT_A_IN | Port A data Input Register | 0 | 00h |
| PORT_B_IN | Port B data Input Register | 1 | 01h |
| PORT_C_IN | Port C data Input Register | 2 | 02h |
| - | Not Used | 3 | 03h |
| - | Not Used | 4 | 04h |
| SPI_IN | Serial Peripheral Interface data Input Register | 5 | 05h |
| I2C_IN | $\mathrm{I}^{2} \mathrm{C}$ Interface data Input Register | 6 | 06h |
| I2C_SR1 | $\mathrm{I}^{2} \mathrm{C}$ Interface Status Register 1 | 7 | 07h |
| I2C_SR2 | $\mathrm{I}^{2} \mathrm{C}$ Interface Status Register 2 | 8 | 08h |
| - | Not Used | 9 | 09h |
| - | Not Used | 10 | 0Ah |
| USP_H | User Stack Pointer (MSB) | 11 | 0Bh |
| USP_L | User Stack Pointer (LSB) | 12 | 0Ch |
| - | Not Used | 13-20 | $\begin{aligned} & \text { 0Dh- } \\ & 014 \mathrm{~h} \end{aligned}$ |
| PWMO_COUNT_IN_H | PWM/Timer 0 Counter Input Register (MSB) | 21 | 015h |
| PWMO_COUNT_IN_L | PWM/Timer 0 Counter Input Register (LSB) | 22 | 016h |
| PWMO_STATUS | PWM/Timer 0 Status Register | 23 | 017h |
| PWMO_CAPTURE_H | PWM/Timer 0 Capture Register (MSB) | 24 | 018h |
| PWMO_CAPTURE_L | PWM/Timer 0 Capture Register (LSB) | 25 | 019h |
| PWM1_COUNT_IN_H | PWM/Timer 1 Counter Input Register (MSB) | 26 | 01Ah |
| PWM1_COUNT_IN_L | PWM/Timer 1 Counter Input Register (LSB) | 27 | 01Bh |
| PWM1_STATUS | PWM/Timer 1 Status Register | 28 | 01Ch |
| PWM1_CAPTURE_H | PWM/Timer 1 Capture Register (MSB) | 29 | 01Dh |
| PWM1_CAPTURE_L | PWM/Timer 1 Capture Register (LSB) | 30 | 01Eh |
| - | Not Used | 31-35 | $\begin{aligned} & \text { 01Fh- } \\ & \text { 023h } \end{aligned}$ |
| SCI_IN | Serial Communication Interface RX data Input Register | 36 | 024h |
| SCI_STATUS | Serial Communication Interface Status Register | 37 | 025h |
| FLAGS | Flag Register | 38 | 026h |
| AD_OVF | 10-bit A/D Converter Overflow Register | 39 | 027h |
| IAP_SR | In Application Programming Status Register | 40 | 028h |

Table 3.1 Input Registers

| Mnemonic | Description | Address |  |
| :---: | :---: | :---: | :---: |
| CHANO_H | 10-bit A/D Converter Channel 0 data Input Register (MSB) | 41 | 029h |
| CHANO_L | 10-bit A/D Converter Channel 0 data Input Register (LSB) | 42 | 02Ah |
| CHAN1_H | 10-bit A/D Converter Channel 1 data Input Register (MSB) | 43 | 02Bh |
| CHAN1_L | 10-bit A/D Converter Channel 1 data Input Register (LSB) | 44 | 02Ch |
| CHAN2_H | 10-bit A/D Converter Channel 2 data Input Register (MSB) | 45 | 02Dh |
| CHAN2_L | 10-bit A/D Converter Channel 2 data Input Register (LSB) | 46 | 02Eh |
| CHAN3_H | 10-bit A/D Converter Channel 3 data Input Register (MSB) | 47 | 02Fh |
| CHAN3_L | 10-bit A/D Converter Channel 3 data Input Register (LSB) | 48 | 030h |
| CHAN4_H | 10-bit A/D Converter Channel 4 data Input Register (MSB) | 49 | 031h |
| CHAN4_L | 10-bit A/D Converter Channel 4 data Input Register (LSB) | 50 | 032h |
| CHAN5_H | 10-bit A/D Converter Channel 5 data Input Register (MSB) | 51 | 033h |
| CHAN5_L | 10-bit A/D Converter Channel 5 data Input Register (LSB) | 52 | 034h |
| CHAN6_H | 10-bit A/D Converter Channel 6 data Input Register (MSB) | 53 | 035h |
| CHAN6_L | 10-bit A/D Converter Channel 6 data Input Register (LSB) | 54 | 036h |
| CHAN7_H | 10-bit A/D Converter Channel 7 data Input Register (MSB) | 55 | 037h |
| CHAN7_L | 10-bit A/D Converter Channel 7 data Input Register (LSB) | 56 | 038h |

## Table 3.2 Output Registers

| Mnemonic | Description | Address |  |
| :--- | :--- | :---: | :---: |
| PORT_A_OUT | Port A data Output Register | 0 | 00 h |
| PORT_B_OUT | Port B data Output Register | 1 | 01 h |
| PORT_C_OUT | Port C data Output Register | 2 | 02 h |
| - | Not Used | 3 | 03 h |
| - | Not Used | 4 | 04 h |
| SPI_OUT | Serial Peripheral Interface data Output Register | 5 | 05 h |
| I2C_OUT | I $^{2} \mathrm{C}$ Interface data Output Register | 6 | 06 h |
| PWMO_COUNT_OUT_H | PWM/Timer 0 Counter Output Register (MSB) | 7 | 07 h |
| PWMO_COUNT_OUT_L | PWM/Timer 0 Counter Output Register (LSB) | 8 | 08 h |
| PWMO_RELOAD_H | PWM/Timer 0 Reload Register (MSB) | 9 | 09 h |
| PWMO_RELOAD_L | PWM/Timer 0 Reload Register (LSB) | 10 | 0 Ah |
| PWM1_COUNT_OUT_H | PWM/Timer 1 Counter Output Register (MSB) | 11 | 0Bh |
| PWM1_COUNT_OUT_L | PWM/Timer 1 Counter Output Register (LSB) | 12 | 0 Ch |
| PWM1_RELOAD_H | PWM/Timer 1 Reload Register (MSB) | 13 | 0 Dh |
| PWM1_RELOAD_L | PWM/Timer 1 Reload Register (LSB) | 14 | 0 Eh |
| SCI_OUT | Serial Communication Interface TX data Output Register | 23 | 017 h |

Table 3.3 Option Bytes

| Mnemonic | Description | Address |  |
| :--- | :--- | :---: | :---: |
| OSC_CR | Oscillator Control Register | 0 | 00 h |
| CLK_SET | Clock Parameters | 1 | 01 h |
| OSC_SET | Oscillator Set-Up | 2 | 02 h |
| PLDV_CR | Programmable Low Voltage Detector Control Register | 3 | 03 h |
| WDT_EN | HW/SW Watchdog selector | 4 | 04 h |
| PG_LOCK | First Page Write Protected | 5 | 05 h |
| PG_UNLOCK | First Page not Write Protected | 6 | 06 h |
| WAKEUP | Wake Up from Halt Time | 7 | 07 h |

## Table 3.4 Configuration Registers

| Mnemonic | Description | Address |  |
| :---: | :---: | :---: | :---: |
| INT_MASK | Interrupt Mask Register | 0 | 00h |
| INT_POL | Interrupts Polarity | 1 | 01h |
| INT_PRL_H | Interrupt Priority Register (higher priority) | 2 | 02h |
| INT_PRL_M | Interrupt Priority Register (medium priority) | 3 | 03h |
| INT_PRL_L | Interrupt Priority Register (lower priority) | 4 | 04h |
| USTP_H | User Stack Top Pointer (MSB) | 5 | 05h |
| USTP_L | User Stack Top Pointer (LSB) | 6 | 06h |
| WDT_CR | Watchdog Configuration Register | 7 | 07h |
| AD_CR1 | 10-bit A/D Converter Control Register 1 | 8 | 08h |
| PWM0_CR1 | PWM/Timer 0 Configuration Register 1 | 9 | 09h |
| PWM0_CR2 | PWM/Timer 0 Configuration Register 2 | 10 | 0Ah |
| PWM0_CR3 | PWM/Timer 0 Configuration Register 3 | 11 | 0Bh |
| PWM1_CR1 | PWM/Timer 1 Configuration Register 1 | 12 | 0Ch |
| PWM1_CR2 | PWM/Timer 1 Configuration Register 2 | 13 | 0Dh |
| - | Not Used | 14 | 0Eh |
| - | Not Used | 15 | 0Fh |
| 12C_CR | $\mathrm{I}^{2} \mathrm{C}$ Interface Control Register | 16 | 010h |
| I2C_CCR | $1^{2} \mathrm{C}$ Interface Clock Control Register | 17 | 011h |
| I2C_OAR1 | $1^{2} \mathrm{C}$ Interface Own Address Register 1 | 18 | 012h |
| I2C_OAR2 | $I^{2} \mathrm{C}$ Interface Own Address Register 2 | 19 | 013h |
| SPI_CR | Serial Peripheral Interface Control Register | 20 | 014h |
| SPI_STATUS_CR | Serial Peripheral Interface Control-Status Register | 21 | 015h |
| SCI_CR1 | Serial Communication Interface Control Register 1 | 22 | 016h |
| SCI_CR2 | Serial Communication Interface Control Register 2 | 23 | 017h |
| PORT_A_PULLUP | Port A Pull Up enable/disable Register | 24 | 018h |
| PORT_A_OR | Port A Option Register | 25 | 019h |

Table 3.4 Configuration Registers

| Mnemonic | Description | Address |  |
| :---: | :---: | :---: | :---: |
| PORT_A_DDR | Port A Data Direction Register | 26 | 01Ah |
| PORT_A_AF | Port A Alternate Function selection Register | 27 | 01Bh |
| PORT_B_PULLUP | Port B Pull Up enable/disable Register | 28 | 01Ch |
| PORT_B_OR | Port B Option Register | 29 | 01Dh |
| PORT_B_DDR | Port B Data Direction Register | 30 | 01Eh |
| PORT_B_AF | Port B Alternate Function selection Register | 31 | 01Fh |
| PORT_C_PULLUP | Port C Pull Up enable/disable Register | 32 | 020h |
| PORT_C_OR | Port C Option Register | 33 | 021h |
| PORT_C_DDR | Port C Data Direction Register | 34 | 022h |
| PORT_C_AF | Port C Alternate Function selection Register | 35 | 023h |
| - | Not Used | 36-42 | $\begin{aligned} & \text { 024h- } \\ & \text { 02Ah } \end{aligned}$ |
| SCI_CR3 | Serial Communication Interface Control Register 3 | 43 | 02Bh |
| SSP_H | System Stack Pointer (MSB) | 44 | 02Ch |
| SSP_L | System Stack Pointer (LSB) | 45 | 02Dh |
| CPU_CLK | CPU Clock Prescaler | 46 | 02Eh |
| AD_CR2 | 10-bit A/D Converter Control Register 2 | 47 | 02Fh |

## 4 MEMORY PROGRAMMING

ST52F510/F513/F514 provides an on-chip user programmable non-volatile memory, which allows fast and reliable storage of user data.
Program/Data Memory addressing space is composed by a Single Voltage Flash Memory and a RAM memory bench. The ST52F513/514 devices also have a Data EEPROM bench to store permanent data with long term retention and a high number of write/erase cycles.
All the Program Data memory addresses can execute code, including RAM and EEPROM benches.
The memory is programmed by setting the $\mathrm{V}_{\mathrm{pp}}$ pin equal to $\mathrm{V}_{\text {dd }}$. Data and commands are transmitted through the $\mathrm{I}^{2} \mathrm{C}$ serial communication protocol. The same procedure is used to perform "In-Situ" the programming of the device after it is mounted in the user system. Data can also be written in runtime with the In-Application Programming (IAP).
The Memory can be locked by the user during the programming phase, in order to prevent external operation such as reading the program code and assuring protection of user intellectual property. Flash and EEPROM pages can be protected by unintentional writings.

Remark: the memory contents are protected by the Error Correction Code (ECC) algorithm that uses a 4-bit redundancy to correct one bit errors.
Warning: when entering the ISP, the default values for Option Bytes are considered, so a Voltage Supply higher than the PLVD lower threshold must be applied to program the device.

### 4.1 Program/Data Memory Organization

The Program/Data Memory is organized as described in Section 3.3. The various sales types have different amounts of each type of memory. Table 4.1 describes the memory benches amount and page allocation for each sales type.
The addressing spaces are organized in pages of 256 bytes. Each page is composed by blocks of 32 bytes. Memory programming is performed one block at a time in order to speed-up the programming time (about 2.5 ms per block).
The whole location address is composed as follows:

15
$\begin{array}{llll}8 & 7 & 5 & 4\end{array}$
0


Table 4.1 Sales Type Memory Organization

| Device | Flash Memory |  | RAM Memory |  | EEPROM Memory |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | Amount | Pages | Amount | Page | Amount | Page(s) |
| ST52F510c2p6 | 4096 bytes | 0 to 15 | 256 bytes | 32 | - | - |
| ST52F510c3p6 | 8192 bytes | 0 to 31 | 256 bytes | 32 | - | - |
| ST52F513c2p6 | 3840 bytes | 0 to 14 | 256 bytes | 32 | 256 bytes | 15 |
| ST52F513c3p6 | 7936 bytes | 0 to 30 | 256 bytes | 32 | 256 bytes | 31 |
| ST52F514c1p6 | 4096 bytes | 0 to 15 | 256 bytes | 32 | 1024 bytes | $16-19$ |
| ST52F514c3p6 | 4096 bytes | 0 to 15 | 256 bytes | 32 | 4096 bytes | $16-31$ |

legend:
c: $Y=16$ pins, $F=20$ pins, $G=28$ pins, $K=32 / 34$ pin
$\mathrm{p}: \mathrm{B}=\mathrm{DIP}, \mathrm{M}=\mathrm{SO}, \mathrm{T}=\mathrm{TQFP}$

### 4.2 Memory Programming

The Programming procedure writes the user program and data into the Flash Memory, EEPROM and Option Bytes. The programming procedures are entered by setting the $\mathrm{V}_{\mathrm{pp}}$ pin equal to $\mathrm{V}_{\mathrm{dd}}$ and releasing the Reset signal. The following pins are used in Programming mode:
■ $V_{\text {PP }} \quad$ used to switch to programming mode

- $V_{D D}$ device supply
- $V_{S S}$ device ground
- RESET device reset
- SCL $\mathrm{I}^{2} \mathrm{C}$ serial clock
- SDA $\quad \mathrm{I}^{2} \mathrm{C}$ serial data

During the device programming, the internal clock is used, so the OSCin and OSCout pins don't have to be considered.
4.2.1 Programming Mode start. The following sequence starts the Programming Mode:

1. $V_{P P}$ is set to $V_{D D}$
2. The device is Reset ( $\mathrm{RESET}=\mathrm{V}_{S S}$ )
3. The Reset is released ( $\mathrm{RESET}=\mathrm{V}_{\mathrm{DD}}$ )
4. The internal oscillator starts at 10 MHz
5. The memory is turned on
6. The $\mathrm{I}^{2} \mathrm{C}$ Interface and Ports are initialized
7. The $I^{2} C$ Interface is configured to work as Slave, Receiver, 7 -bit address and waits for data
8. The Start signal is sent to the chip followed by the Slave Address 1010000 and the direction bit set to 0 (the addressed slave waits for data). The device sends the acknowledge
9. The Programming Mode code 00000000 is sent and acknowledged
10. A command code is sent to the device
11. The procedure related to the command is executed

Table 4.2 Programming Mode Commands

| Command | Code | Data in | Data out | Erase | Description |
| :--- | :---: | :---: | :---: | :---: | :--- |
| BlockWrite | 00000001 | 32 | - | Yes | Write the currently addressed block with the 32 bytes <br> following the command. The Block locations are erased <br> before being written. |
| ByteWrite | 00000010 | 2 | - | Yes | Write the byte addressed by the next data sent in the <br> currently addressed page. |
| BlockErase | 00000011 | 1 |  | Yes | Erase the block addressed by 3 MSB of the next data sent <br> and inside the currently addressed page. |
| ByteErase | 00000100 | 1 |  | Yes | Erase the byte addressed by the next data sent and inside <br> the currently addressed page. |
| ByteRead | 00000101 | 1 | 1 | - | Read the byte addressed by the next data sent and inside <br> the current page. The read data is sent by the device after <br> the re-send of the Slave Address with the R/W bit changed. |
| GlobalErase | 00001001 | - | - | Yes | All the memory is erased. <br> FastBlockWrite 00001011 |
| 32 | - | No | Write the currently addressed block with the 32 bytes <br> following the command. The Block locations aren't erased. |  |  |
| SetPage | 00001100 | 1 | - | - | The currently addressed page is set with the next data sent. |$|$| ReadData |
| :--- |
| 00001101 |
| - |

Figure 4.1 Commands and Data Communication Sequences

## Programming mode start sequence

| S | 10100000 | A | 00000000 | A | Command | A | Data1 | A | $\ldots .$. | DataN | A | P |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Execution of commands for writing data:

| Command | A | Data1 | A | ..... | DataN | A | Command | A | Data1 | A | $\ldots .$. | DataN | A | P |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Execution of commands for reading data:

| Command | A | Address | A | P | S | 10100001 | A | Data read | NA | P |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

S=Start, $\mathrm{P}=$ Stop, $\mathrm{A}=$ Acknowledge, $\mathrm{NA}=$ Non-acknowledge
From Slave to Master $\quad \square \quad$ From Master to Slave

The generic procedure of commands execution, with the data communication in both directions is displayed in Figure 4.1.

Remark: the Slave Address 1010000 must be sent after a Stop (i.e. each time the data direction changes, to specify the $R / W$ bit). For example: if a command to send data to the device has been executed, a command for receiving data must be followed by the slave address and the R/W bit must be set to 1. The Programming Mode code doesn't need to be specified again .

Warning: After entering the Programming Mode, the currently pointed address is the Page 48, Block 3, byte 0 (Lock Byte).
The list of the available commands in Programming Mode is showed in Table 4.2
4.2.2 Fast Programming procedure. The fastest way to program the device memory is the use of the FastBlockWrite command. The following procedure can be used to write the memory with a new program and new data, starting from the first memory location:

1. The Programming Mode is entered with the sequence described above
2. The memory is erased (all bits are put to 0) with the GlobalErase command. The device holds the SCL line low, releasing it after the command is completed (about 2 ms ). This command also unlocks the device if locked.
3. The FastBlockWrite command is sent and the device acknowledges it
4. The 32 bytes of data to be written in the first memory Block are sent in a sequence. The device acknowledges each of them
5. After the device acknowledges the 32nd byte, it holds the SCL line until the parallel writing of the 32 byte is completed (about 2.5 ms )
6. The Block Pointer is incremented by sending the IncBlock command
7. The procedure is repeated from point 3 until there is data to be sent to the memory

Note: the Block Pointer assumes values between 0 to 7 (there are 8 blocks in a page). When the Block Pointer is equal to 7 , the IncBlock command puts this pointer to 0 and increments the Page Pointer. The Page Pointer, after page writing is completed, does't have to be incremented in the procedure above described.
4.2.3 Random data writing. A single byte can be written in a specified memory location by using the following procedure:

1. The Programming Mode is entered with the sequence described in Section 4.2.1
2. The SetPage command is sent, followed by the page number where the data should be written
3. The ByteWrite command is sent followed by two bytes
4. The first bytes that follows the ByteWrite command is the address inside the pointed page where the data must be written.
5. The second byte is the data to be written
6. The device held the SCL line low until the data is not stored in the memory (about 4.5 ms : 2 ms for erasing and 2.5 for writing)

A similar procedure can be used to write a single block:

1. The SetPage command is sent, followed by the page number where the data should be written
2. The IncBlock command is sent as many times as the block number inside the page (for example: to address the block 3 the IncBlock must be sent 3 times)
3. The WriteBlock command is sent followed by the 32 data bytes to be written.
4. After the 32th byte is sent, the device holds the SCL line low until all the data are not stored in the memory (about $4.5 \mathrm{~ms}: 2 \mathrm{~ms}$ for erasing and 2.5 for writing: the same time for a single byte)
The procedures described previously can be repeated as many time as needed, without exiting from Programming Mode or re-sending the Slave Address again.

The commands ByteErase and BlockErase, used instead of ByteWrite and BlockWrite, erase (put all bit to 0 ) the specified memory location or block.
4.2.4 Option Bytes Programming. The Option Byte addresses cannot be accessed with a sequential procedure like the one described in Section 4.2.2. Actually, the pointers are automatically incremented up to the last block or address in page 31. A further increment sets all the pointers to 0 .
The Option Byte addresses (located at page 48, block 0, addresses 0-7) must be accessed with a direct addressing procedure as the one described in Section 4.2.3.
If the Fast Programming procedure is used, it must be followed by a Random Block Writing procedure to program the Option Bytes. The other 24 bytes of the block can be written with dummy or user values. The blocks $0,1,2$ and 3 of Page 48 can be used for writing data as well (see Section 4.5) and for locking the device (see Section 4.4).

Figure 4.2 Programming Procedures

## Fast Programming Procedure



Random Byte Writing Procedure

| SetPage | A | Page Address | A | ByteWrite | A | Byte Address | A | Data | A | Command | ..... |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Random Block Writing Procedure

| SetPage | A | Page Address | A | IncBlock | A |  | IncBlock |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |



Option Byte Writing Procedure


$\cdots$| Dummy 0 | A | ..... | Dummy 23 | A |
| :--- | :--- | :--- | :--- | :--- |

S=Start, $\mathrm{P}=$ Stop, $\mathrm{A}=$ Acknowledge, $\mathrm{NA}=$ Non-acknowledge
$\square$ From Slave to Master $\quad \square \quad$ From Master to Slave

Figure 4.3 Reading and Erasing Procedures
Fast Reading Procedure

| S | 10100000 | A | 00000000 | A | ReadData | A | P | S | 10100001 | A | Data read | NA | P |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |



Random Byte Reading Procedure


Byte Erasing Procedure


## Block Erasing Procedure


(*) Block address is specified by the 3 most significative bits of the whole given address (less significative bits are don't care) $_{\text {( }}$
S=Start, $P=$ Stop, $A=A c k n o w l e d g e, ~ N A=N o n-a c k n o w l e d g e ~$
From Slave to Master $\quad \square \quad$ From Master to Slave

### 4.3 Memory Verify

To verify the memory contents or just to read part of data stored in memory, the ByteRead and the ReadData command can be used. The first instruction needs the specification of the address; the second one allows the sequential reading of consecutive memory locations.
Since the device is "Slave" for the $I^{2} C$ protocol, after receiving a command for reading, it must be configured as Slave Transmitter to send the data. In order to do so, the Slave Address (1010000) must be sent again with the R/W byte set to 1 , as stated by the communication protocol.
4.3.1 Fast read procedure. The memory can be read sequentially by using the following procedure:

1. The Programming mode is entered with the sequence described in Section 4.2.1
2. The pointers address the memory location 0
3. The ReadData command is sent and the device acknowledges it.
4. The Master generates a Stop condition followed by a Start condition
5. The Slave Address with the R/W byte set to 1 (10100001) is sent. The device receives the Slave Address and acknowledges it.
6. The device sends the data to be read in the serial data line SDA. The current absolute address is post-incremented.
7. The Master device doesn't send the acknowledge and generates a stop condition.
8. To read the next data, the Master generates a Start condition followed by the Slave Address with the R/W byte set to 0 (10100000). The device receives the Slave Address and acknowledges it.
9. The sequence restarts from point 3 until there is data to be read.

Remark: for the same reasons explained in Section 4.2.4 the Option Bytes cannot be read with this procedure: they can be read with a direct addressing procedure as the one explained in the next section.
4.3.2 Random data reading. To read a specified memory location, the following procedure should be used:

1. The Programming mode is entered with the sequence described in Section 4.2.1
2. The SetPage command is sent, followed to the page number where the data to be read is located
3. The ByteRead command is sent, followed by an address inside the page
4. The Master generates a Stop condition followed by a Start condition
5. The Slave Address with the R/W byte set to 1 (10100001) is sent. The device receives the Slave Address and acknowledges it.
6. The device sends the data to be read in the serial data line SDA.
7. The Master device doesn't send the acknowledge and generates a stop condition.
8. To send the next command, the Master should generate a Start condition followed by the Slave Address with the R/W byte set to 0 (10100000).

### 4.4 Memory Lock

The Program/Data Memory space can be locked to inhibit the reading of contents and protect the intellectual property.
To lock the device, the user must set all the bit of the Lock Byte to ' 1 '. The Lock Byte is located on Page 48 (030h), Block 3, byte 0 inside the block i.e. byte 96 ( 060 h ) inside the page.
After writing 255 (0FFh) into the Lock Byte, with the procedure described in the Section 4.2.3, the memory is locked and the only command allowed are the following:

- GlobalErase: this command, writing ' 0 ' in all the memory, also unlock the device.
- ReadData: the only block that can be read is the Block 3 in Page 48 (030h); this allows the reading of the Lock Byte and the ID Code locations (see Section 4.5).
- ReadStatus: this command allows the detection of an error condition in Programming mode operation (see Section 4.6). It can also be used to check if the device is locked. The most significative bit return the Lock Bit ( $0=$ unlocked, $1=$ locked).

Remark: the Lock Byte is checked when entering the Programming Mode. For this reason after writing the Lock Byte, all the commands can be carried out until the Programming mode is exited.

Figure 4.4 Device Lock Procedure

## Device Lock Procedure

$\ldots$| SetPage | A | 00110000 | A | ByteWrite | A | 01100000 | A | 11111111 | A | Command |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| $\ldots$ |  |  |  |  |  |  |  |  |  |  |$.. .$.

Device Lock and ID Code Writing Procedure


Device Lock Reading Procedure

(*) $^{*}$ The most significative bit return the Lock Bit ( $0=$ unlocked, $1=$ locked)

S=Start, $\mathrm{P}=$ Stop, $\mathrm{A}=$ Acknowledge, $\mathrm{NA}=$ Non-acknowledgeFrom Slave to Master $\square$From Master to Slave

Figure 4.5 Error Handling Procedure
Wrong command/data case handling:


S=Start, $\mathrm{P}=$ Stop, $\mathrm{A}=$ Acknowledge, $\mathrm{NA}=$ Non-acknowledge

From Slave to Master $\quad \square \quad$ From Master to Slave

When the device is locked, if memory reading is attempted, with the exception of the Lock Byte and ID Code block, the device returns no data and an error sequence. If memory writing is attempted in any memory location, the device doesn't carry out the command and returns an error sequence.
To unlock the device the GlobalErase command must be executed before any writing or reading command.

### 4.5 ID Code

Block 3 on Page 48 (030h) can also be read if the device is locked. The first byte of the block is the Lock Byte, the other 31 locations are available to the user for writing data, as for example identification codes to distinguish the firmware version loaded in the device.
The ID Code must be written before locking the device: after the device is locked it can only be read. The use of the Block writing procedure is the fastest way: both the ID Code and the Lock bytes are written together. The lock Bytes are sent first and then the 31 bytes of ID Code follow.
The blocks 0,1 and 2 on Page 48 can be also be used for writing data, but they cannot be accessed when the device is locked.
Note: the ID Code cannot be modified if the device is locked: it can only be read.

### 4.6 Error cases

If a wrong command or data is sent to the device, it generates an error condition by not sending the acknowledge after the first successive data or command. Figure 4.5 shows the error sequence.
The error case can be handled by using the ReadStatus command. This command can be sent after the error condition is detected; the device returns a Status Byte containing the error code. The ReadStatus command sequence is showed in Figure 4.5. The list of the error codes is illustrated in Table 4.3.

Remark: after the ReadStatus command execution or after any error, the Start Sequence must be carried out before sending a new command.

The Most Significative Bit of the error codes indicates (when set to ' 1 ') that the memory is locked. When a command, that is not allowed when the memory is locked, is sent, the "Not Allowed" code is sent. If another code is sent with the MSB to ' 1 ' it indicates that the error condition is not caused by the memory lock, but by the event related with the code sent.

Warning: when the data writing into a non existing location is attempted, no error condition is generated. The user must take care in specifying the correct page address.

Table 4.3 Error codes

| Name | Code | Description |
| :--- | :---: | :--- |
| Device Locked | xyyyyyyy | x=lock bit (1=device locked), yyyyyyy=error code |
| Wrong Direction | $x 0000001$ | A transmit direction, not correct in the running sequence, has been set |
| Stop Missed | $x 0000010$ | The Master missed generating a necessary Stop Condition |
| Data Missing | $x 0000011$ | The Master missed to send necessary data to the device |
| Receive Error | $x 0000100$ | The data sent by the Master hasn't been received correctly by the device |
| Wrong Command | $\times 0000101$ | The Master sent a wrong command code |
| Not Allowed | $x 0000110$ | A command not allowed when the device is locked has been sent |
| Wrong Mode | $x 0010000$ | A code different form the Programming mode code (00000000) has been sent |

### 4.7 In-Situ Programming (ISP)

The Program/Data Memory can be programmed using the ISP mode. This mode allows the device to be programmed when it is mounted in the user application board.
This feature can be implemented by adding a minimum number of components and board impact.
The programming procedures and pins used are identical to the ones described before for the standard Programming Mode. All the features previously described in this chapter are applicable in ISP mode.
If RESET, SCL and SDA pins are used in the user application board for other purposes, it is recommended to use a serial resistor to avoid a conflict when the other devices force the signal level.
The ISP can be applied by using the standard tools for the device programming. The ISP can be applied by using the standard tools for the device programming. The ST52F510 Starter Kit supplies a cable to perform the ISP. The user application board should supply a suited connector type for the cable (see Starter Kit User Manual).

### 4.8 In-Application Programming (IAP)

The In Application Programming Mode (IAP) allows the writing of user data in the Flash and EEPROM memories when the user program is running.
There are two ways to write data in IAP mode: single byte write and Block write. Both procedures take about 4.5 ms to complete the writing: the Block write allows the writing of 32 byte in parallel.
Remark: during data writing, the execution of the user program is stopped until the procedure is completed. Interrupt requests stop the writing operation and the data may be not stored. The bit ABRT in the IAP_SR Input register signals that the data writing hasn't been completed. To assure writing completion, the user should globally disable the interrupts (UDGI instruction) before starting IAP data writing.
4.8.1 Single byte write. Writing of a single byte in the Non-Volatile Program/Data memory is performed by using the LDER instruction (both direct and indirect addressing). The memory page should be indicated before the LDER instruction with the PGSET or PGSETR instruction. The byte address inside the page is specified by the LDER instruction itself.
As soon as the instruction is executed, the data writing starts and is performed in about 4.5 ms .
4.8.2 Block write. This procedure allows the writing of 32 bytes in parallel. These bytes should belong to the same block.
Before the writing in the Program/Data memory, data must be buffered in the Register File in the first 32 locations ( $0-31,00 h-020 \mathrm{~h}$ ) by using the normal instructions to load the Register File locations.
Then the data writing starts by using the BLKSET instruction. The destination block is addressed by specifying the memory page with the PGSET or PGSETR instruction before to start the writing; the block inside the page is addressed with the argument of the BLKSET instruction.
Example:

## PGSET 5

BLKSET 4
This instruction sequence writes the contents of the first 32 bytes of the Register File in the locations 1408-1439 (0580h-059Fh).
Warning: the user should be careful in specifying the correct page and block: the addressing of an not existing block can cause the unwanted writing of a different block.

As soon as the BLKSET instruction is executed, the data writing starts and is performed in about 4.5 ms .

This procedure may also be used to write few data, taking in account that all the 32 byte are written in the block anyway.

### 4.8.3 Memory Corruption Prevention.

The user can protect some pages (or all the memory) from unintentional writings. The only constraint is that the protected pages must be consecutive.
Two Option Bytes allow the specification of the page to be protected: PG_LOCK (Option Byte 5) and PG_UNLOCK (Option Byte 6). PG_LOCK is used to specify the first protected page; PG_UNLOCK is used to specify the first page not protected after the protected ones. The pages between the two addresses are protected.
When writing in a protected page is attempted, the procedure is aborted and the bit PRTCD of IAP_SR Input register is set.
If the PG_LOCK and PG_UNLOCK have the same value, no page is protected. By default, the two Option Bytes are programmed with the value 0 , so the memory is not write protected by default.
In Programming Mode the protection is not considered and the pages can be written unless the device is locked.

### 4.8.4 Option Bytes.

First Protected Page (PG_LOCK)
Option Byte 5 (05h)
Reset Value: 00000000 (00h)

| 7 |  |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :---: |
| LCK7 | LCK6 | LCK5 | LCK4 | LCK3 | LCK2 | LCK1 | LCK0 |

Bit 7-0: LCK7-0 First Page write protected
In this register the address of first page to be protected in writing is specified. The pages following this one are protected up to the page specified by the PG_UNLOCK Option Byte (not included among the protected ones).

First Page not Protected (PG_UNLOCK)
Option Byte 6 (06h)
Reset Value: 00000000 (00h)


## Bit 7-0: UNLCK7-0 First Page not write protected

In this register the address of first page not write protected after the protected ones is specified. The pages following this one aren't protected.

### 4.8.5 Input Register.

IAP Status Register (IAP_SR)
Input Register 40 (028h) Read only
Reset Value: 00000000 (00h)
7
0

| - | - | - | - | - | - | PRTCD | ABRT |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-2: Not Used

## Bit 1: PRTCD Page Protected

0 : The writing has been completed
1: The writing has been aborted because the page is protected.

Bit 0: ABRT Writing operation aborted
0 : The writing has been completed
1: The writing has been aborted because an interrupt or another unspecified cause occurred.

The ABRT and PRTCD bits are reset after the next successful data writing in the Flash of EEPROM memory.

## 5 INTERRUPTS

The Control Unit (CU) responds to peripheral events and external events through its interrupt channels.
When such events occur, if the related interrupt is not masked and doesn't have a priority order, the current program execution can be suspended to allow the CU to execute a specific response routine.
Each interrupt is associated with an interrupt vector that contains the memory address of the related interrupt service routine. Each vector is located in the Program/Data Memory space at a fixed address (see Figure 3.2 Program/Data Memory Organization).

### 5.1 Interrupt Processing

If interrupts are pending at the end of an arithmetic or logic instruction, the interrupt with the highest priority is acknowledged. When the interrupt is acknowledged the flags and the current PC are saved in the stacks and the associated Interrupt routine is executed. The start address of this routine (Interrupt Vector) is located in three bytes of the Program/Data Memory between address 3 and 32 (03h-020h). See Table 5.1 for the list of the Interrupt Vector addresses.
The Interrupt routine is performed as a normal code. At the end of each instruction, the CU checks if a higher priority interrupt has sent an interrupt request. An Interrupt request with a higher priority stops lower priority Interrupts. The Program Counter and the flags are stored in their own stacks.
With the instruction RETI (Return from Interrupt) the flags and the Program Counter (PC) are restored from the top of the stacks. These stacks have already been described in Paragraph 3.4.
An Interrupt request cannot stop fuzzy rule processing, but only after the end of a fuzzy rule or at the end of a logic or arithmetic instruction, unless a Global Interrupt Disable instruction has been executed before (see below).
Remark: A fuzzy routine can be interrupted only in the Main program. When a Fuzzy function is running inside another interrupt routine an interrupt request can cause side effects in the Control Unit. For this reason, in order to use a Fuzzy function inside an interrupt routine, the user MUST include the Fuzzy function between an UDGI (MDGI) instruction and an UEGI (MEGI) instruction (see the following paragraphs), in order to disable the interrupt request during the execution of the fuzzy function.

Figure 5.1 Interrupt Flow


### 5.2 Global Interrupt Request Enabling

When an Interrupt occurs, it generates a Global Interrupt Pending (GIP). After a GIP a Global Interrupt Request (GIR) will be generated and Interrupt Service Routine associated with the interrupt with higher priority will start.
In order to avoid possible conflicts between the interrupt masking set in the main program, or inside high level language compiler macros, the GIP is put in AND through the User Global Interrupt Mask or the Macro Global Interrupt Mask (see Figure 5.2).
The UEGI/UDGI instruction switches the User Global Interrupt Mask enabling/disabling the GIR for the main program.
MEGI/MDGI instructions switch the Macro Global Interrupt Mask on/off in order to ensure that the macro will not be interrupted.

Figure 5.2 Global Interrupt Request


### 5.3 Interrupt Sources

ST52F510/F513/F514 manages interrupt signals generated by the internal peripherals or generated by software by the TRAP instruction or coming from the Port pins. There are two kinds of interrupts coming from the Port pins: the NMI and the Ports Interrupts.
NMI (Not Maskable Interrupt) is associated with pin PA7 when it is configured as Alternate Function. This interrupt source doesn't have a configurable level priority and cannot be masked. The fixed priority level is lower than the software TRAP and higher than all the other interrupts. The NMI can be configured to be active on the rising or the falling edge.
The Port Interrupts sources are connected with Port A and Port B pins. The pins belonging to the same Port are associated with the same interrupt vector: there is one vector for Port A and one for Port B. In order to use one port pin as interrupt, it must be configured as an interrupt source (see I/O Ports chapter). In this manner, up to 16 Port Interrupt sources are available. By reading the Port the sources that belong to the same Port can be discriminated. The Port Interrupts can be configured to be active on the rising or the falling edge.

Warning: changing the NMI or Port Interrupt polarity an interrupt request is generated.
All the interrupt sources are filtered, in order to avoid false interrupt requests caused by glitches.
The Trap instruction is something between a interrupt and a call: it generated an interrupt request at top priority level and the control is passed to the associated interrupt routine which vector is located in the fixed addresses 30-32. This routine cannot be interrupted and it is serviced even if the interrupts are globally disabled.

### 5.4 Interrupt Maskability and Priority Levels

Interrupts can be masked by the corresponding INT_MASK Configuration Register 0 (00h). An interrupt is enabled when the mask bit is "1". Vice versa, when the bit is " 0 ", the interrupt is masked and the eventual requests are kept pending.
All the interrupts, with the exception of the NMI and TRAP that have fixed level priority, have a configurable priority level. The configuration of the priority levels is completed by writing three consecutive Configuration Registers: INT_PRL_H, INT_PRL_M, INT_PRL_L, addresses from 2 to 4 (02h-04h). The 24 bits of these registers are divided into 8 groups of three bits: each group is associated with a priority level. The three bits of each group are written with the code number associated with the interrupt source. See Table 5.1 to know the codes.

Remark: The priority levels Configuration Registers must be programmed with different values for each 3-bit groups to avoid erroneous operation. For this reason the Interrupt priority must be fixed at the beginning of the main program, because the reset values of the Configuration Registers correspond to an undefined configuration (all zeros). During program execution the interrupt priority can only be modified within the Main Program: it cannot be changed within an interrupt service routine.

### 5.5 Interrupt RESET

When an interrupt is masked, all requests are not acknowledged and remain pending. When the pending interrupt is enabled it is immediately serviced. This event may be undesired; in order to avoid this a RINT instruction may be inserted followed by the code number that identifies the interrupt to reset the pending request. See Table 5.1 to know the codes.

Note: Similarly to the CALL instruction, after a TRAP the flags are not stacked.

Figure 5.3 Example of Interrupt Requests


### 5.6 Register Description

## Interrupt Mask Register (INT_MASK)

Configuration Register 0 (00h) Read/Write
Reset Value: 00000000 (00h)

| MSKPB | MSKPA | MSKI2C | MSKSPI | MSKSCI | MSKT1 | MSKTO | MSKAD |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7: MSKPB Interrupt Mask Port B
0: Port B interrupt masked
1: Port B interrupt enabled
Bit 6: MSKPA Interrupt Mask Port A
0: Port A interrupt masked
1: Port A interrupt enabled
Bit 5: MSKI2C Interrupt Mask ${ }^{2}$ C Interface $0: I^{2} \mathrm{C}$ Interface interrupt masked 1: $I^{2} \mathrm{C}$ Interface interrupt enabled

Bit 4: MSKSPI Interrupt Mask SPI
0: SPI interrupt masked
1: SPI interrupt enabled

Bit 3: MSKSCI Interrupt Mask SCI
0: SCI interrupt masked
1: SCI interrupt enabled
Bit 2: MSKT1 Interrupt Mask PWM/Timer 1
0: Pwm/Timer 1 interrupt masked
1: Pwm/Timer 1 interrupt enabled

Bit 1: MSKTO Interrupt Mask Pwm/Timer 0
0: Pwm/Timer 0 interrupt masked
1: Pwm/Timer 0 interrupt enabled

Bit 0: MSKAD Interrupt Mask A/D Converter
0: A/D interrupt masked
1: A/D interrupt enabled

Interrupt Polarity Register (INT_POL)
Configuration Register 1 (01h) Read/Write
Reset Value: 00000000 (00h)

7
0


Bit 7-5: Not Used

Bit 4-3: See Timer 0 Registers Description

## Bit 2: POLPB Port B Interrupt Polarity

0 : The Port B interrupt is triggered on the rising edge of the applied external signal.
1: The Port B interrupt is triggered on the falling edge of the applied external signal.

Bit 1: POLPA Port A Interrupt Polarity
0 : The Port A interrupt is triggered on the rising edge of the applied external signal.
1: The Port A interrupt is triggered on the falling edge of the applied external signal.

Bit 0: POLNMI Non Maskable Interrupt Polarity
0 : The NMI is triggered on the rising edge of the applied external signal.
1: The NMI is triggered on the falling edge of the applied external signal.

High Priority Register (INT_PRL_H)
Configuration Register 2 (02h) Read/Write
Reset Value: 11111010 (OFAh)
7

| PRL23 | PRL22 | PRL21 | PRL20 | PRL19 | PRL18 | PRL17 | PRL16 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

## Medium Priority Register (INT_PRL_M)

Configuration Register 3 (03h) Read/Write
Reset Value: 11000110 (0C6h)
7

| PRL15 | PRL14 | PRL13 | PRL12 | PRL11 | PRL10 | PRL9 | PRL8 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

## Low Priority Register (INT_PRL_L)

Configuration Register 4 (04h) Read/Write
Reset Value: 10001000 (088h)
7
0

| PRL7 | PRL6 | PRL5 | PRL4 | PRL3 | PRL2 | PRL1 | PRL0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

These three register are used to configure the priority level of each interrupt source. The 24 bits of these registers (PRL24-PRL0) are divided into 8 groups of three bits: each group is associated with a priority level (from level 1, the highest, to level 8, the lowest: level 0 is fixed for the NMI that can be interrupted only by the TRAP). The three bits of each group are written with the code number associated with the interrupt source (see Table 5.1).

PRL2-PRL1: Interrupt priority level 1 (highest)
PRL5-PRL3: Interrupt priority level 2
PRL8-PRL6: Interrupt priority level 3
PRL11-PRL9:Interrupt priority level 4
PRL14-PRL12: Interrupt priority level 5
PRL17-PRL15: Interrupt priority level 6
PRL20-PRL18: Interrupt prioritylevel 7
PRL23-PRL21: Interrupt priority level 8 (lowest)
Example: writing the code 110 into PRL8-PRL6 bits the priority level 3 is assigned to the Port A Interrupt.
Warning: the Priority Level configuration registers must be always configured.

Table 5.1 Interrupt sources paramethers

| Interrupt Source | Priority type | PRL code | RINT code | Maskable | Vector Addresses |
| :--- | :---: | :---: | :---: | :---: | :---: |
| A/D Converter | Programmable | 000 | 0 | Yes | $3-5$ (03h-05h) |
| PWM/Timer 0 | Programmable | 001 | 1 | Yes | $6-8$ (06h-08h) |
| PWM/Timer 1 | Programmable | 010 | 2 | Yes | $9-11$ (09h-0Bh) |
| SCI | Programmable | 011 | 3 | Yes | $12-14$ (0Ch-0Eh) |
| SPI | Programmable | 100 | 4 | Yes | $15-17(0 \mathrm{Fh}-011 \mathrm{~h})$ |
| IC Interface $^{\text {Port A }}$ | Programmable | 101 | 5 | Yes | $18-20(012 \mathrm{~h}-014 \mathrm{~h})$ |
| Port B | Programmable | 110 | 6 | Yes | $21-23(015 \mathrm{~h}-017 \mathrm{~h})$ |
| NMI | Programmable | 111 | 7 | Yes | $24-26$ (018h-01Ah) |
| TRAP | Fixed | - | 8 | No | $27-29(01 \mathrm{Bh}-01 \mathrm{Dh})$ |

## 6 CLOCK, RESET \& POWER SAVING MODES

### 6.1 Clock

The ST52F510/F513/F514 Clock Generator module generates the internal clock for the internal Control Unit, ALU and on-chip peripherals. The Clock is designed to require a minimum of external components.
ST52F510/F513/F514 devices supply the internal oscillator in four clock modes:

- External oscillator
- External clock

■ External RC oscillator
■ Internal clock
The device always starts in internal clock mode, excluding any external clock source. After the start-up phase the clock is configured according to the user definition programmed in the Option Byte 0 (OSC_CR). The internal clock generator can supply an internal clock signal with a fixed frequency of $10 \mathrm{MHz} \pm 1 \%$, without the need for external components. In order to obtain the maximum accuracy, the frequency can be calibrated by configuring the related Option byte 2 (OSC_SET).
The external oscillator mode uses a quartz crystal or a ceramic resonator connected to OSCin and OSCout as illustrated in Figure 6.1. This figure also illustrates the connection of an external clock.
The ST52F510/F513/F514 oscillator circuit generates an internal clock signal with the same period and phase as the OSCIN input pin. The maximum frequency allowed is 24 MHz .
When the external oscillator is used, the loop gain can be adapted to the various frequencies values by configuring the three bits of the Option Byte 1 CLK_SET (see Register Decription, Table 6.2).
When an external clock is used, it must be connected to the pin OSCIN while OSCOUT can
be floating. In this case, Option Byte 1 bits must be written with 0 (000).
The crystal oscillator start-up time is a function of many variables: crystal parameters (especially $\mathrm{R}_{\mathrm{s}}$ ), oscillator load capacitance (CL), IC parameters, environment temperature and supply voltage.
The crystal or ceramic leads and circuit connections must be as short as possible. Typical values for CL1, CL2 are 10 pF for a 20 MHz crystal.
The clock signal can also be generated by an external RC circuit offering additional cost savings.
Figure 6.1 illustrates the possible connections. Frequency is a function of resistor, capacitance, supply voltage and operating temperature; some indicative values when $\mathrm{Vdd}=5 \mathrm{~V}$ and $\mathrm{T}=25^{\circ}$, are shown in Table 6.1.
The clock signal generates two internal clock signals: one for the CPU and one for the peripherals. The CPU clock frequency can be reduced, in order to decrease current consuption, by setting the CPU_CLK Configuration Register 46 (02Eh). The CPU clock can be reduced up to 64 times (see Register Description).

Table 6.1 RC Oscillator indicative frequencies

| $\mathbf{C}(\mathbf{p F})$ | $\mathbf{R}(\Omega)$ | $\mathbf{f}_{\text {osc }} \mathbf{( K H z )}$ | Variation |
| :---: | :---: | :---: | :---: |
| 20 pF | 9.5 K | 5000 | $6.6 \%$ |
|  | 10 K | 4870 | $7.1 \%$ |
|  | 20 K | 3000 | $5.3 \%$ |
|  | 50 K | 1360 | $3.3 \%$ |
|  | 100 K | 724 | $2.8 \%$ |
| 100 pF | 10 K | 1720 | $7.5 \%$ |
|  | 20 K | 926 | $8 \%$ |
|  | 50 K | 424 | $11.2 \%$ |
|  | 100 K | 248 | $15 \%$ |

Figure 6.1 Oscillator Connections
(CRYSTAL CLOCK

### 6.2 Reset

Four Reset sources are available:

- RESET pin (external source)
- WATCHDOG (internal source)
- POWER ON Reset (Internal source)
- PLVD Reset (Internal source)

When a Reset event occurs, the user program restarts from the beginning.
6.2.1 External Reset. Reset is an input pin. An internal reset does not affect this pin. A Reset signal originated by external sources is recognized immediately. The RESET pin may be used to ensure Vdd has risen to a point where the ICU can operate correctly before the user program is run. Reset must be set to Vdd in working mode.
A Pull up resistor of $100 \mathrm{~K} \Omega$ guarantees that the RESET pin is at level " 1 " when no HALT or PowerOn events occur. If an external resistor is connected to the RESET pin a minimum value of $10 \mathrm{~K} \Omega$ must be used.
6.2.2 Reset Procedures. After the Reset pin is set to Vdd or following a Power-On Reset event, the device is not started until the internal supply voltage has reached the nominal level of 2.5 V (corresponding roughly to $\mathrm{Vdd}=2.8 \mathrm{~V}$ ).

After this level has been reached, the internal oscillator ( 10 MHZ ) is started and a delay period of 4.096 clock cycles is initiated, in order to allow the oscillator to stabilize and to ensure that recovery has taken place from the Reset state.
If the device has been configured to work with the internal clock, the user program starts, otherwise the Option Byte 7 (WAKEUP) is read and another count starts before running the user program. The duration of the count depends on the contents of the Option Byte 7 (WAKEUP), that works as a prescaler, according to the follwing formula:

$$
\text { Delay }=4096 \times(W A K E U P+1) \times T c l k
$$

This delay has been introduced in order to ensure that the oscillator has become stable after its restart.
If the Reset is generated by the PLVD or the Watchdog, the oscillator is not turned off; for this reason the CPU is then restarted immediately, without the delay.
After a RESET procedure is completed, the core reads the instruction stored in the first 3 bytes of the Program/Data Memory, which contains a JUMP instruction to the first instruction of the user program. The Assembler tool automatically generates this Jump instruction with the first instruction address.

Figure 6.2 Reset Block Diagram


### 6.3 Programmable Low Voltage Detector

The on-chip Programmable Low Voltage Detector (PLVD) circuit prevents the processor from falling into an unpredictable status if the power supply drops below a certain level.
When Vdd drops below the detection level, the PLVD causes an internal processor Reset that remains active as long as Vdd remains below the trigger level.
The PLVD resets the entire device except the Power-on Detector and the PLVD itself.
The PLVD can be enabled/disabled at reset by setting the Option Byte 3 (PLVD_CR) bits.
When Vdd increases above the Trigger Level, the PLVD reset is deactivated and the user program is started from the beginning.
The detection levels are programmable by means of the Option Byte 3 (PLVD_CR). There are three levels for the PLVD falling voltages ( $2.9 \mathrm{~V}, 3.4 \mathrm{~V}$, 3.9 V ) and for rising voltages ( $3.1 \mathrm{~V}, 3.65 \mathrm{~V}, 4.2 \mathrm{~V}$ ). The hysteresis for each level are respectively 200 $\mathrm{mV}, 250 \mathrm{mV}$ and 300 mV .
The PLVD circuit will only detect a drop if Vdd voltage stays below the safe threshold for at least $5 \mu$ s before activation/deactivation of the PLVD in order to filter voltage spikes.

Remark: the PLVD function isn't active when it is in HALT mode. In that case the device is reset if the Vdd voltage stays below the threshold of 2 V .

### 6.4 Power Saving modes

There are two types of Power Saving modes: WAIT and HALT mode. These conditions may be entered by using the WAIT or HALT instructions.
6.4.1 Wait Mode. Wait mode places the ICU in a low power consumption status by stopping the CPU. All peripherals and the watchdog remain active. During WAIT mode the Interrupts are enabled. The ICU remains in Wait mode until an Interrupt or a RESET occurs, whereupon the Program Counter jumps to the interrupt service routine or, if a Reset occurs, to the beginning of the user program.
6.4.2 Halt Mode. Halt mode is the lowest ICU power consumption mode, which is entered by executing the HALT instruction. The internal oscillator is turned off, causing all internal processing to be terminated, including the operations of the on-chip peripherals. Halt mode cannot be used when the watchdog is enabled. If the HALT instruction is executed while the watchdog system is enabled, it will be skipped without modifying the normal CPU operations.

The ICU can exit Halt mode upon reception of an NMI, a Port Interrupt or a Reset. The internal oscillator ( 10 MHZ ) is started and a delay period of 4.096 clock cycles is initiated, in order to allow the oscillator to stabilize and to ensure that recovery has taken place from the Reset state.
If the device has been configured to work with the internal clock, the user program is started, otherwise the Option Byte 7 (WAKEUP) is read and another count is started before running the user program. The count duration depends on the contents of the Option Byte 7 (WAKEUP), that works as prescaler, according to the follwing formula:

$$
\text { Delay }=4096 \times(W A K E U P+1) \times T c l k
$$

This delay has been introduced in ordet to ensure that the oscillator has become stable after it is restarted.
After the start up delay, by exiting with the NMI or a Port interrupt, the CPU restarts operations by serving the associated interrupt routine.

Note: if the Port Interrupt is masked, the ICU doesn't exit the Halt mode with this interrupt.

Figure 6.3 WAIT Flow Chart


Figure 6.4 HALT Flow Chart


### 6.5 Register Description

The following section describes the Register which are used to configure the Clock, Reset and PLVD.

### 6.5.1 Configuration Register.

## CPU Clock Prescaler (CPU_CLK)

Configuration Register 46 (02Eh) Read/Write Reset Value: 00000000 (00h)

7
0

| - | - | CPUCK5 | CPUCK4 | CPUCK3 | CPUCK2 | CPUCK1 | CPUCKO |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-6: Not Used

Bit 5-0: CPUCK5-0 CPU Clock Prescaler bits
The CPU Clock frequency is divided by a factor described in the following table

| CPUCK5-0 | CPU Clock |
| :---: | :---: |
| 000000 | $\mathrm{f}_{\mathrm{CPU}}=\mathrm{f}_{\mathrm{OSC}}$ |
| 000001 | $\mathrm{f}_{\mathrm{CPU}}=\mathrm{f}_{\mathrm{OSC}} / 2$ |
| 000010 | $\mathrm{f}_{\mathrm{CPU}}=\mathrm{f}_{\mathrm{OSC}} / 4$ |
| 000100 | $\mathrm{f}_{\mathrm{CPU}}=\mathrm{f}_{\mathrm{OSC}} / 8$ |
| 001000 | $\mathrm{f}_{\mathrm{CPU}}=\mathrm{f}_{\mathrm{OSC}} / 16$ |
| 010000 | $\mathrm{f}_{\mathrm{CPU}}=\mathrm{f}_{\mathrm{OSC}} / 32$ |
| 100000 | $\mathrm{f}_{\mathrm{CPU}}=\mathrm{f}_{\mathrm{OSC}} / 64$ |
| others | $\mathrm{f}_{\mathrm{CPU}}=\mathrm{f}_{\mathrm{OSC}} / 64$ |

### 6.5.2 Option Bytes.

## Clock Mode (OSC_CR)

Option Byte 0 (00h)
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$



Bit 7-2: Not Used

Bit 1-0: CKMOD1-0 Clock Mode
00: Internal Oscillator
01: External Clock or quartz
1x: External RC oscillator

## External Clock Parameters (CLK_SET)

Option Byte 1 (01h)
Reset Value: 00000000 (00h)
$\qquad$

| - | - | - | - | - | CKPAR2 | CKPAR1 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| - | CKPAR0 |  |  |  |  |  |

Bit 7-3: Not Used

Bit 2-0: CKPAR2-0 Oscillator Gains
These three bits enable/disable the loop gains when a external clock or quartz are used for generating the clock. The following table decribes the possible configuration options. Table 6.2 illustrates the reccomended values for the most common frequencies used, time to start the oscillations and the settling time to have a duty cycle of $40 \%-60 \%$ (at steady state it is $50 \%$ ).

| CKPAR2-0 | Enabled Gain Stages |
| :---: | :---: |
| 000 | No Gains (External Clock Mode) |
| 001 | 1 gain stage enabled |
| 010 | not allowed |
| 011 | 3 gain stage enabled |
| 100 | not allowed |
| 101 | 6 gain stage enabled |
| 110 | not allowed |
| 111 | 8 gain stage enabled |

Warning: If an External Clock is used instead of a quartz or ceramic resonator, it is reccomended that no gain be enabled (CKPAR2-0=000) in order lo lower the current consumption.

Table 6.2 Recomended Gains for the most common frequencies

| Frequency | Recommend <br> Gain Stages | CKPAR2-0 | Oscillation <br> Start Times | Settling Times for <br> 40\% duty-cycle |
| :---: | :---: | :---: | :---: | :---: |
| External Clock | 0 | 000 | - | - |
| 1 MHz | 1 | 001 | $367 \mu \mathrm{~s}$ | $27 \mu \mathrm{~s}$ |
| 4 MHz | 1 | 001 | $84 \mu \mathrm{~s}$ | $10 \mu \mathrm{~s}$ |
| 8 MHz | 3 | 011 | $75 \mu \mathrm{~s}$ | $9 \mu \mathrm{~s}$ |
| 10 MHz | 3 | 011 | 79 s | $5 \mu \mathrm{~s}$ |
| 12 MHz | 6 | 101 | $352 \mu \mathrm{~s}$ | $8 \mu \mathrm{~s}$ |
| 16 MHz | 6 | 111 | $165 \mu \mathrm{~s}$ | $7 \mu \mathrm{~s}$ |
| 20 MHz | 8 |  | $11 \mu \mathrm{~s}$ |  |

(1) The recommended values have been chosen to have the best tradeoff beetwen start time and current consumption. Higher gains give shorter Start times; lower gains give less current consumption.
(2) Indicative values by design at $25^{\circ}$ Celsius, $\mathrm{V}_{\mathrm{DD}}=2.6 \mathrm{~V}$. Not Tested in production.

## Internal Oscillator Calibration (OSC_SET)

Option Byte 2 (02h)
Reset Value: 00010100 (14h)
7
0

| - | - | OSPAR5 | OSPAR4 | OSPAR3 | OSPAR2 | OSPAR1 | OSPAR0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-6: Not Used
Bit 5-0: OSPAR5-0 Internal Oscillator Parameters
These bits are used in order to calibrate the precision of the internal oscillator working at 10 MHz . The six bits enable some current generators with steps of $0.05 \mu \mathrm{~A}$ corresponding to interval of frequency of 100 KHz .

Warning: the maximum configuration value allowed is 101000 (40). The value coresponding to the 10 MHz by design is 010100 (20).

## PLVD Control Register (PLVD_CR)

Option Byte 3 (03h)
Reset Value: 00000010 (02h)

## 7

 0

Bit 7-2: Not Used
Bit 1-0: PLVD1-0 PLVD detection levels
00: PLVD disabled
01: Medium detection level
10: Lowest detection level
11: Highest detection level

## Wake-Up Time Prescaler (WAKEUP)

Option Byte 7 (07h)
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| WK7 | WK6 | WK5 | WK4 | WK3 | WK2 | WK1 | WK0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: WK7-0 Wake-up prescaler
This byte determinates the time delay for the stabilization of the oscillator after an External Reset or a POR and after the wake-up from Halt. The time delay is computed according to the following formula:

$$
\text { Delay }=4096 \times(W A K E U P+1) \times T c l k
$$

Warning: If the internal clock is used as clock source the prescaler is not used.

## 7 I/O PORTS

### 7.1 Introduction

ST52F510/F513/F514 are characterized by flexible individually programmable multi-functional I/O lines. The ST52F510/F513/F514 supplies devices with up to 3 Ports (named from A to C) with up to 22 I/O lines.
Each pin can be used as a digital I/O or can be connected with a peripheral (Alternate Function). The I/O lines belonging to Port A and Port B can also be used to generate Port Interrupts.
The I/O Port pins can be configured in the following modes:
■ Input high impedance (reset state)

- Input with pull-up
- Output with pull-up
- Output push-pull
- Output with weak pull-up
- Output open drain
- Interrupt with pull-up

■ Interrupt without pull-up
These eight modes can be selected by programming three Configuration Registers for each Port. All the pins that belong to the same Port can be configured separately by setting the corresponding bits in the three registers (see Register Description).
To avoid side effects, the Configuration Registers register are latched only when the Direction Register (PORT_x_DDR) is written. For this reason this register must be always written when modifying the pin configuration.
All the I/O digital pins are TTL compatible and have a Schmitt Trigger. The output buffer can supply high current sink (up to 8 mA ).

Figure 7.1 Digital Pin


### 7.2 Input Mode

The pins configured as input can be read by accessing the corresponding Port Input Register by means of the LDRI instruction. The addresses for Port A, B and C are respectively 0 (00h), 1 (01h), and 2 ( 02 h ).
When executing the LDRI instruction all the signals connected to the input pins of the Port are read and the logical value is copied in the specified Register File location. If some pins are configured in output, the port buffer contents, which are the last written logical values in the output pins, are read.

### 7.3 Output Mode

The pins configured as output can be written by accessing the corresponding Port Output Register by means of the LDPR, LDPI and LDPE instructions. The addresses for Port A, B and C are respectively, 0 ( 00 h ), 1 ( 01 h ), and 2 ( 02 h ).
When executing the above mentioned instructions, the Port buffer is written and the Port pin signals are modified. If some pins are configured as input or as interrupt, the values are ignored.

### 7.4 Interrupt Mode

The pins configured as Interrupt Mode can generate a Port Interrupt request. Only Port A and Port $B$ pins can be configured in this mode.
An Interrupt vector is associated to each Port: there are two Port Interrupts available but more pins of the ports can act as source at the same time.
The Configuration Registers switch the signals deriving from interrupt pins to an OR gate that generates the interrupt request signal. The signal deriving from the pins can be read, allowing the discrimination of the interrupt sources when more than one pin can generate the interrupt signal.
The interrupt trigger can be configured either in the rising or falling edge of the external signal.

Figure 7.2 Analog Pin


### 7.5 Alternate Functions

The Alternate Function allows the pins to be connected with the peripheral signals or NMI. Not all Port pins have an Alternate Function associated.
A Configuration Register (PORT_x_AF) for each Port is used to switch from the Digital I/O function or the Alternate Function.
Some pins can have two Alternate Functions: one input function and one output function. To switch between the two functions, the PORT_x_AF must be configured in Alternate Function mode and the PORT_x_DDR Configuration Register must be switched in Input mode or in Output mode.
NMI is considered an Alternate Function. For this reason an NMI interrupt request can't be generated unless the PA7 pin is configured in Alternate Function and in one of the Input modes.

Figure 7.3 Port Pin Architecture

When an on-chip peripheral is configured to use a pin, the correct I/O mode of the related pin should be selected by selecting one of the appropriate modes. See the Registers description in order to obtain the right configurations.Some peripherals, as for example the $\mathrm{I}^{2} \mathrm{C}$ peripheral, directly drive the pin configuration according to the current function, overriding the user configuration.

### 7.6 Register Description

In order to configure the Port's pins, the three Configuration Registers PORT_x_PULLUP, PORT_x_OR and PORT_x_DDR must be configured. The combination of these three registers determine the pin's configuration, according to the scheme shown in Table 7.1.
In order to select between the digital functions or Alternate functions PORT_x_AF register must be configured. Each bit of the configuration registers configures the pin of the corresponding position (example: PORT_A_DDR bit 5 configures the pin PA5).


### 7.6.1 Configuration Registers.

Port A Pull-Up Register (PORT_A_PULLUP)
Configuration Register 24 (018h) Read/Write
Reset Value: 00000000 (00h)

| 7 |
| :--- |
| 7 |
| PUA7 |
| PUA6 |

Bit 7-0: PUA7-0 Port A pull-up (see Table 7.1)
0: Port A pin without pull-up
1: Port A pin with pull-up

## Port A Option Register (PORT_A_OR)

Configuration Register 25 (019h) Read/Write
Reset Value: 00000000 (00h)

| 7 |  |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :---: |
| ORA7 | ORA6 | ORA5 | ORA4 | ORA3 | ORA2 | ORA1 | ORA0 |

Bit 7-0: ORA7-0 Port A option (see Table 7.1)

Port A Data Direction Register (PORT_A_DDR)
Configuration Register 26 (01Ah) Read/Write
Reset Value: 00000000 (00h)
7
0

| DDRA7 | DDRA6 | DDRA5 | DDRA4 | DDRA3 | DDRA2 | DDRA1 | DDRA0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: DDRA7-0 Port A direction (see Table 7.1)
0 : Port A pin configured as input
1: Port A pin configured as output

Port A Alternate Fuction (PORT_A_AF)
Configuration Register 27 (01Bh) Read/Write
Reset Value: 00000000 (00h)
7
0

| AFA7 | AFA6 | AFA5 | AFA4 | AFA3 | AFA2 | AFA1 | AFA0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit 7: AFA7 Alternate Function PA7
0 : Digital I/O
1: INT
Bit 6: AFA6 Alternate Function PA6
0 : Digital I/O
1: T0OUT

Bit 5: AFA5 Alternate Function PA5
0: Digital I/O
1: TCLK

Bit 4: AFA4 Alternate Function PA4
0 : Digital I/O
1: TSTRT

Bit 3: AFA3 Alternate Function PA3
0 : Digital I/O
1: TRES
Bit 2: AFA2 Alternate Function PA2
0 : Digital I/O
1: T1OUT
Bit 1: AFA1 Alternate Function PA1
0 : Digital I/O
1: SDA

Bit 0: AFAO Alternate Function PA0
0 : Digital I/O
1: SCL
Table 7.1 Pin mode configuration

| MODE | PU | OR | DDR |
| :--- | :---: | :---: | :---: |
| Input high impedance | 0 | 0 | 0 |
| Input with pull-up | 1 | 0 | 0 |
| Interrupt without pull-up | 0 | 1 | 0 |
| Interrupt with pull-up | 1 | 1 | 0 |
| Output push-pull | 0 | 0 | 1 |
| Output with pull-up | 1 | 0 | 1 |
| Output open drain | 0 | 1 | 1 |
| Output weak pull-up | 1 | 1 | 1 |

Port B Pull-Up Register (PORT_B_PULLUP)
Configuration Register 28 (01Ch) Read/Write
Reset Value: 00000000 (00h)
7 0

| PUB7* $^{*}$ | PUB6* $^{*}$ | PUB5** $^{*}$ | PUB4** $^{*}$ | PUB3** $^{* *}$ | PUB2** | PUB1 | PUB0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

(*) Not used in 16/20 pin package devices
(**) Not used in 16 pin package devices
Bit 7-0: PUB7-0 Port B pull-up (see Table 7.1)
0 : Port B pin without pull-up
1: Port B pin with pull-up
Port B Option Register (PORT_B_OR)
Configuration Register 29 (01Dh) Read/Write
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| ORB7* $^{*}$ | ORB6* | ORB5** | ORB4** | ORB3** | ORB2** | ORB1 | ORB0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

(*) Not used in 16/20 pin package devices
(**) Not used in 16 pin package devices
Bit 7-0: ORB7-0 Port B option (see Table 7.1)

Port B Data Direction Register (PORT_B_DDR)
Configuration Register 30 (01Eh) Read/Write
Reset Value: 00000000 (00h)

(*) Not used in 16/20 pin package devices
(**) Not used in 16 pin package devices
Bit 7-0: DDRB7-0 Port B direction (see Table 7.1)
0 : Port B pin configured as input
1: Port B pin configured as output

## Port B Alternate Fuction (PORT_B_AF)

Configuration Register 31 (01Fh) Read/Write
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| AFB7 | AFB6 | AFB5 | AFB4 | AFB3 | AFB2 | AFB1 | AFB0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Note: This register is not used in 16 pin devices

Bit 7: AFB7 Alternate Function PB7
0 : Digital I/O
1: AIN7

Bit 6: AFB6 Alternate Function PB6
0 : Digital I/O
1: AIN6

Bit 5: AFB5 Alternate Function PB5
0 : Digital I/O
1: AIN5
Bit 4: AFB4 Alternate Function PB4
0 : Digital I/O
1: AIN4
Bit 3: AFB3 Alternate Function PB3
0 : Digital I/O
1: AIN3
Bit 2: AFB2 Alternate Function PB2
0: Digital I/O
1: AIN2

Bit 1: AFB1 Alternate Function PB1
0: Digital I/O
1: AIN1
Bit 0: AFBO Alternate Function PB0
0: Digital I/O
1: AINO / VREF

Port C Pull-Up Register (PORT_C_PULLUP) Configuration Register 32 (020h) Read/Write
Reset Value: 00000000 (00h)
7
0

| - | - | PUC5 | PUC4 | PUC3 | PUC2 | PUC1 | PUC0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Note: This register is not used in 16/20 pin devices
Bit 7-6: Not Used
Bit 5-0: PUC5-0 Port C pull-up (see Table 7.1)
0 : Port C pin without pull-up

1: Port C pin with pull-up
Port C Option Register (PORT_C_OR)
Configuration Register 33 (021h) Read/Write Reset Value: 00000000 (00h)

7

| - | - | ORC5 | ORC4 | ORC3 | ORC2 | ORC1 | ORC0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Note: This register is not used in 16/20 pin devices
Bit 7-6: Not Used
Bit 5-0: ORC5-0 Port C option (see Table 7.1)

Port C Data Direction Register (PORT_C_DDR)
Configuration Register 34 (022h) Read/Write
Reset Value: 00000000 (00h)
7

| - | - | DDRC5 | DDRC4 | DDRC3 | DDRC2 | DDRC1 | DDRC0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Note: This register is not used in 16/20 pin devices

## Bit 7-6: Not Used

Bit 5-0: DDRC5-0 Port C direction (see Table 7.1)
0 : Port C pin configured as input
1: Port C pin configured as output
Note: in order to achieve low current consuption, the port pins must be configured as input pull-up, even though they are not existing in the package. For example in 20 pin devices, the pins PB6-7 and PC0-7 must be configured in input pull-up.

## Port C Alternate Fuction (PORT_C_AF)

Configuration Register 35 (023h) Read/Write
Reset Value: 00000000 (00h)

7
0

| - | - | AFC5 | AFC4 | AFC3 | AFC2 | AFC1 | AFC0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit 7-6: Not Used

Bit 5: AFC5 Alternate Function PC5
0: Digital I/O

## 1: TRES

Bit 4: AFC4 Alternate Function PC4
0 : Digital I/O
1: TX
Bit 3: AFC3 Alternate Function PC3
0: Digital I/O
1: SS

Bit 2: AFC2 Alternate Function PC2
0 : Digital I/O
1: MISO

Bit 1: AFC1 Alternate Function PC1
0: Digital I/O
1: MOSI

Bit 0: AFCO Alternate Function PC0
0: Digital I/O
1: SCK

### 7.6.2 Input Registers.

Port A Data Input Register (PORT_A_IN)
Input Register 0 (00h) Read only
Reset Value: XXXX XXXX

| 7 |  |  |  | 0 |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| PAl7 | PAl6 | PAl5 | PAl4 | PAl3 | PAl2 | PAl1 | PAl0 |

Bit 7-0: PAI7-0 Port A Input data
The logical level applied in the Port A pins, configured as digital input, can be achieved by reading this register.

Port B Data Input Register (PORT_B_IN) Input Register 1 (01h) Read only Reset Value: XXXX XXXX

7 0

| PBI7* | PB16* | PBI5** | PB14** | PBI3** | PBI2** | PBI1 | PBI0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

(*) Not used in 16/20 pin package devices
(**) Not used in 16 pin package devices

Bit 7-0: PBI7-0 Port B Input data
The logical level applied in the Port B pins, configured as digital input, can be achieved by reading this register.

## Port C Data Input Register (PORT_C_IN)

Input Register 2 (02h) Read only
Reset Value: XXXX XXXX


Note: This register is not used in 16/20 pin devices

## Bit 7-6: Not Used

Bit 5-0: PCI5-0 Port C Input data

The logical level applied in the Port C pins, configured as digital input, can be achieved by reading this register.

### 7.6.3 Output Registers.

Port A Data Output Register (PORT_A_OUT)
Output Register 0 (00h) Write only
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| PAO7 | PAO6 | PAO5 | PAO4 | PAO3 | PAO2 | PAO1 | PAO0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

The logical values written in these register bits are put in the Port A pins configured as digital output.

Port B Data Output Register (PORT_B_OUT)
Output Register 1 (01h) Write only
Reset Value: 00000000 (00h)
7
0

(*) Not used in 16/20 pin package devices
(**) Not used in 16 pin package devices
Bit 7-0: PBO7-0 Port B Input data
The logical values written in these register bits are put in the Port B pins configured as digital output.

## Port C Data Output Register (PORT_C_OUT)

Output Register 2 (02h) Write only Reset Value: 00000000 (00h)
7

7 |  |  |  |  | 0 |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| - | - | PCO5 | PCO4 | PCO3 | PCO2 | PCO1 | PCO0 |

Note: This register is not used in 16/20 pin devices
Bit 7-6: Not Used

Bit 5-0: PCO5-0 Port C Input data
The logical values written in these register bits are put in the Port C pins configured as digital output.

Bit 7-0: PA07-0 Port A Output data

## 8 FUZZY COMPUTATION (DP)

The ST52F510/F513/F514 Decision Processor (DP) main features are:
■ Up to 8 Inputs with 8-bit resolution;

- 1 Kbyte of Program/Data Memory available to store more than 300 to Membership Functions (Mbfs) for each Input;
- Up to 128 Outputs with 8-bit resolution;
- Possibility of processing fuzzy rules with an UNLIMITED number of antecedents;

■ UNLIMITED number of Rules and Fuzzy Blocks.
The limits on the number of Fuzzy Rules and Fuzzy program blocks are only related to the Program/Data Memory size.

### 8.1 Fuzzy Inference

The block diagram shown in Figure 8.1 describes the different steps performed during a Fuzzy algorithm. The ST52F510/F513/F514 Core allows for the implementation of a Mamdami type fuzzy inference with crisp consequents. Inputs for fuzzy inference are stored in 8 dedicated Fuzzy input registers. The LDFR instruction is used to set the Input Fuzzy registers with values stored in the Register File. The result of a Fuzzy inference is stored directly in a location of the Register File.

### 8.2 Fuzzyfication Phase

In this phase the intersection (alpha weight) between the input values and the related Mbfs (Figure 8.2) is performed.
Eight Fuzzy Input registers are available for Fuzzy inferences.

Figure 8.2 Alpha Weight Calculation


After loading the input values by using the LDFR assembler instruction, the user can start the fuzzy inference by using the FUZZY assembler instruction. During fuzzyfication: input data is transformed in the activation level (alpha weight) of the Mbf's.

### 8.3 Inference Phase

The Inference Phase manages the alpha weights obtained during the fuzzyfication phase to compute the truth value $(\omega)$ for each rule.
This is a calculation of the maximum (for the OR operator) and/or minimum (for the AND operator) performed on alpha values according to the logical connectives of Fuzzy Rules.
Several conditions may be linked together by linguistic connectives AND/OR, NOT operators and brackets.
The truth value $\omega$ and the related output singleton are used by the Defuzzyfication phase, in order to complete the inference calculation.

Figure 8.1 Fuzzy Inference


Figure 8.3 Fuzzyfication


### 8.4 Defuzzyfication

In this phase the output crisp values are determined by implementing the consequent part of the rules.
Each consequent Singleton $X_{i}$ is multiplied by its weight values $\omega_{\mathrm{i}}$, calculated by the Decision processor, in order to compute the upper part of the Defuzzyfication formula.
Each output value is obtained from the consequent crisp values ( $\mathrm{X}_{\mathrm{i}}$ ) by carrying out the following Defuzzyfication formula:

$$
Y_{i}=\frac{\sum_{j}^{N} X_{i j} \omega_{i j}}{\sum_{j}^{N} \omega_{i j}}
$$

where:
$\mathrm{i}=$ identifies the current output variable
$\mathrm{N}=$ number of the active rules on the current output
$\omega_{\mathrm{ij}}=$ weight of the j-th singleton
$X_{i j}=$ abscissa of the $j$-th singleton
The Decision Processor outputs are stored in the RAM location i-th specified in the assembler instruction OUT i.

### 8.5 Input Membership Function

The Decision Processor allows the management of triangular Mbfs. In order to define an Mbf, three different parameters must be stored on the Program/Data Memory (see Figure 8.4):

- the vertex of the Mbf: V;
- the length of the left semi-base: LVD;
- the length of the right semi-base: RVD;

In order to reduce the size of the memory area and the computational effort the vertical range of the vertex is fixed between 0 and 15 (4 bits)
By using the previous memorization method different kinds of triangular Membership Functions may be stored. Figure 8.5 shows some examples of valid Mbfs that can be defined in ST52F510/ F513/F514.
Each Mbf is then defined storing 3 bytes in the first Kbyte of the Program/Data Memory.
The Mbf is stored by using the following instruction:

## MBF n_mbf Ivd v rvd

where:
$n \_m b f$ is a tag number that identifies the Mbf $l v d, v$, and $r v d$ are the parameters that describe the Mbf's shape as described above.

Figure 8.4 Mbfs Parameters


Figure 8.5 Example of valid Mbfs


### 8.6 Output Singleton

The Decision Processor uses a particular kind of membership function called Singleton for its output variables. A Singleton doesn't have a shape, like a traditional Mbf, and is characterized by a single point identified by the couple ( $\mathrm{X}, \mathrm{w}$ ), where w is calculated by the Inference Unit as described earlier. Often, a Singleton is simply identified with its Crisp Value X.

Figure 8.6 Output Membership Functions


### 8.7 Fuzzy Rules

Rules can have the following structures:
if $A$ op $B$ op C...........then $Z$
if (A op B) op (C op D op E...) .then Z
where op is one of the possible linguistic operators (AND/OR)
In the first case the rule operators are managed sequentially; in the second one, the priority of the operator is fixed by the brackets.
Each rule is codified by using an instruction set, the inference time for a rule with 4 antecedents and 1 consequent is about 3 microseconds at 20 MHz .
The Assembler Instruction Set used to manage the Fuzzy operations is reported in the table below.

Table 8.1 Fuzzy Instructions Set

| Instruction | Description |
| :--- | :--- |
| MBF $n \_m b f / v d v$ rvd | Stores the Mbf $n \_m b f$ with the shape identified by the parameters /vd, $v$ and $r v d$ |
| IS $n m$ | Fixes the alpha value of the input n with the Mbf m |
| ISNOT $n m$ | Calculates the complementary alpha value of the input n with the Mbf m. |
| FZAND | Implements the Fuzzy operation AND |
| FZOR | Implements the Fuzzy operation OR |
| CON crisp | Multiplies the crisp value with the last $\omega$ weight <br> $n \_o u t ~$ |
| OUT $n \_o u t$ | Starts the computation of a sigle fuzzy variable |
| FUZZY | Modify the priority in the rule evaluation |
| ( ) |  |

## Example 1:

IF Input ${ }_{1}$ IS NOT Mbf AND Input $_{4}$ is $\mathrm{Mbf}_{12} \mathrm{OR} \mathrm{Input}_{3}$ IS Mbf $_{8}$ THEN Crisp ${ }_{1}$
is codified by the following instructions:
ISNOT 11 calculates the NOT $\alpha$ value of Input ${ }_{1}$ with $\mathrm{Mbf}_{1}$ and stores the result in internal registers
FZAND implements the operation AND between the previous and the next alpha value evaluated
IS 412 fixes the $\alpha$ value of Input $_{4}$ with $\mathrm{Mbf}_{12}$ and stores the result in internal registers
FZOR implements the operation OR between the previous and the next alpha value evaluated
IS 38 fixes the $\alpha$ value of Input $_{3}$ with $\mathrm{Mbf}_{8}$ and stores the result in internal registers
CON crisp $_{1}$ multiplies the result of the last $\Omega$ operation with the crisp value $\operatorname{crisp}_{1}$

Example 2, the priority of the operator is fixed by the brackets:

( parenthesis open to change the priority
IS 31 fixes the $\alpha$ value of $\operatorname{Input}_{3}$ with $\mathrm{Mbf}_{1}$ and stores the result in internal registers
FZAND implements the operation AND between the previous and the next alpha value evaluated
ISNOT 415 calculates the NOT $\alpha$ value of Input ${ }_{4}$ with $\mathrm{Mbf}_{15}$ and stores the result in internal registers
) parenthesis closed
FZOR implements the operation OR between the previous and the next alpha value evaluated ( parenthesis open to change the priority

IS 16 fixes the $\alpha$ value of $\operatorname{Input}_{1}$ with $\mathrm{Mbf}_{6}$ and stores the result in internal registers
FZOR implements the operation OR between the previous and the next alpha value evaluated
ISNOT 214 calculates the NOT $\alpha$ value of Input ${ }_{6}$ with $\mathrm{Mbf}_{14}$ and stores the result in internal registers
) parenthesis closed
CON crisp $_{2}$ multiplies the result of the last $\Omega$ operation with the crisp value crisp $_{2}$

At the end of the fuzzy rules related to the current Fuzzy Variable, by using the instruction OUT reg, the specified register is written with the computed value. Afterwards, the control of the algorithm returns to the CU . The next Fuzzy Variable evaluation must start again with a FUZZY instruction.

## 9 INSTRUCTION SET

ST52F510/F513/F514 supplies 107 (98 + 9 Fuzzy) instructions that perform computations and control the device. Computational time required for each instruction consists of one clock pulse for each Cycle plus 2 clock pulses for the decoding phase. Total computation time for each instruction is reported in Table 9.1
The ALU of ST52F510/F513/F514 can perform multiplication (MULT) and division (DIV). Multiplication is performed by using 8 bit operands storing the result in 2 registers ( 16 bit values), see Figure 2.3.
Division is performed between a 16 bit dividend and an 8 bit divider, the result and the remainder are stored in two 8-bit registers (see Figure 2.4).

### 9.1 Addressing Modes

ST52F510/F513/F514 instructions allow the following addressing modes:

- Inherent: this instruction type does not require an operand because the opcode specifies all the information necessary to carry out the instruction. Examples: NOP, SCF.
- Immediate: these instructions have an operand as a source immediate value. Examples: LDRC, ADDI.
- Direct: the operands of these instructions are specified with the direct addresses. The
operands can refer (according to the opcode) to addresses belonging to the different addressing spaces. Example: SUB, LDRE.
- Indirect: data addresses that are required are found in the locations specified as operands. Both source and/or destination operands can be addressed indirectly. The operands can refer, (according to the opcode) to addresses belonging to different addressing spaces. Examples: LDRR(reg1),(reg2);

LDER mem_addr,(reg1).

- Bit Direct: operands of these instructions directly address the bits of the specified Register File locations. Examples: BSET, BTEST.


### 9.2 Instruction Types

ST52F510/F513/F514 supplies the following instruction types:

- Load Instructions
- Arithmetic and Logic Instructions
- Bitwise instructions
- Jump Instructions
- Interrupt Management Instructions

■ Control Instructions

The instructions are listed in Table 9.1

Table 9.1 Instruction Set

| Load Instructions |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mnemonic | Instruction | Bytes | Cycles | Z | S | C |
| BLKSET | BLKSET const | 2 | $\left(^{*}\right)$ | - | - | - |
| GETPG | GETPG regx | 2 | 7 | - | - | - |
| LDCE | LDCE confx,memy | 3 | $8 / 9$ | - | - | - |
| LDCI | LDCI confx, const | 3 | 7 | - | - | - |
| LDCNF | LDCNF regx, conf | 3 | 7 | - | - | - |
| LDCR | LDCR confx, regy | 3 | 8 | - | - | - |
| LDER | LDER memx, regy | 3 | 10 | - | - | - |
| LDER | LDER (regx),(regy) | 3 | 11 | - | - | - |
| LDER | LDER (regx), regy | 3 | 10 | - | - | - |
| LDER | LDER memx,(regy) | 3 | 11 | - | - | - |
| LDFR | LDFR fuzzyx, regy | 3 | 8 | - | - | - |


| Load Instructions (continued) |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| LDPE | LDPE outx, memy | 3 | $8 / 9$ | - | - | - |
| LDPE | LDPE outx, (regy) | 3 | $9 / 10$ | - | - | - |
| LDPI | LDPI outx, const | 3 | 7 | - | - | - |
| LDPR | LDPR outx, regy | 3 | 8 | - | - | - |
| LDRC | LDRC regx, const | 3 | 7 | - | - | - |
| LDRE | LDRE regx, memy | 3 | $8 / 9$ | - | - | - |
| LDRE | LDRE (regx), (regy) | 3 | $10 / 11$ | - | - | - |
| LDRE | LDRE (regx), memy | 3 | $9 / 10$ | - | - | - |
| LDRE | LDRE regx, (regy) | 3 | $9 / 10$ | - | - | - |
| LDRI | LDRI regx, inpx | 3 | 7 | - | - | - |
| LDRR | LDRR regx, regy | 3 | 9 | - | - | - |
| LDRR | LDRR (regx), (regy) | 3 | 10 | - | - | - |
| LDRR | LDRR (regx), regy | 3 | 9 | - | - | - |
| LDRR | LDRR regx, (regy) | 3 | 10 | - | - | - |
| PGSET | PGSET const | 2 | 4 | - | - | - |
| PGSETR | PGSETR regx | 2 | 5 | - | - | - |
| POP | POP regx | 2 | 7 | - | - | - |
| PUSH | PUSH regx | 2 | 8 | - | - | - |


| Arithmetic Instructions |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mnemonic | Instruction | Bytes | Cycles | Z | S | C |
| ADD | ADD regx, regy | 3 | 9 | । | - | । |
| ADDC | ADDC regx, regy | 3 | 9 | । | - | । |
| ADDI | ADDI regx, const | 3 | 8 | । | - | । |
| ADDIC | ADDIC regx, const | 3 | 8 | । | - | । |
| ADDO | ADDO regx, regy | 3 | 11 | । | । | । |
| ADDOC | ADDOC regx, regy | 3 | 11 | । | । | । |
| ADDOI | ADDOI regx, const | 3 | 10 | । | । | । |
| ADDOIC | ADDOICregx,const | 3 | 10 | । | । | । |
| AND | AND regx, regy | 3 | 9 | । | - | - |
| ANDI | ANDI regx,const | 3 | 8 | । | - | - |
| CP | CP regx, regy | 3 | 8 | । | । | - |
| CPI | CPI regx,const | 3 | 7 | । | । | - |
| DEC | DEC regx | 2 | 7 | । | । | - |


| Arithmetic Instructions (continued) |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| DIV | DIV regx, regy | 3 | 16 | I | 1 | 1 |
| INC | INC regx | 2 | 7 | 1 | - | 1 |
| MIRROR | MIRROR regx | 2 | 7 | 1 | - | - |
| MULT | MULT regx, regy | 3 | 11 | 1 | - | - |
| NOT | NOT regx | 2 | 7 | 1 | - | - |
| OR | OR regx, regy | 3 | 9 | 1 | - | - |
| ORI | ORI regx, const | 3 | 8 | 1 | - | - |
| SUB | SUB regx, regy | 3 | 9 | 1 | 1 | - |
| SUBI | SUBI regx, const | 3 | 8 | 1 | 1 | - |
| SUBIS | SUBIS regx, const | 3 | 8 | I | I | - |
| SUBO | SUBO regx, regy | 3 | 11 | 1 | 1 | 1 |
| SUBOI | SUBOI regx, const | 3 | 10 | I | 1 | I |
| SUBOIS | SUBOISregx,const | 3 | 10 | I | 1 | I |
| SUBOS | SUBOS regx, regy | 3 | 11 | I | 1 | I |
| SUBS | SUBS regx, regy | 3 | 9 | I | I | - |
| RCF | RCF | 1 | 4 | - | - | I |
| RSF | RSF | 1 | 4 | - | I | - |
| RZF | RZF | 1 | 4 | I | - | - |
| SCF | SCF | 1 | 4 | - | - | I |
| SSF | SSF | 1 | 4 | - | 1 | - |
| SZF | SZF | 1 | 4 | 1 | - | - |
| XOR | XOR regx, regy | 3 | 9 | I | - | - |
| XORI | XORI regx, cons | 3 | 8 | I | - | - |


| Bitwise Instructions |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mnemonic | Instruction | Bytes | Cycles | Z | S | C |
| ASL | ASL regx | 2 | 7 | I | - | 1 |
| ASR | ASR regx | 2 | 7 | I | 1 | - |
| BNOT | BNOT regx, bit | 3 | 8 | I | - | - |
| BRES | BRES regx, bit | 3 | 8 | 1 | - | - |
| BSET | BSET regx, bit | 3 | 8 | I | - | - |
| BTEST | BTEST regx, bit | 3 | 7 | I | - | - |
| MTEST | MTEST regx,const | 3 | 7 | I | - | - |
| RLC | RLC regx | 2 | 7 | I | - | I |


| Bitwise Instructions (continued) |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| ROL | ROL regx | 2 | 7 | । | - | 1 |
| ROR | ROR regx | 2 | 7 | । | । | - |
| RRS | RRS regx | 2 | 7 | । | । | - |


| Jump Instructions |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mnemonic | Instruction | Bytes | Cycles | Z | S | C |  |
| CALL | CALL addr | 3 | 11 | - | - | - |  |
| JP | JP addr | 3 | 6 | - | - | - |  |
| JPC | JPC addr | 3 | $5 / 6$ | - | - | - |  |
| JPNC | JPNC addr | 3 | $5 / 6$ | - | - | - |  |
| JPNS | JPNS addr | 3 | $5 / 6$ | - | - | - |  |
| JPNZ | JPNZ addr | 3 | $5 / 6$ | - | - | - |  |
| JPS | JPS addr | 3 | $5 / 6$ | - | - | - |  |
| JPZ | JPZ addr | 3 | $5 / 6$ | - | - | - |  |
| RET | RET | 1 | 8 | - | - | - |  |


| Interrupt Management Instructions |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mnemonic | Instruction | Bytes | Cycles | Z | S | C |
| HALT | HALT | 1 | $4 / 13$ | - | - | - |
| MEGI | MEGI | 1 | $6 / 11$ | - | - | - |
| MDGI | MDGI | 1 | 5 | - | - | - |
| RETI | RETI | 1 | 9 | - | - | - |
| RINT | RINT INT | 2 | 6 | - | - | - |
| UDGI | UDGI | 1 | 5 | - | - | - |
| UEGI | UEGI | 1 | $6 / 11$ | - | - | - |
| TRAP | TRAP | 1 | 9 | - | - | - |
| WAITI | WAITI | 1 | $7 / 10$ | - | - | - |


| Control Instructions |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Mnemonic | Instruction | Bytes | Cycles | Z | S | C |  |
| FUZZY | FUZZY | 1 | 4 | - | - | - |  |
| NOP | NOP | 1 | 5 | - | - | - |  |
| WDTRFR | WDTRFR | 1 | 6 | - | - | - |  |
| WDTSLP | WDTSLP | 1 | 5 | - | - | - |  |

## Notes:

| regx, regy: | Register File Address |
| :--- | :--- |
| memx, memy: | Program/Data Memory Addresses |
| confx, confy: | Configuration Registers Addresses |
| outx: | Output Registers Addresses |
| inpx: | Input Registers Addresses |
| const: | Constant value |
| fuzzyx: | Fuzzy Input Registers |
| I | flag affected |
| - | flag not affected |

(*) The instruction BLKSET determines the start of a 32 byte block writing in Flash or EEPROM Program/ $_{\text {/ }}$ Data Memory. During this phase (about 4 ms ), the CPU is stopped to executing program instructions. The duration of the BLKSET instruction can be identified with this time.

## 10 10-BIT A/D CONVERTER

### 10.1 Introduction

ST52F510/F513/F514 A/D Converter is a 10-bit analog to digital converter with up to 8 analog inputs. The A/D converter offers a typical conversion time of $10 \mu \mathrm{~s}$ in fast mode and of $20 \mu \mathrm{~s}$ in slow mode. This period also includes the time of the integral Sample and Hold circuitry, which minimizes the need for external components and allows quick sampling of the signal for the minimum warping effect and integral conversion error.
In addition the peripheral performs a calibration procedure in order to get the maximum precision allowed in the data of conversion. The calibration procedure is performed in two phases: the precharging phase and the tuning phase. The precharging process can be executed, after the peripheral start, to set-up the internal references and to speed-up the tuning process. The tuning process is carried-out during the channels conversion.

Note: The user must be take in account both the pre-charging time and some dummy conversion (at least 20) for the tuning before starting the data acquisition. It is recommended to repeat this procedure at the start-up and after a long time peripheral stop.

The pre-charging process starts by starting the peripheral by setting to 1 the STR bit of the AD_CR Configuration Register. To speed-up the calibration procedure, the pre-charging phase can be skipped when not necessary (for example when consecutive single conversions are performed). The user can disable the pre-charging by setting the PRECH bit in the AD_CR Configuration Register.
The A/D peripheral converts the input voltage with a process of successive approximations using a fixed clock frequency derived from the 10 MHz internal oscillator, divided by a factor that depends on the speed mode: about 1.6 MHz in Fast Mode and 800 kHz in Slow Mode. The speed mode is chosen by the SCK bit of the AD_CR Configuration Register.
The conversion range is found between the analog $V_{S S}$ and the $A / D V_{\text {REF }}$ references. The $V_{\text {REF }}$ can be either internal, derived from the $\mathrm{V}_{\mathrm{DD}}$, or external by using the VREF pin. The external reference voltage allows the application of more precise and stable reference voltages. The two modes are selected by using the REF bit of the AD_CR Configuration Register.
Remark: the voltage applied to the VREF pin must be in the range $\mathrm{V}_{\mathrm{SS}}-\mathrm{V}_{\mathrm{DD}}$.
The external reference voltage $\mathrm{V}_{\text {REF }}$ is applied to the analog pin PBO. This pin shares the alternate functions with the first analog channel Ain0: if the

Figure 10.1 A/D Converter Structure

external reference mode is chosen the Ain0 channel is not used and the first channel of the conversion sequence becomes Ain1.
The converter uses a fully differential analog input configuration for a better noise immunity and precision performances.
Up to 8 multiplexed Analog Inputs are available. A single signal or a group of signals can be converted sequentially by simply programming the starting address of the last analog channel to be converted. Single or continuous conversion modes are available.
The result of the conversion of each A/D channel is stored in the 8-bit Input Register pairs (addresses from 41 to 56 ( $029 \mathrm{~h}-038 \mathrm{~h}$ )) according to the 8 -bit or 10 -bit mode. The resolution of conversion ( 8 or 10 bit) can be chosen by programming the RESOL bit of the AD_CR Configuration Register. In 8-bit mode the eight most significative bits (9:2) of the result of conversion is stored in the least significative byte of the register pair and the most significative is put to zero. In 10-bit mode the two most significative bits ( $9: 8$ ) are stored in the most significative byte of the register pair; the other bits (7:0) are stored in the least significative byte.
In 10-bit mode the result of the conversion must be read in two steps: the MSB and the LSB. The peripheral has been designed to avoid the side effects that can occur when the register are modified between the reading of the two bytes. In fact the latching of the input register pair is disabled after the reading of the first byte and it is enabled again after the reading of the second byte. User should pay attention to complete the two readings to guarantee the data of the conversion to be latched.
When the converted signal is higher than $\mathrm{V}_{\text {REF }}$, an overflow occurs. In this case the 8/10 bits result are all set to 1 and the A/D Overflow Register bit (address 39027 h ) corresponding to the channel is set to 1 . The bit is reset at the next conversion having no overflow occurrence.
ST52F510/F513/F514 Interrupt Unit provides one maskable channel for the End of Conversion and for the overflow control. It is possible to set the interrupt source on EOC or on overflow or on both by programming the INT0 and INT1 bits in the AD_CR Configuration Registers.
Note: the A/D Converter interrupts are not enabled unless the bit 0 (MSKAD) of the Configuration Register 0 (INT_MASK) is enabled (set to 1).
A Power-Down programmable bit (POW) allows the $A / D$ converter to be set to a minimum consumption idle status. A stabilization time is required, after the Power On, before accurate conversions can be performed.

### 10.2 Functional Description

The conversion is monotonic, meaning that the result never decreases if the analog input doesn't and never increases if the analog input doesn't.
If input voltage is less than Vss (voltage supply low) then the result is equal to 00 h .
The A/D converter is linear and the digital result of the conversion is provided by the following formula:

$$
\text { Digitalresult }=\frac{255 \times \text { InputVoltage }}{\text { ReferenceVoltage }}
$$

## Where Reference Voltage is $\mathrm{V}_{\text {ref }} \mathrm{V}_{\text {ss }}$.

The accuracy of the conversion is described in the Electrical Characteristics Section of the device datasheets.
The A/D converter is not affected by the WAIT mode.
When the ICU enters HALT mode with the A/D converter enabled, the converter is disabled until HALT mode is exited and the start-up delay has elapsed.

### 10.3 Operating Modes

Four main operating modes can be selected by setting the values of the CONT and SEQ bit in the A/D Configuration Register AD_CR.
10.3.1 One Channel Single Mode. In this mode (CONT=0, SEQ=0), the A/D provides an EOC signal after the end of the conversion of the specified channel; then the A/D waits for a new start event. The channel is identified by the bits $\mathrm{CH} 2-\mathrm{CH} 0$ in the Configuration Register AD_CR, while the bit STR is used to command the Start/ Stop.
10.3.2 Multiple Channels Single Mode. In this mode (CONT=0, SEQ=1) the A/D provides an EOC signal after the end of the channels sequence conversion identified by the three AD_CR Configuration Register bits $\mathrm{CH} 2-0$; then A/D waits for a new start event.
10.3.3 One Channel Continuous Mode. In this mode (CONT=1, SEQ=0) a continuous conversion flow is entered by a start event on the selected channel. At the end of each conversion, the relative Input Register is updated with the last conversion result, while the former value is lost.
The conversion continues until a stop command is executed by writing a ' 0 ' in the apposite AD_CR Configuration Register bit STR.
10.3.4 Multiple Channels Continuous Mode. In this mode (CONT=1, SEQ=1) a continuous conversion flow is entered by a start event on the selected channel sequence. The CH2-0 bits indicate the last channel of the sequence.
At the end of each conversion the relative Input Registers are updated with the last conversion results, while the former values are lost.
The conversion continues until a stop command is executed by writing a ' 0 ' in the apposite AD_CR Configuration Register bit STR.

### 10.4 Power Down Mode

Before enabling any A/D operation modes, set the Power On bit (POW) of the Configuration Register AD_CR to ' 1 ' and then start the A/D Converter by setting the STR bit. It is suggested to execute the pre-charging after the Power on to speed-up the auto calibration process. Clearing the Power On bit is useful when the A/D is not used, reducing the total chip power consumption. This state is also the reset configuration and it is forced by hardware when the core is in HALT state (after a HALT instruction execution).

### 10.5 A/D Converter Register Description

The following registers are related to the use of the A/D Converter.

### 10.5.1 A/D Converter Configuration Registers.

A/D Converter Control Register 1 (AD_CR1)
Configuration Register 8 (08h) Read/Write
Reset Value: 00000000 (00h)


| CH 2 | CH 1 | CH 0 | SCK | SEQ | POW | CONT | STR |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit 7-5: CH2-CH0 Channel Number
The number specified identifies the number of channels to be converted (Multiple Channel mode) or the channel to be converted (One Channel mode)

Bit 4: SCK A/D speed mode
0: Slow mode $(800 \mathrm{kHz})$
1: Fast mode ( 1600 kHz )

Bit 3: SEQ One/Multiple Channel Mode
0 : One Channel Mode
1: Multiple Channel Mode
Bit 2: POW A/D Converter Power Down/Up
0: Power down
1: Power up
Bit 1: CONT Single/Continuous Mode
0 : Single Mode
1: Continuous Mode

Bit 0: STR A/D Converter Start bit
0: A/D Converter stopped
1: A/D Converter started

A/D Converter Control Register 2 (AD_CR2) Configuration Register 47 (02Fh) Read/Write Reset Value: 00000000 (00h)

7
0

| - | - | - | PRECH | REF | RESOL | INT1 | INT0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit 7-5: not used

Bit 4: PRECH Pre-charging process on/off 0 : Pre-charge on (default)
1: Pre-charge off
Bit 3: REF Voltage Reference (VREF) source
0: Internal from Vdd
1: External from VREF pin
Bit 2: RESOL $8 / 10$ bits resolution
0: 10 bits
1:8 bits
Bit 1: INT1 Overflow interrupt mask
0 : interrupt disabled
1: interrupt enabled (if MSKAD=1)
Bit 0: INT0 End of Conversion interrupt mask 0 : interrupt disabled
1: interrupt enabled (if MSKAD=1)

### 10.5.2 Input Registers.

A/D Converter Overflow Register (AD_OVF)
Input Register 39 (027h) Read only
Reset Value: 00000000 (00h)

$$
7
$$

0

| OVF7 | OVF6 | OVF5 | OVF4 | OVF3 | OVF2 | OVF1 | OVF0 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit 7-0: OVF7-OVF0 Overflow Flag
0 : no overflow occurred in the last conversion 1: overflow occurred in the last conversion

## A/D Converter Data Registers

The converted digital values of the analog level applied to AIN0-7 pins, are buffered in the following register couples:

A/D Channel 0 data MSB (CHANO_H) Input Register 41 (029h) Read only Reset Value: 00000000 (00h)

A/D Channel 0 data LSB (CHANO_L) Input Register 42 (02Ah) Read only Reset Value: 00000000 (00h)

A/D Channel 1 data MSB (CHAN1_H) Input Register 43 (02Bh) Read only Reset Value: 00000000 (00h)

A/D Channel 1 data LSB (CHAN1_L) Input Register 44 (02Ch) Read only Reset Value: 00000000 (00h)

A/D Channel 2 data MSB (CHAN2_H) Input Register 45 (02Dh) Read only Reset Value: 00000000 (00h)

A/D Channel 2 data LSB (CHAN2_L) Input Register 46 (02Eh) Read only

Reset Value: 00000000 (00h)
A/D Channel 3 data MSB (CHAN3_H) Input Register 47 (02Fh) Read only Reset Value: 00000000 (00h)

A/D Channel 3 data LSB (CHAN3_L) Input Register 48 (030h) Read only Reset Value: 00000000 (00h)

A/D Channel 4 data MSB (CHAN4_H) Input Register 49 (031h) Read only Reset Value: 00000000 (00h)

A/D Channel 4 data LSB (CHAN4_L) Input Register 50 (032h) Read only Reset Value: 00000000 (00h)

A/D Channel 5 data MSB (CHAN5_H) Input Register 51 (033h) Read only Reset Value: 00000000 (00h)

A/D Channel 5 data LSB (CHAN5_L)
Input Register 52 (034h) Read only Reset Value: 00000000 (00h)

A/D Channel 6 data MSB (CHAN6_H) Input Register 53 (035h) Read only Reset Value: 00000000 (00h)

A/D Channel 6 data LSB (CHAN6_L) Input Register 54 (036h) Read only Reset Value: 00000000 (00h)

A/D Channel 7 data MSB (CHAN7_H) Input Register 55 (037h) Read only Reset Value: 00000000 (00h)

A/D Channel 7 data LSB (CHAN7_L)
Input Register 56 (038h) Read only Reset Value: 00000000 (00h)

## 11 WATCHDOG TIMER

### 11.1 Functional Description

The Watchdog Timer (WDT) is used to detect the occurrence of a software fault, usually generated by external interference or by unforeseen logical conditions, which causes the application program to abandon its normal sequence. The WDT circuit generates an ICU reset on expiry of a programmed time period, unless the program refreshes the WDT before the end of the programmed time delay. Sixteen different delays can be selected by using the WDT configuration register.
After the end of the delay programmed by the configuration register, if the WDT is active, it starts a reset cycle pulling the reset signal low.
Once the WDT is activated, the application program has to refresh the counter (by the WDTRFR instruction) during normal operation in order to prevent an ICU reset.
In ST52F510/F513/F514 devices it is possible to choose between "Hardware" or "Software" Watchdog. The Hardware WDT allows the counting to avoid unwanted stops for external interferences. The first mode is always enabled unless the Option Byte 4 (WDT_EN) is written with a special code (10101010b): only this code can switch the WDT in "Software" Mode, the other 255 possibilities keep the "Hardware" Mode enabled.
The WDT is started and refreshed by using the WDTRFR instruction. When the software mode is enabled, the WDTSLP instruction stops the WDT avoiding timeout resets.
When the WDT is in Hardware Mode, neither the WDTSLP instruction nor external interference can stop the counting. The "Hardware" WDT is always enabled after a Reset.

The working frequency of WDT (PRES CLK in the Figure 11.1) is equal to the clock master. The clock master is divided by 500, obtaining the WDT CLK signal that is used to fix the timeout of the WDT.
According to the WDT_CR Configuration Register values, a WDT delay between 0.1 ms and 937.5 ms can be defined when the clock master is 5 MHz . By changing the clock master frequency the timeout delay can be calculated according to the configuration register values. The first 4 bits of the WDT_CR register are used, obtaining 16 different delays.

Table 11.1 Watchdog Timing Range ( 5 MHz )

| WDT timeout period (ms) |  |
| :---: | :---: |
| $\min$ | 0.1 |
| $\max$ | 937.5 |

### 11.2 Register Description

## SW Watchdog Enable (WDT_EN)

Option Byte 4 (04h)
Reset Value: 00000000 (00h)
7
0

| WDTEN7 | WDTEN6 | WDTEN5 | WDTEN4 | WDTEN3 | WDTEN2 | WDTEN1 | WDTEN0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: WDTEN7-0 SW Watchdog Enable byte Writing the code 10101010 in this byte the Software Watchdog mode is enabled.

Figure 11.1 Watchdog Block Diagram


Watchdog Control Register (WDT_CR)
Configuration Register 7 (07h) Read/Write Reset Value: 00000001 (00h) 7


Bit 3-0: D3-0 Watchdog Clock divisor factor bits The Watchdog Clock (WDT CLK) is divided by the numeric factor determined by these bits, according with Table 11.2 and the following formula:
$\operatorname{Timeout}(\mathrm{ms})=\frac{5 \times 10^{5} \times \text { DivisionFactor }}{\operatorname{Clock}(\mathrm{MHz})}$

Bit 7-4: Not Used

Table 11.2 Watchdog Timeout configuration examples

| WDT_CR(3:0) | Division Factor | Timeout Values (ms) |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  | 5 MHz | 10 MHz | 20MHz |
| 0000 | 1 | 0.1 | 0.05 | 0.025 |
| 0001 | 625 | 62.5 | 31.25 | 15.625 |
| 0010 | 1250 | 125 | 62.5 | 31.25 |
| 0011 | 1875 | 187.5 | 93.75 | 46.875 |
| 0100 | 2500 | 250 | 125 | 62.5 |
| 0101 | 3125 | 312.5 | 156.25 | 78.125 |
| 0110 | 3750 | 375 | 187.5 | 93.75 |
| 0111 | 4375 | 437.5 | 218.75 | 109.375 |
| 1000 | 5000 | 500 | 250 | 125 |
| 1001 | 5625 | 562.5 | 281.25 | 140.625 |
| 1010 | 6250 | 625 | 312.5 | 156.25 |
| 1011 | 6875 | 687.5 | 343.75 | 171.875 |
| 1100 | 7500 | 750 | 375 | 187.5 |
| 1101 | 8125 | 812.5 | 406.25 | 203.125 |
| 1110 | 8750 | 875 | 437.5 | 218.75 |
| 1111 | 9375 | 937.5 | 468.75 | 234.375 |

## 12 PWM/TIMERS

### 12.1 Introduction

ST52F510/513/514 offers two on-chip PWM/Timer peripherals. All ST52F510/513/514 PWM/Timers have the same internal structure. The timer consists of a 16-bit counter with a 16-bit programmable Prescaler, giving a maximum count of $2^{32}$ (see Figure 12.1).
Each timer has two different working modes, which can be selected by setting the correspondent bit TxMOD of the PWMx_CR1 Configuration Register: Timer Mode and PWM (Pulse Width Modulation) Mode.
All the Timers have Autoreload Functions; in PWM Mode the reload value can be set by the user.
Each timer output is available on the apposite external pins configured in Alternate Function and in one of the Output modes.
PWM/Timer 0 can also use external START/STOP signals in order to perform Input capture and Output compare, external RESET signal, and external CLOCK to count external events: TSTRT, TRES and TCLK pins. In addition, the START/ STOP and RESET signals have configurable polarity (falling or rising edge).

Remark: To use TRES, TSTRT, TCLK external signals the related pins must be configured in Alternate Function and in one of Input modes.
For each timer, the contents of the 16 -bit counter are incremented on the Rising Edge of the 16-bit prescaler output (PRESCOUT) and it can be read at any instant of the counting phase by accessing the Input Registers PWMx_COUNT_IN_x; the value is stored in two 8 -bit registers (MSB and LSB) for each PWM/Timer.

Figure 12.1 PWM/Timer Counter block diagram

The Input Registers couple PWMx_CAPTURE_x store the counter value after the last Stop signal (only Timer Mode). The counter value is not stored after a Reset Signal.
The peripheral status can also be read from the Input Registers (one for each Timer). These registers report START/STOP, SET/RESET status, TxOUT signal and the counter overflow flag. This last signal is set after the first EOC and it is reset by a Timer RESET (internal or external).

### 12.2 Timer Mode

Timer Mode is selected writing 0 in the TxMOD bit. Each Timer requires three signals: Timer Clock (TMRCLKx), Timer Reset (TxRES) and Timer Start (TxSTRT) (see Figure 12.1). Each of these signals can be generated internally, and/or externally only for Timer 0, by using TRES, TSTRT and TCLK pins.
The Prescaler output (PRESCOUT) increments the Counter value on the rising edge. PRESCOUT is obtained from the internal clock signal (CLKM) or, only for TIMERO, from the external signal provided on the apposite pin.
Note: The external clock signal applied on the TCLK pin must have a frequency that is at least two times smaller than the internal master clock.
The prescaler output period can be selected by setting the TxPRESC bits with one of the 17 division factors available. TMRCLK frequency is divided by a factor equal to the power of two of the prescaler values (up to $2^{16}$ ).
TxRES resets the content of the 16 -bit counter to zero. It is generated by writing 0 in the TxRES bit of the PWMx_CR1 Configuration Register and/or it can be driven by the TRES pin if configured (only Timer0).


Figure 12.2 Timer 0 External Start/Stop Mode


TxSTRT signal starts/stops the Timer from counting only if the peripherals are configured in Timer mode. The Timers are started by writing 1 in the TXSTRT bit of the PWMx_CR1 and are stopped by writing 0 . This signal can be generated internally and/or externally by forcing the TSTRT pin (only TIMERO).
TIMER 0 START/STOP can be given externally on the TSTRT pin. In this case, the TOSTRT signal allows the user to work in two different configurable modes (see Figure 12.2):

- LEVEL (Time Counter): If the TOSTRT signal is high, the Timer starts counting. When the TOSTRT is low the timer stops counting and the 16 -bit current value is stored in the PWMO_COUNT_IN_x Input Registers couple.
- EDGE (Period Counter): After reset, on the first TOSTRT rising edge, TIMER 0 starts counting and at the next rising edge it stops. In this manner the period of an external signal may be measured.
The same above mentioned modes, can be used to reset the Timer0 by using the TRES pin signal.
The polarity of the TOSRTR Start/Stop signal can be changed by setting the STRPOL and RESPOL bits in the INT_POL Configuration Register (01h bit 3 and 4). When these bits are set, the PWM/Timer 0 is Started/Set on the low level or in the falling edge of the signal applied in the pins.
The Timer output signal, TxOUT, is a signal with a frequency equal to the one of the 16 bit-Prescaler output signal, PRESCOUTx, divided by a 16-bit counter set by writing the Output Register couple PWMx_COUNT_OUT_x.

Note: the contents of these registers upgrades the Timer counter after it stops counting. Since the register couple is written in two steps this can cause side effects. In order to avoid this, the user should write the MSB before writing the LSB: actually, the 16 -bit value is latched in parallel when the LSB is written. By writing only the LSB (and MSB equal to 0), the PWM/Timer is used as with an 8 bit counter.

Warning: in Timer Mode the Reload Register couple PWMx_RELOAD_x (see PWM mode) must be set to the higher value FFFFh (65535) otherwise it can affect the count duration.

There can be two types of TxOUT waveforms:

- type 1: TxOUT waveform equal to a square wave with a $50 \%$ duty-cycle
- type 2: TxOUT waveform equal to a pulse signal with the pulse duration equal to the Prescaler output signal.

Figure 12.3 TxOUT Signal Types


Figure 12.4 PWM Mode with Reload


### 12.3 PWM Mode

The PWM working mode for each timer is obtained by setting the TxMOD bit of the Configuration Register PWMx_CR1.
The TxOUT signal in PWM Mode consists of a signal with a fixed period, whose duty cycle can be modified by the user.
The TxOUT period is fixed by setting the 16 -bit Prescaler bits (TxPRESC) in the PWMx_CR2 and the 16 -bit Reload value by writing the relative Output Registers couple PWMx_RELOAD_x. The 16-bit Prescaler divides the master clock CLKM by powers of two, determining the maximum length period.
Reload determines the maximum value that the counter can count before starting a new period. The use of the two 16 -bit values allows the TxOUT period to be set with more precision when needed. By setting the Reload value the counting resolution decreases. In order to obtain the maximum resolution, Reload value should be set to OFFFFh and the period corresponds to the one established by the Prescaler value.
The value set in the 16 -bit counter by writing the Counter Output Registers couple, determines the duty-cycle: when count reaches the Counter value the TxOUT signal changes from high to low level.
The period of the PWM signal is obtained by using the following formula:
T=PWMx_RELOAD * $2^{\text {TxPRESC }}$ TMRCLKx
where TxPRES equals the value set in the TxPRESC bits of the PWMx_CR2 Configuration Register and TMRCLKx is the period of the Timer clock that drives the Prescaler.
The duty cycle of the PWM signal is obtained by the following formula:

$$
d_{\text {cycle }}=\frac{T_{o n}}{T}=\frac{P W M x \operatorname{COUNT}}{P W M x R E L O A D}
$$

Note: the PWM_x_COUNT value must be lower than or equal to the PWM_X_RELOAD value. When it is equal, the TxOUT signal is always at high level. If the Output Register PWM_x_COUNT is 0 , TxOUT signal is always at a low level.
By using a 24 MHz clock a PWM frequency that is close to 100 Khz can be obtained.
The TIMERO clock CLKM can also be supplied with an external signal, applied on the TCLK pin, which must have a frequency that is at least two times smaller than the internal master clock.

Note: he Timers have to complete the previous counting phase before using a new value of the Counter. If the Counter value is changed during counting, the new values of the timer Counter are only used at the end of the previous counting phase. The Counter buffer is written in two steps (one byte per time) and is latched only after the LSB is written. In order to avoid side effects, the user should write the MSB before writing the LSB. By only writing the LSB, the PWM/Timer is used as with a 8 bit counter. The same mechanism is applied to the two bytes of Reload but, differently of the Counter it is set immediately. Nevertheless, it is recommended that the Reload value be written when the Timer is stopped in order to avoid incongruence with the Counter value. The same recommendation is made when reading the two bytes of the counter: It is performed in two steps, so if the timer is running, the carry of the LSB to the MSB can cause the wrong 16 -bit value reading. A Reload value greater than 1 must always be used.

When the Timers are in Reset status, or when the device is reset, the TxOUT pins goes in threestate. If these outputs are used to drive external devices, it is recommended that the related pins be left in the default configuration (Input threestate) or change them in this configuration.
In PWM mode the PWM/Timers can only be Set or Reset: Start/Stop signals do not affect the Timers. TxRES resets the content of the 16 -bit counter to zero. It is generated by writing 0 in the corresponding TxRES bit of the PWMx_CR1 Configuration Register and/or it can be driven by the TRES pin if it is configured (only Timer0).
12.3.1 Simultaneous Start. The PWM/Timers can be started simultaneously when working in PWM mode. The TOSYNC and T1SYNC bits in PWMO_CR3 Configuration Registers mask the reset of each timer; after enabling each single PWM/Timer. They are started by putting off the mask with a single writing in the PWMO_CR3 Register.
Simultaneous start is also possible in Timer mode. The timers start counting simultaneously, but the output pulses are generated according to the modality configured (square or pulse mode).

### 12.4 Timer Interrupts

The PWM/Timer can be programmed to generate an Interrupt Request, both on the falling and the rising of the TxOUT signal and when there's a STOP signal (external or internal).
By using the TxIES, TxIER and TxIEF bits of the Configuration Registers PWMx_CR1, the interrupt sources can be switched on/off. All the interrupt sources may be activated at the same time: sources can be distinguished by reading the PWMx_STATUS Input Register.
The interrupt on the falling edge corresponds to half of a counting period in Timer mode when the waveform is set to Square Wave and to the end of the Ton phase in PWM mode.

Note: when the PWM Counter is set to 0 or 65535, the interrupt occurs at the end of each control period.
In order to be active, the PWM/Timers interrupts must be enabled by writing the Interrupt Mask Register (INT_MASK) in the Configuration Register Space, bits MSKT0 And MSKT1.

### 12.5 PWM/Timer 0 Register Description

The following registers are related to the use of the PWM/Timer 0.

### 12.5.1 PWM/Timer 0 Configuration Registers.

PWM/Timer 0 Control Register 1 (PWMO_CR1)
Configuration Register 9 (09h) Read/Write
Reset Value: 00000000 (00h)
7
0

| TOMOD | TOIES | TOIEF | TOIER | STRMOD | TOSTRT | RESMOD | TORES |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7: TOMOD PWM/Timer 0 Mode
0: Timer Mode
1: PWM Mode

Bit 6: TOIES Interrupt on Stop signal Enable
0 : interrupt disabled
1 : interrupt enabled
Bit 5: TOIEF Interrupt on TOOUT falling Enable
0 : interrupt disabled
1: interrupt enabled
Bit 4: TOIER Interrupt on TOOUT rising Enable
0 : interrupt disabled
1: interrupt enabled

Bit 3: STRMOD Start signal mode
0 : start on level
1: start on edge

Bit 2: TOSTRT PWM/Timer 0 Start bit
0 : Timer 0 stopped
1: Timer 0 started

Bit 1: RESMOD Reset signal mode
0 : reset on level
1: reset on edge

Bit 0: TORES PWM/Timer 0 Reset bit
0 : PWM/Timer 0 reset
1: PWM/Timer 0 set

PWM/Timer 0 Control Register 2 (PWMO_CR2)
Configuration Register 10 (0Ah) Read/Write
Reset Value: 00000000 (00h)

$$
\begin{array}{lll}
7 & 4 & 0
\end{array}
$$

| - | - | TOWAV | TOPRESC |
| :--- | :--- | :--- | :--- |

Bit 7-6: Not Used
Bit 5: TOWAV TOOUT Waveform
0 : pulse (type2)
1: square (type1)
Bit 4-0: TOPRESC PWM/Timer 0 Prescaler
The PWM/Timer 0 clock is divided by a factor equal to $2^{\text {TOPRESC }}$. The maximum value allowed for TOPRESC is 10000 (010h).

PWM/Timer 0 Control Register 3 (PWMO_CR3)
Configuration Register 11 (0Bh) Read/Write
Reset Value: 00000000 (00h)

| 7 | 4 |  | 2 | 0 |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| TISYNC | - | TOSYNC | TOCKS | STRSRC | RESSRC |

Bit 7: TOSYNC PWM/Timer 0 Set/Reset mask
0: Set/Reset activated
1: Set/Reset masked
Bit 6: not used
Bit 5: T1SYNC PWM/Timer 1 Set/Reset mask
0: Set/Reset activated
1: Set/Reset masked
Bit 4: TOCKS PWM/Timer 0 Clock Source
0: Internal clock
1: External Clock from TCLK
Bit 3-2: STRSRC PWM/Timer 0 Start signal source
00: Internal from TOSTRT bit
01: External from TSTRT pin
10: Both internal and external

Bit 1-0: RESSRC PWM/Timer 0 Reset source
00: Internal from TORES bit
01: External from TRES pin
10: Both internal and external

## Interrupt Polarity Register (INT_POL)

Configuration Register 1 (01h) Read/Write
Reset Value: 00000000 (00h)
7 0

| - | - | - | RESPOL | STRPOL | POLPB | POLPA | POLNMI |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-5: Not Used
Bit 4: RESPOL Reset signal polarity
0 : Reset on low level/rising edge
1: Reset on high level/falling edge
Bit 3: STRPOL Start signal polarity
0 : Start on high level/rising edge
1: Start on low level/falling edge
Bit 2-0: See Interrupt Registers Description

### 12.5.2 PWM/Timer 0 Input Registers.

## PWM/Timer 0 Counter High Input Register (PWMO_COUNT_IN_H)

Input Register 21 (015h) Read only
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| TOCI15 | TOC114 | TOCl13 | T0CI12 | TOCl11 | TOC110 | T0C19 | TOCl8 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T0CI15-8 PWM/Timer 0 Counter MSB
In this register the current value of the Timer 0 Counter MSB can be read.

## PWM/Timer 0 Counter Low Input Register (PWMO_COUNT_IN_L)

Input Register 22 (016h) Read only
Reset Value: 00000000 (00h)
7 0

| TOC17 | TOC16 | TOC15 | T0C14 | TOC13 | TOC12 | TOCI1 | TOCIO |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T0CI7-0 PWM/Timer 0 Counter MSB
In this register the current value of the Timer 0 Counter LSB can be read.

PWM/Timer 0 Status Register (PWMO_STATUS) Input Register 23 (017h) Read only
Reset Value: 00000000 (00h)
7
0


Bit 7-4: Not Used

Bit 3: TOOVFL PWM/Timer 0 counter overflow flag 0 : no overflow occurred since last reset
1: overflow occurred

Bit 2: TOOUT TOOUT pin value
0 : TOOUT pin is at logical level 0
1: TOOUT pin is at logical level 1
Bit 1: TORST Reset Status
0 : PWM/Timer 0 is reset
1: PWM/Timer 0 is set

Bit 0: TOSST Start Status
0 : PWM/Timer 0 is stopped
1: PWM/Timer 0 is running

## PWM/Timer 0 Capture High Input Register

 (PWMO_CAPTURE_H)Input Register 24 (018h) Read only
Reset Value: 00000000 (00h)

| T0CP15 | T0CP14 | T0CP13 | T0CP12 | T0CP11 | T0CP10 | T0CP9 | T0CP8 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T0CP15-8 PWM/Timer 0 Capture MSB

In this register the counter value after the last stop can be read.

PWM/Timer 0 Capture Low Input Register (PWMO_CAPTURE_L)
Input Register 25 (019h) Read only
Reset Value: 00000000 (00h)
7
0

| TOCP7 | TOCP6 | TOCP5 | T0CP4 | T0CP3 | TOCP2 | T0CP1 | T0CP0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T0CP7-0 PWM/Timer 0 Capture LSB

In this register the counter value after the last stop can be read.

### 12.5.3 PWM/Timer 0 Output Registers.

## PWM/Timer 0 Counter High Output Register

 (PWMO_COUNT_OUT_H)Output Register 7 (07h) Write only
Reset Value: 00000000 (00h)
7
0

| T0CO15 | T0CO14 | T0CO13 | T0CO12 | T0CO11 | T0CO10 | T0CO9 | T0CO8 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T0CO15-8 PWM/Timer 0 Counter MSB This register is used to write the Timer 0 Counter value (MSB).
Note: this register is latched after writing the LSB part (PWM_COUNT_OUT_L: see below). For this reason this register must be written before the LSB.

## PWM/Timer 0 Counter Low Output Register (PWMO_COUNT_OUT_L)

Output Register 8 (08h) Write only
Reset Value: 00000000 (00h)
7
0

| T0CO7 | TOCO6 | T0CO5 | T0CO4 | TOCO3 | TOCO2 | T0CO1 | T0CO0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T0CO7-0 PWM/Timer 0 Counter MSB

This register is used to write the Timer 0 Counter value (LSB).
Note: writing this register, the PWMO_COUNT_OUT_x couple is latched in the internal registers of the peripherals. For this reason, this register should be written after the MSB one.

PWM/Timer 0 Reload High Output Register (PWMO_RELOAD_H)<br>Output Register 9 (09h) Write only<br>Reset Value: 11111111 (OFFh)

7

## 0

| T0REL15 | TOREL14 | TOREL13 | T0REL12 | TOREL11 | T0REL10 | T0REL9 | T0REL8 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit 7-0: TOREL15-8 PWM/Timer 0 Reload MSB
This register is used to write the Timer 0 Reload value (MSB).

Note: this register is latched after writing the LSB part (PWMO_RELOAD_L: see below). For this reason, this register must be written before the LSB.

## PWM/Timer 0 Reload Low Output Register (PWMO_RELOAD_L)

Output Register 10 (0Ah) Write only
Reset Value: 11111111 (OFFh)
7
0

| TOREL7 | TOREL6 | TOREL5 | TOREL4 | TOREL3 | TOREL2 | TOREL1 | TOREL0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: TOREL7-0 PWM/Timer 0 Reload LSB
This register is used to write the Timer 0 Reload value (LSB).

Note: by writing this register, the PWMO_RELOAD_x couple is latched in the internal registers of the peripherals. For this reason this register should be written after the MSB one.

### 12.6 PWM/Timer 1 Register Description

The following registers are related to the use of the PWM/Timer 1.

### 12.6.1 PWM/Timer 1 Configuration Registers.

PWM/Timer 1 Control Register 1 (PWM1_CR1)
Configuration Register 12 (0Ch) Read/Write Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| T1MOD | T1IES | T1IEF | T1IER | - | T1STRT | - | T1RES |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7: T1MOD PWM/Timer 1 Mode
0: Timer Mode
1: PWM Mode

Bit 6: T1IES Interrupt on Stop signal Enable
0 : interrupt disabled
1 : interrupt enabled
Bit 5: T1IEF Interrupt on T1OUT falling Enable 0 : interrupt disabled
1: interrupt enabled
Bit 4: T1IER Interrupt on T1OUT rising Enable
0 : interrupt disabled
1: interrupt enabled
Bit 3: not used
Bit 2: T1STRT PWM/Timer 1 Start bit
0 : Timer 0 stopped
1: Timer 0 started
Bit 1: not used
Bit 0: T1RES PWM/Timer 1 Reset bit
0 : PWM/Timer 0 reset
1: PWM/Timer 0 set

PWM/Timer 1 Control Register 2 (PWM1_CR2)
Configuration Register 13 (0Dh) Read/Write
Reset Value: 00000000 (00h)
7
5
0

| - | - | T1WAV | T1PRESC |
| :--- | :--- | :--- | :--- |

Bit 7-6: Not Used
Bit 5: T1WAV T1OUT Waveform
0: pulse (type2)
1: square (type1)

Bit 4-0: T1PRESC PWM/Timer 1 Prescaler The PWM/Timer 1 clock is divided by a factor equal to $2^{\text {T1PRESC }}$. The maximum value allowed for T1PRESC is 10000 (010h).

### 12.6.2 PWM/Timer 1 Input Registers.

PWM/Timer 1 Counter High Input Register
(PWM1_COUNT_IN_H)
Input Register $26(01 \mathrm{Ah})$ Read only
Reset Value: $00000000(00 \mathrm{~h})$ 7

| T1Cl15 | T1Cl14 | T1Cl13 | T1Cl12 | T1Cl11 | T1Cl10 | T1Cl9 | T1Cl8 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T1Cl15-8 PWM/Timer 1 Counter MSB

In this register the current value of the Timer 1 Counter MSB can be read.

```
PWM/Timer 1 Counter Low Input Register (PWM1_COUNT_IN_L)
Input Register 27 (01Bh) Read only
Reset Value: 00000000 (00h)
```

7
7

|  | 0 |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| T 1 Cl 7 | T 1 Cl 6 | T 1 Cl 5 | T 1 Cl 4 | T 1 Cl 3 | T 1 Cl 2 | T 1 Cl 1 | T 1 Cl 0 |

Bit 7-0: T1CI7-0 PWM/Timer 1 Counter LSB
In this register the current value of the Timer 1 Counter LSB can be read.

PWM/Timer 1 Status Register (PWM1_STATUS) Input Register 28 ( 01 Ch ) Read only Reset Value: 00000000 (00h)

7
0

| - | - | - | - | T1OVFL | T1OUT | T1RST | T1SST |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-4: Not Used

Bit 3: T10VFL PWM/Timer 1 counter overflow flag 0 : no overflow occurred since last reset 1: overflow occurred

Bit 2: T1OUT T1OUT pin value
0 : T1OUT pin is at logical level 0
1: T1OUT pin is at logical level 1
Bit 2: T1RST Reset Status
0 : PWM/Timer 1 is reset
1: PWM/Timer 1 is set

Bit 2: T1SST Start Status
0 : PWM/Timer 1 is stopped
1: PWM/Timer 1 is running

PWM/Timer 1 Capture High Input Register (PWM1_CAPTURE_H)
Input Register 29 (01Dh) Read only
Reset Value: 00000000 (00h)
7 0

| T1CP15 | T1CP14 | T1CP13 | T1CP12 | T1CP11 | T1CP10 | T1CP9 | T1CP8 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T1CP15-8 PWM/Timer 1 Capture MSB

In this register the counter value after the last stop can be read.

PWM/Timer 1 Capture Low Input Register (PWM1_CAPTURE_L)
Input Register 30 (01Eh) Read only
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| T1CP7 | T1CP6 | T1CP5 | T1CP4 | T1CP3 | T1CP2 | T1CP1 | T1CP0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T1CP7-0 PWM/Timer 1 Capture LSB
In this register the counter value after the last stop can be read.

### 12.6.3 PWM/Timer 1 Output Registers.

PWM/Timer 1 Counter High Output Register (PWM1_COUNT_OUT_H)
Output Register 11 (0Bh) Write only
Reset Value: 00000000 (00h)
7 0

| T 1 CO 15 | T 1 CO 14 | T 1 CO 13 | T 1 CO 12 | T 1 CO 11 | T 1 CO 10 | T 1 CO 9 | T 1 CO 8 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T1CO15-8 PWM/Timer 1 Counter MSB
This register is used to write the Timer 1 Counter value (MSB).
Note: this register is latched after writing the LSB part (PWM1_COUNT_OUT_L: see below). For this reason, this register must be written before the LSB.

## PWM/Timer 1 Counter Low Output Register

 (PWM1_COUNT_OUT_L)Output Register 12 (0Ch) Write only
Reset Value: 00000000 (00h)


| T 1 CO 7 | T 1 CO 6 | T 1 CO 5 | T 1 CO 4 | T 1 CO 3 | T 1 CO 2 | T 1 CO 1 | T 1 CO 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T1CO7-0 PWM/Timer 0 Counter MSB
This register is used to write the Timer 1 Counter value (LSB).

Note: by writing this register, the PWM1_COUNT_OUT_x couple is latched in the internal registers of the peripherals. For this reason this register should be written after the MSB one.

## PWM/Timer 1 Reload High Output Register (PWM1_RELOAD_H)

Output Register 13 (0Dh) Write only
Reset Value: 11111111 (0FFh)
7 0

| T1REL15 | T1REL14 | T1REL13 | T1REL12 | T1REL11 | T1REL10 | T1REL9 | T1REL8 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T1REL15-8 PWM/Timer 0 Reload MSB
This register is used to write the Timer 1 Reload value (MSB).
Note: this register is latched after writing the LSB part (PWM1_RELOAD_L: see below). For this reason, this register must be written before the LSB.

## PWM/Timer 1 Reload Low Output Register (PWMO_RELOAD_L)

Output Register 14 (0Eh) Write only
Reset Value: 11111111 (OFFh)
7 0

| T1REL7 | T1REL6 | T1REL5 | T1REL4 | T1REL3 | T1REL2 | T1REL1 | T01REL0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: T1REL7-0 PWM/Timer 1 Reload LSB

This register is used to write the Timer 1 Reload value (LSB).
Note: by writing this register, the PWM1_RELOAD_x couple is latched in the internal registers of the peripherals. For this reason, this register should be written after the MSB one.

## 13 SERIAL COMMUNICATION INTERFACE

The Serial Communication Interface (SCI) integrated into ST52F510/F513/F514 provides a general purpose shift register peripheral, several widely distributed devices to be linked, through their SCl subsystem. SCl gives a serial interface providing communication with the speed from less than 300 up to over 115200 baud, and a flexible character format.
SCI is a full-duplex UART-type asynchronous system with standard Non Return to Zero (NRZ) format for the transmitted/received bit. The length of the transmitted word is 10/11 bits (1 start bit, 8/ 9 data bits, 1 stop bit).
SCl is composed of three modules: Receiver, Transmitter and Baud-Rate Generator.

### 13.1 SCI Receiver block

The SCI Receiver block manages the synchronization of the serial data stream and stores the data characters. The SCl Receiver is mainly composed of two sub-systems: Recovery Buffer Block and SCDR_RX Block.
SCI receives data deriving from the RX pin and drives the Recovery Buffer Block, which is a highspeed shift register operating at a clock frequency (CLOCK_RX) 16 times higher than the fixed baud rate (CLOCK_TX). This sampling rate, higher than the Baud Rate clock, detects the START condition, Noise error and Frame error.
When the SCI Receiver is in IDLE status, it is waiting for the START condition, which is obtained with a logic level of 0 , consecutive to a logic level 1. This condition is detected if, with the fixed
sampling time, a logic level 0 is sampled after three logic levels of 1 .
The recognition of the START bit forces the SCI Receiver Block to start a data acquisition sequence.
The data acquisition sequence is configured by the apposite Configuration Register, allowing the following data frame formats (see Figure 13.1):

Figure 13.1 SCI transmitted word structures


- 8 bit length, 1 stop bit, no parity bit
- 8 bit length, 2 stop bit, no parity bit
- 8 bit length, 1 stop bit, with parity bit
- 9 bit length, 1 stop bit, no parity bit

The parity bit (if used) can be configured for even or odd parity check. If the 9 -bit length format is configured, this bit is used in transmission for the ninth bit (see below). The ninth bit received can be read in the R8 bit of the SCI Status Register, address 37 (035h) bit 2 (see Figure 13.3).

Figure 13.2 SCI Block Diagram


Recognition of a STOP condition transfers data received from the Recovery Buffer to the SCDR_RX buffer, adding the eventual ninth data bit. After this operation, RXF flag (bit 5) of SCl Status Input Register is set to logic level 1. The Control Unit reads data from the SCDR_RX buffer (in read-only mode) by reading the SCI_IN Input Register (address 36 024h) with the LDRI instruction and provides a reset at logic level 0 to the RXF flag.
If data of the Recovery Buffer is ready to be transferred into the SCDR_RX buffer, but the previous one has not been read by the Core, an OVERRUN Error takes place: the SCI Status Register flag OVERR (bit 4) indicates the error condition. In this case, information that is stored in the SCDR_RX buffer is not altered, but the one that has caused the OVERRUN error can be overwritten by new data deriving from the serial data line.

### 13.1.1 Recovery Buffer Block .

This block is structured as a synchronized finite state machine on the CLOCK_RX signal.
When the Recovery Buffer Block is in IDLE state it waits for the reception of the correct 1 and 0 sequence representing START.
Recognition takes place by sampling the input RX at CLOCK_RX frequency, which has a frequency that is 16 times higher than CLOCK_TX. For this reason, while the external transmitter sends a single bit, the Recovery Buffer Block samples 16 states (from SAMPLE1 to SAMPLE16).
Analysis of the RX input signal is carried out by checking three samples for each bit received.
If these three samples are not equal, then the noise error flag, NSERR (bit 7), of SCI Status Register is set to 1 and the data received value will be the one assumed by the majority of the samples.

The procedure described above, allows SCI not to becomes IDLE, because of a limited noise due to an erroneous sampling, the transmission is recognized as correct and the noise flag error is set.
At the end of the cycle of the reception of a bit, the Recovery Buffer Block will repeat the same steps 9 times: one step for each bit received, plus one for the stop acquisition (10 times in case of 9 -bit data, double stop or parity check).
At the end of data reception the Recovery Buffer Block will supply information about eventual frame errors by setting the 1 FRERR flag (bit 6) of the SCI Status Register to 1.
A frame error can occur if the parity check hasn't been successfully achieved or if the STOP bit has not been detected.
If the Recovery Buffer Block receives 10 consecutive bits at logic level 0, a Line Break condition occurs and the related Interrupt Request is sent.

### 13.1.2 SCDR_RX Block.

It is a finite state machine synchronized with the clock master signal, CKM.
The SCDR_RX block waits for the signal of complete reception from the Recovery Buffer in order to load the word received. Moreover, the SCDR_RX block loads the values of FRERR and NSERR flag bits of the Status Register, and sets the RXF flag to 1 .
By using the LDRI instruction data is transferred to Register File and RXF flag is reset to 0 , to indicate that the SCDR_RX block is empty.
If new data arrives before the previous one has been transferred to Register File, the overrun error occurs and the OVERR flag of Status Register is set to 1 .

Figure 13.3 SCI Status Register


### 13.2 SCI Transmitter Block

The SCI Transmitter Block consists of the following blocks: SCDR_TX and SHIFT REGISTER, synchronized, respectively, with the clock master signal (CKM) and the CLOCK_TX.
The whole block receives the settings for the following transmission modes through the Configuration Register:
■ 8 bit length, 1 stop bit, no parity bit

- 8 bit length, 2 stop bit, no parity bit
- 8 bit length, 1 stop bit, with parity bit

■ 9 bit length, 1 stop bit, no parity bit
In case of 9 bit frame transmission, the most significative bit arrives through the bit PAR/T8 (bit
2) of the SCI_CR1 Configuration Register. In an 8bit transmission, instead, this bit is used to configure the data format: in particular to choose the polarity control (even or odds) to implement the parity check (see above).
After a RESET, the SCDR TX block is in IDLE state until it receives an enabling signal by writing the TXSTRT bit of the SCI_CR2 Configuration Register.
The data is loaded on the Peripheral Register SCI_OUT (address 23 017h) by using the instruction LPPR, LDPI or LDPE. If the transmission is enabled, the data to be transmitted is transferred from the Output Register to SCDR_TX block and the TXEM flag (bit 1) of the SCl Status Register is reset to 0 to indicate SCDR_TX block is full.
If the core supplies new data, this could not be loaded in the SCDR TX block until the current data has not been unloaded on the Shift Register block. Meaning that only when TXEM is 1 data can be loaded in the SCDR_TX Block.
When the SHIFT REGISTER Block loads the data to be transmitted on an internal buffer, the TXEND flag (bit 0) of the SCI Status Register is reset to 0 to indicate the beginning of a new transmission. At the end of transmission TXEND is set to 1 , allowing new data coming from SCDR_TX to be loaded in the SHIFT REGISTER.

It is important to underline that TXEND = 1 does not mean SCDR_TX is ready to receive a new data. For this reason, it is better to utilize the TXEM signal to synchronize the load instruction to the SCI TRANSMITTER block
If the TXSTRT bit is reset, the transmission is stopped, but the SCI Transmitter block completes the transmission in progress before resetting.

### 13.3 Baud Rate Generator Block

The Baud Rate Generator Block performs the division of the clock master signal (CKM) in a set of synchronism frequencies for the serial bit reception/transmission on the external line.
Reception frequency (CLOCK_RX) is 16 times higher than the transmission frequency (CLOCK_TX).
To adapt the Baud Rate Generator to the clock master frequency supplied by the user, a 12 -bit Prescaler must be programmed by loading the Configuration Registers SCI_CR2 (PRESC_H bit 11:8 of the 12 bit prescaler) and SCI_CR3 (PRESC_L bit 7:0 of the 12 bit prescaler). The prescaler allows the programming of all standard Baud Rates by using the most common clock master sources.
The Prescaler value can be obtained by the following formula:

$$
P R E S C=\operatorname{round}\left(\frac{C K M}{16 \times B A U D}\right)
$$

Where CKM is the clock master frequency (expressed in Hz ) and BAUD is the desired Baud Rate (expressed in bit/second). The obtained value is rounded to the nearest integer value. This rounding can cause an error in the obtained Baud Rate. This error must be lower than 3\%. To verify that the PRESC value satisfies this constrain, the obtained Baud Rate must be computed by inverting the previous formula:

$$
\overline{B A U D}=\frac{C K M}{16 \times P R E S C}
$$

then the following relation can be used to verify that the difference with the desired Baud Rate is lower than 3\%:

$$
\frac{|B A U D-\overline{B A U D}|}{B A U D}<0.03
$$

Table 13.1 shows the recommended Prescaler values for common clock master frequencies. To get more precision in Baud Rate, standard quartz frequencies for serial communication can be used. The corresponding Prescaler values for these frequencies are showed in the Table 13.2.

Table 13.1 Recommended Prescaler values for common frequencies (Baud/MHz)

|  | $\mathbf{1}$ | $\mathbf{4}$ | $\mathbf{5}$ | $\mathbf{8}$ | $\mathbf{1 0}$ | $\mathbf{1 2}$ | $\mathbf{1 6}$ | $\mathbf{2 0}$ | $\mathbf{2 4}$ |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{1 2 0 0}$ | 52 | 208 | 260 | 417 | 521 | 625 | 833 | 1042 | 1250 |
| $\mathbf{2 4 0 0}$ | 26 | 104 | 130 | 208 | 260 | 313 | 417 | 521 | 625 |
| $\mathbf{4 8 0 0}$ | 13 | 52 | 65 | 104 | 130 | 156 | 208 | 260 | 313 |
| $\mathbf{9 6 0 0}$ | - | 26 | 33 | 52 | 65 | 78 | 104 | 130 | 156 |
| $\mathbf{1 9 2 0 0}$ | - | 13 | 16 | 26 | 33 | 39 | 52 | 65 | 78 |
| $\mathbf{3 8 4 0 0}$ | - | - | 8 | 13 | 16 | 20 | 26 | 33 | 39 |
| $\mathbf{5 7 6 0 0}$ | - | - | - | - | 11 | 13 | 17 | 22 | 26 |
| $\mathbf{1 1 5 2 0 0}$ | - | - | - | - | - | - | - | 11 | 13 |

Table 13.2 Recommended Prescaler values for serial communication quartz (Baud/MHz)

|  | 1.843 | 2.458 | 3.686 | 4.915 | 6.144 | 7.373 | 9.830 | 11.059 | 12.288 | 14.746 | 19.661 | $\mathbf{2 2 . 1 1 8}$ |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathbf{1 2 0 0}$ | 96 | 128 | 192 | 256 | 320 | 384 | 512 | 576 | 640 | 768 | 1024 | 1152 |
| $\mathbf{2 4 0 0}$ | 48 | 64 | 96 | 128 | 160 | 192 | 256 | 288 | 320 | 384 | 512 | 576 |
| 4800 | 24 | 32 | 48 | 64 | 80 | 96 | 128 | 144 | 160 | 192 | 256 | 288 |
| 9600 | 12 | 16 | 24 | 32 | 40 | 48 | 64 | 72 | 80 | 96 | 128 | 144 |
| $\mathbf{1 9 2 0 0}$ | 6 | 8 | 12 | 16 | 20 | 24 | 32 | 36 | 40 | 48 | 64 | 72 |
| $\mathbf{3 8 4 0 0}$ | 3 | 4 | 6 | 8 | 10 | 12 | 16 | 18 | 20 | 384 | 32 | 36 |
| 57600 | 2 | - | 4 | - | - | 8 | - | 12 | 13 | 16 | 21 | 24 |
| $\mathbf{1 1 5 2 0 0}$ | 1 | - | 2 | - | - | 4 | - | 6 | - | 8 | - | 12 |

### 13.4 SCI Register Description

The following registers are related to the use of the SCI peripheral.

### 13.4.1 SCI Configuration Registers.

SCI Control Register 1 (SCI_CR1)
Configuration Register 22 (016h) Read/Write
Reset Value: 00000000 (00h)


| RXFINT | OVRINT | BRKINT | TXEMINT TXENINT | PAR/T8 | FRM |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7: RXFINT SCDR_RX buffer full interrupt mask 0 : interrupt disabled
1: interrupt enabled

Bit 6: OVRINT Overrun interrupt mask
0 : interrupt disabled
1: interrupt enabled
Bit 5: BRKINT Break interrupt mask
0 : interrupt disabled
1: interrupt enabled
Bit 4: TXEMINT SCDR_TX buffer empty interrupt
0 : interrupt disabled
1: interrupt enabled

Bit 3: TXENINT TX end interrupt mask
0 : interrupt disabled
1: interrupt enabled

Bit 2: PAR/T8 Parity type selection or TX 9th bit 0 : parity odd if enabled, else TX 9th bit=0 1: parity even if enabled, else TX 9th bit=1

Bit 1-0: FRM Frame type selection
00: 8 bit, no parity, 1 stop bit
01: 8 bit, no parity, 2 stop bit
10: 8 bit, parity, 1 stop bit
11: 9 bit, no parity, 1 stop bit

Note: the SCl interrupts are not enabled unless the bit 3 (MSKSCI) of the Configuration Register 0 (INT_MASK) is enabled (set to 1).

SCI Control Register 2 (SCI_CR2)
Configuration Register 23 (017h) Read/Write
Reset Value: 00000000 (00h)

| 4 | 2 |  | 0 |
| :---: | :---: | :---: | :---: |
| PRESC_H | - | RXSTRT | TXSTRT |

Bit 7-4: PRESC_H Baud Rate prescaler (bit 11:8)
These bits are the higher part of the prescaler (see SCI_CR3 Configuration Register) which determinates the baud rate of the communication, according to Table 13.1 and Table 13.2, as explained in Paragraph 13.3.

Bit 3-2: not used

Bit 1: RXSTRT Reception enable
0: RX disabled
1: RX enabled

Bit 0: TXSTRT Transmission enable
0: TX disabled
1: TX enabled

## SCI Control Register 3 (SCI_CR3)

Configuration Register 43 (02Bh) Read/Write Reset Value: 00000000 (00h)

7
0

PRESC_L

Bit 7-0: PRESC_L Baud Rate prescaler (bit 7:0) These bits are the lower part of the prescaler (see SCI_CR2 Configuration Register) which determinates the baud rate of the communication, according to Table 13.1 and Table 13.2, as explained in Paragraph 13.3.

### 13.4.2 SCI Input Registers.

SCI RX data Input Register (SCI_IN)
Input Register 36 (024h) Read only
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| $R X 7$ | $R X 6$ | $R X 5$ | $R X 4$ | $R X 3$ | $R X 2$ | $R X 1$ | $R X 0$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-0: RX7-0 RX Data
In this register the last received serial data can be read.

## SCI Status Register (SCI_STATUS)

Input Register 37 (025h) Read only Reset Value: 00000011 (03h)

7
0

| NSERR | FRERR | RXF | OVERR | - | R8 | TXEM | TXEND |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7: NSERR Noise error
0: noise error not occurred
1: noise error occurred
Bit 6: FRERR Frame error
0 : frame error not occurred
1: frame error occurred

Bit 5: RXF RX data register full
0 : RX data register already read
1: $R X$ data register full but not read yet

Bit 4: OVERR Overrun error
0 : overrun error not occurred
1: overrun error occurred
Bit 3: not used
Bit 2: R8 Received 9th bit
0: RX 9th bit=0
1: RX 9th bit=1

Bit 1: TXEM TX data register empty
0 : TX data register full
1: TX data register empty
Bit 0: TXEND TX end flag
0 : data transferred to the shift register
1: data transmission completed

### 13.4.3 SCI Output Register.

SCI TX data Output Register (SCI_OUT)
Input Register 23 (017h) Write only
Reset Value: 00000000 (00h)


| TX7 | TX6 | TX5 | TX4 | TX3 | TX2 | TX1 | TX0 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

## Bit 7-0: TX7-0 TX Data

In this register the serial data to be transmitted can be written.

## $14 \mathrm{I}^{2} \mathrm{C}$ BUS INTERFACE $\left(\mathbf{I}^{2} \mathrm{C}\right)$

### 14.1 Introduction

The $I^{2} C$ Bus Interface serves as an interface between the microcontroller and the serial $I^{2} \mathrm{C}$ bus, providing both multimaster and slave functions and controls all $\mathrm{I}^{2} \mathrm{C}$ bus-specific sequencing, protocol, arbitration and timing. The $I^{2}$ Bus Interface supports fast $\mathrm{I}^{2} \mathrm{C}$ mode ( 400 kHz ).

### 14.2 Main Features

- Parallel-bus $/{ }^{2} \mathrm{C}$ protocol converter
- Multi-master capability
- 7-bit/10-bit Addressing
- Transmitter/Receiver flag
- End-of-byte transmission flag
- Transfer problem detection
$1^{2} \mathrm{C}$ Master Features:
- Clock generation
- $\mathrm{I}^{2} \mathrm{C}$ bus busy flag
- Arbitration Lost Flag
- End of byte transmission flag
- Transmitter/Receiver Flag
- Start bit detection flag
- Start and Stop generation
$I^{2} \mathrm{C}$ Slave Features:
- Stop bit detection
- $I^{2} \mathrm{C}$ bus busy flag
- Detection of misplaced start or stop condition
- Programmable $\mathrm{I}^{2} \mathrm{C}$ Address detection
- Transfer problem detection
- End-of-byte transmission flag
- Transmitter/Receiver flag

Figure 14.1 $\quad I^{2} C$ BUS Protocol

### 14.3 General Description

In addition to receiving and transmitting data, this interface converts it from serial to parallel format and vice versa, using either an interrupt or polled handshake. The interrupts are enabled or disabled via software. The interface is connected to the $I^{2} \mathrm{C}$ bus by a data pin (SDA) and by a clock pin (SCL). The interface can be connected both with a standard $I^{2} \mathrm{C}$ bus and a Fast $I^{2} \mathrm{C}$ bus. This selection is made via software.

### 14.3.1 Mode Selection.

The interface can operate in the following four modes:

- Slave transmitter/receiver
- Master transmitter/receiver

By default, it operates in slave mode.
The interface automatically switches from slave to master after it generates a START condition and from master to slave in case of arbitration loss or a STOP generation, providing Multi-Master capability.

### 14.3.2 Communication Flow.

In Master mode, Communication Flow initiates data transfer and generates the clock signal. A serial data transfer always begins with a start condition and ends with a stop condition. Both start and stop conditions are generated in master mode by software.
In Slave mode the interface is capable of recognizing its own address (7 or 10-bit) and the General Call address. The General Call address detection may be enabled or disabled by software.
Data and addresses are transferred as 8 -bit bytes, (MSB first). The first byte(s) follow the start condition is the address (one in 7-bit mode, two in 10 -bit mode), which is always transmitted in Master mode.A 9th clock pulse follows the 8 clock cycles of a byte transfer, during which the receiver must send an acknowledge bit to the transmitter. Refer to Figure 14.1.


Acknowledge may be enabled and disabled via software.
The $I^{2} \mathrm{C}$ interface address and/or general call address can be selected via software.
The speed of the $I^{2} \mathrm{C}$ interface may be selected between Standard $(0-100 \mathrm{KHz})$ and Fast $\mathrm{I}^{2} \mathrm{C}$ (100400 KHz ).

### 14.3.3 SDA/SCL Line Control.

Transmitter mode: the interface holds the clock line low before transmission, in order to wait for the microcontroller to write the byte in the Data Register.
Receiver mode: the interface holds the clock line low after reception to wait for the microcontroller to read the byte in the Data Register.
SCL frequency is controlled by a programmable clock divider which depends on the $\mathrm{I}^{2} \mathrm{C}$ bus mode.

When the $I^{2} \mathrm{C}$ cell is enabled, the SDA and SCL pins must be configured as floating open-drain I/O. The value of the external pull-up resistance used depends on the application.

### 14.4 Functional Description

By default the $I^{2} \mathrm{C}$ interface operates in Slave mode ( $\mathrm{M} / \mathrm{SL}$ bit is cleared) except when it initiates a transmit or receive sequence.
First, the interface frequency must be configured using the related bits of the Configuration Registers.

### 14.4.1 Slave Mode.

As soon as a start condition is detected, the address is received from the SDA line and sent to the shift register; then it is compared with the address of the interface or the General Call address (if selected by software).

Figure $14.2 I^{2} \mathrm{C}$ Interface Block Diagram


Note: In 10-bit addressing mode, the comparison includes the header sequence (11110xx0) and the two most significant bits of the address.
Header matched (10-bit mode only): the interface generates an acknowledgement pulse if the ACK bit is set.
Address not matched: the interface ignores it and waits for another Start condition.

Address matched: the interface generates in sequence:

- Acknowledge pulse if the ACK bit is set.
- EVF and ADSL bits are set with an interrupt if the ITE bit is set.
Afterwards, the interface waits for the I2C_SR1 register to be read, holding the SCL line low (see Figure 14.3 Transfer sequencing EV1).
Next, in 7-bit mode read the I2C IN register to determine from the least significant bit (Data Direction Bit) if the slave must enter Receiver or Transmitter mode.
In 10-bit mode, after receiving the address sequence the slave is always in receive mode. It will enter transmit mode on receiving a repeated Start condition followed by the header sequence with matching address bits and the least significant bit set (11110xx1).


## Slave Receiver

Following reception of the address and after the I2C_SR1 register has been read, the slave receives bytes from the SDA line into the I2C_IN register via the internal shift register. After each byte, the interface generates the following in sequence:

- Acknowledge pulse if the ACK bit is set
- EVF and BTF bits are set with an interrupt if the ITE bit is set.
Afterwards, the interface waits for the I2C SR1 register to be read followed by a read of the I2C_IN register, holding the SCL line low (see Figure 14.3 Transfer sequencing EV2).


## Slave Transmitter

Following the address reception and after the I2C_SR1 register has been read, the slave sends bytes from the I2C_OUT register to the SDA line via the internal shift register.
The slave waits for a read of the I2C_SR1 register followed by a write in the I2C_OUT register, holding the SCL line low (see Figure 14.3 Transfer sequencing EV3).
When the acknowledge pulse is received:

- The EVF and BTF bits are set by hardware with an interrupt if the ITE bit is set.


## Closing slave communication

After the last data byte is transferred a Stop Condition is generated by the master. The interface detects this condition and sets:

- EVF and STOPF bits with an interrupt if the ITE bit is set.
Afterwards, the interface waits for a read of the I2C_SR2 register (see Figure 14.3 Transfer sequencing EV4).


## Error Cases

- BERR: Detection of a Stop or a Start condition during a byte transfer. In this case, the EVF and the BERR bits are set with an interrupt if the ITE bit is set.
If it is a Stop then the interface discards the data, released the lines and waits for another Start condition.
If it is a Start then the interface discards the data and waits for the next slave address on the bus.
- AF: Detection of a non-acknowledge bit. In this case, the EVF and AF bits are set with an interrupt if the ITE bit is set.
Note: In both cases, the SCL line is not held low; however, SDA line can remain low due to possible «0» bits transmitted last. At this point, both lines must be released by software.


## How to release the SDA / SCL lines

Set and subsequently clear the STOP bit while BTF is set. The SDA/SCL lines are released after the current byte is transferred.

### 14.4.2 Master Mode.

To switch from default Slave mode to Master mode a Start condition generation is needed.

## Start condition

Setting the START bit while the BUSY bit is cleared causes the interface to switch to Master mode (M/SL bit set) and generates a Start condition.
Once the Start condition is sent:

- The EVF and SB bits are set by hardware with an interrupt if the ITE bit is set.
Afterwards, the master waits for a read of the I2C_SR1 register followed by a write in the I2C_OUT register with the Slave address, holding the SCL line low (see Figure 14.3 Transfer sequencing EV5).


## Slave address transmission

At this point, the slave address is sent to the SDA line via the internal shift register.
In 7-bit addressing mode, one address byte is sent.
In 10-bit addressing mode, sending the first byte including the header sequence causes the following event:

- The EVF bit is set by hardware with interrupt generation if the ITE bit is set.
Afterwards, the master waits for a read of the I2C_SR1 register followed by a write in the I2C_OUT register, holding the SCL line low (see Figưre 14.3 Transfer sequencing EV9).
The second address byte is sent by the interface.
After completion of this transfer (and acknowledge from the slave if the ACK bit is set):
- The EVF bit is set by hardware with interrupt generation if the ITE bit is set.
Afterwards, the master waits for a read of the I2C_SR1 register followed by a write in the I2C_CR register (for example set PE bit), holding the SCL line low (see Figure 14.3 Transfer sequencing EV6).
Next, the master must enter Receiver or Transmitter mode.

Note: In 10-bit addressing mode, in order to switch the master to Receiver mode, software must generate a repeated Start condition and resend the header sequence with the least significant bit set (11110xx1).

## Master Receiver

Following the address transmission and after I2C_SR1 and I2C_CR registers have been accessed, the master receives bytes from the SDA line into the I2C_IN register via the internal shift register. After each byte the interface generates in sequence:

- Acknowledge pulse if the ACK bit is set
- EVFand BTF bits are set by hardware with an interrupt if the ITE bit is set.
Afterwards, the interface waits for a read of the I2C_SR1 register followed by a read of the I2C_IN register, holding the SCL line low (see Figure 14.3 Transfer sequencing EV7).

In order to close the communication: before reading the last byte from the I2C_IN register, set the STOP bit to generate the Stop condition. The interface automatically goes back to slave mode (M/SL bit cleared).
Note: In order to generate the non-acknowledge pulse after the last data byte received, the ACK bit must be cleared just before reading the second last data byte.

## Master Transmitter

Following the address transmission and after the I2C_SR1 register has been read, the master sends bytes from the I2C_OUT register to the SDA line via the internal shift register.
The master waits for a read of the I2C_SR1 register followed by a write in the I2C_OUT register, holding the SCL line low (see Figure 14.3 Transfer sequencing EV8).

When the acknowledge bit is received, the interface sets:

- EVF and BTF bits with an interrupt if the ITE bit is set.
In order to close the communication: after writing the last byte to the I2C_OUT register, set the STOP bit to generate the Stop condition. The interface automatically returns to slave mode (M/ SL bit cleared).


## Error Cases

- BERR: Detection of a Stop or a Start condition during a byte transfer. In this case, the EVF and BERR bits are set by hardware with an interrupt if ITE is set.
- AF: Detection of a non-acknowledge bit. In this case, the EVF and AF bits are set by hardware with an interrupt if the ITE bit is set. To resume, set the START or STOP bit.
- ARLO: Detection of an arbitration lost condition. In this case the ARLO bit is set by hardware (with an interrupt if the ITE bit is set and the interface automatically goes back to slave mode (the M/SL bit is cleared).
Note: In all these cases, the SCL line is not held low; however, the SDA line can remain low due to possible «0» bits transmitted last. Both lines must be released via software.

Figure 14.3 Tranfer Sequencing
7-bit Slave receiver:

| S | Address | A |  | Data1 | A |  | Data2 | A |  | DataN | A |  | P |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  |  |  | EV1 |  |  | EV2 |  |  | EV2 |  |  | EV2 |  | EV4 |

## 7-bit Slave transmitter:



## 7-bit Master receiver:

| S | Address | A |  | Data1 | A |  | Data2 | A |  | DataN | NA |  |  | P |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| EV5 |  |  | EV6 |  |  | EV7 |  |  | EV7 |  |  | EV7 |  |  |

7-bit Master transmitter:


10-bit Slave receiver:


10-bit Master transmitter:


Legend:
S=Start, $\mathrm{P}=$ Stop, $\mathrm{A}=$ Acknowledge, NA=Non-acknowledge
$E V x=E v e n t$ (with interrupt if ITE=1)
EV1: $E V F=1, A D S L=1$, cleared by reading I2C_SR1 register.
EV2: $E V F=1, B T F=1$, cleared by reading I2C_SR1 register followed by reading I2C_IN register.
EV3: $\mathrm{EVF}=1, \mathrm{BTF}=1$, cleared by reading I2C_SR1 register followed by writing I2C_OUT register. $E V 3-1$ : $E V F=1, A F=1, B T F=1, S C L=0$; $A F$ is cleared by reading I2C_SR2. BTF is cleared by releasing the lines (STOP=1,STOP=0) or by readyng I2C_SR1 and writing I2C_OUT register (I2C_OUT=FFh).Note: If lines are released by $S T O P=1, S T O P=0$, the subsequent $E V 4$ is not seer EV4: EVF=1, STOPF=1, cleared by reading I2C_SR2 register.
EV5: $E V F=1$, SB=1, cleared by reading I2C_SR1 register followed by writing I2C_OUT register.
EV6: EVF=1, cleared by reading I2C_SR1 register followed by writing I2C_CR (for example PE= EV7: $E V F=1, B T F=1$, cleared by reading I2C_SR1 register followed by reading I2C_IIN register.
EV8: $E V F=1, B T F=1$, cleared by reading I2C_SR1 register followed by writing I2C_OUT register. EV9: $\mathrm{EVF}=1, \mathrm{ADD} 10=1$, cleared by reading I2C_SR1 register followed by writing I2C_OUT regist

Figure 14.4 Event Flags and Interrupt Generation


| Interrupt Event | Event Flag | Enable Control Bit | Exit from Wait | Exit from Halt |
| :---: | :---: | :---: | :---: | :---: |
| 10-bit Address Sent Event (Master Mode) | ADD10 | ITE | Yes | No |
| End of Byte Transfer Event | BTF |  | Yes | No |
| Address Matched Event (Slave Mode) | ADSEL |  | Yes | No |
| Start Bit Generation Event (Master Mode) | SB |  | Yes | No |
| Acknowledge Failure Event | AF |  | Yes | No |
| Stop Detection Event (Slave Mode) | STOPF |  | Yes | No |
| Arbitration Lost Event (Multimaster configuration) | ARLO |  | Yes | No |
| Bus Error Event | BERR |  | Yes | No |

Note: The $P^{2} C$ interrupt events are connected to the same interrupt vector. They generate an interrupt if the corresponding Enable Control Bit (ITE) is set and the Interrupt Mask bit (MSKI2C) in the INT_MASK Configuration Register is unmasked (set to 1, see Interrupts Chapter).

### 14.5 Register Description

In the following sections describe the registers used by the $\mathrm{I}^{2} \mathrm{C}$ Interface are described.

### 14.5.1 $\mathbf{I}^{2} \mathrm{C}$ Interface Configuration Registers.

## $\mathbf{I}^{2} \mathrm{C}$ Control Register (I2C_CR)

Configuration Register 16 (010h) Read/Write
Reset Value: 00000000 (00h)


| - | - | PE | ENGC | START | ACK | STOP | ITE |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-6: Not Used. They must be held to 0 .

Bit 5: PE Peripheral Enable.
This bit is set and cleared by software
0 : peripheral disabled
1: peripheral enabled

## Notes:

- When PE=0, all the bits of the I2C_CR register and the SR register except the Stop bit are reset. All outputs are released while $\mathrm{PE}=0$
- When $\mathrm{PE}=1$, the corresponding $\mathrm{I} / \mathrm{O}$ pins are selected by hardware as alternate functions.
- To enable the $I^{2} \mathrm{C}$ interface, write the I2C_CR register TWICE with $\mathrm{PE}=1$ as the first write only activates the interface (only PE is set).


## Bit 4: ENGC Enable General Call

This bit is set and cleared by software. It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : General Call disabled
1: General Call enabled
Note: The 00h General Call address is acknowledged (01h ignored).

Bit 3: START Generation of a Start Condition
This bit is set and cleared by software. It is also cleared by hardware when the interface is disabled $(\mathrm{PE}=0)$ or when the Start condition is sent (with interrupt generation if ITE=1).

- In Master Mode

0 : No Start generation
1: Repeated Start generation

- In Slave Mode

0 : No Start generation
1: Start generation when the bus is free
Bit 2: ACK Acknowledge enable
This bit is set and cleared by software. It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : No acknowledge returned
1: Acknowledge returned after an address byte or a data byte is received

Bit 1: STOP Reset signal mode
This bit is set and cleared by software. It is also cleared by hardware in master mode. Note: This bit is not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).

- In Master Mode

0 : No Stop generation
1: Stop generation after the current byte transfer or after the current Start condition is sent. The STOP bit is cleared by hardware when the Stop condition is sent.

- In Slave Mode

0 : No Start generation
1: Release the SCL and SDA lines after the current byte transfer ( $\mathrm{BTF}=1$ ). In this mode the STOP bit has to be cleared by software.

## Bit 0: ITE Interrupt Enable

0 : Interrupt disabled
1: Interrupt enabled

## I ${ }^{2}$ C Clock Control Register (I2C_CCR)

Configuration Register 17 (011h) Read/Write
Reset Value: 00000000 (00h)
7
0

| FM/SM | CC6 | CC5 | CC4 | CC3 | CC2 | CC1 | CC0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7: FM/SM Fast/Standard $\mathrm{I}^{2} \mathrm{C}$ Mode.
This bit is set and cleared by software. It is not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).

1: Standard $\mathrm{I}^{2} \mathrm{C}$ Mode (recommended up to 100 kHz )
0 : Fast I ${ }^{2} \mathrm{C}$ Mode (recommended up to 400 kHz)

Bit 6-0: CC6-CCO 7-bit clock divider
These bits select the speed of the bus ( $\mathrm{F}_{\mathrm{SCL}}$ ) depending on the $I^{2} \mathrm{C}$ mode. They are not cleared when the interface is disabled ( $\mathrm{PE}=0$ ). The speed can be computed as follows:

- Standard mode (FM/SM=1): $\mathrm{F}_{\text {SCL }}<=100 \mathrm{kHz}$

$$
\mathrm{F}_{\mathrm{SCL}}=\mathrm{f}_{\mathrm{CPU}} /(3 \mathrm{x}[\mathrm{CC} 6 . . \mathrm{CC} 0]+11)
$$

- Fast mode (FM/SM=0): $\mathrm{F}_{\mathrm{SCL}}>100 \mathrm{kHz}$

$$
\mathrm{F}_{\mathrm{SCL}}=\mathrm{f}_{\mathrm{CPU}} /(2 x[\mathrm{CC} 6 . . \mathrm{CC} 0]+9)
$$

Warning: For safety reason, CC6-CCO bits must be configured with a value >=3 for the Standard mode and $>=2$ for the Fast mode.

## $1^{2} \mathrm{C}$ Own Address Register 1 (I2C_OAR1)

Configuration Register 18 (012h) Read/Write
Reset Value: 00000000 (00h)

$$
7
$$

0

| ADD7 | ADD6 | ADD5 | ADD4 | ADD3 | ADD2 | ADD1 | ADD0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

## 7-bit Addressing Mode

bit 7-1: ADD7-ADD1 Interface address.
These bits define the $I^{2} \mathrm{C}$ bus address of the interface. They are not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).

Bit 0: ADDO Address direction bit.
This bit is "don't care", the interface acknowledges either 0 or 1 . It is not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).
Note: Address 01 h is always ignored.

## 10-bit Addressing Mode

bit 7-0: ADD7-ADDO Interface address.
These are the least significant bits of the $1^{2} \mathrm{C}$ bus address of the interface. They are not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).
$1^{2} \mathrm{C}$ Own Address Register 2 (I2C_OAR2)
Configuration Register 19 (013h) Read/Write
Reset Value: 00000000 (00h)
$\begin{array}{llll}7 & 2 & 0\end{array}$

| - | - | - | - | - | ADD9 | ADD8 | - |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-3: Not Used
bit 7-1: ADD8-ADD8 Interface address.
These are the most significant bits of th $\mathrm{I}^{2} \mathrm{C}$ bus address of the interface (10-bit mode only). They are not cleared when the interface is disabled ( $\mathrm{PE}=0$ ).

Bit 0: Reserved

### 14.5.2 $I^{2} \mathrm{C}$ Interface Input Registers.

$\mathrm{I}^{2} \mathrm{C}$ Data Input Register (I2C_IN) Input Register 6 (06h) Read only
Reset Value: 00000000 (00h)
7
0

| I2CDI7 | I2CDI6 | I2CDI5 | I2CDI4 | I2CDI3 | I2CDI2 | I2CDI1 | I2CDI0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

bit 7-0: I2CDI7-I2CDI0 Received data.
These bits contain the byte to be received from the bus in Receiver mode: the first data byte is received automatically in the I2C_IN register using the least significant bit of the address.
Then, the next data bytes are received one by one after reading the I2C_IN register.
$1^{2} \mathrm{C}$ Status Register 1 (I2C_SR1)
Input Register 7 (07h) Read only
Reset Value: 00000000 (00h)
7
0

| EVF | ADD10 | TRA | BUSY | BTF | ADSL | M/SL | SB |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

## Bit 7: EVF Event Flag

This bit is set by hardware as soon as an event occurs. It is cleared by software reading I2C_SR2 register in case of error event or as described in Figure 14.3. It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : No event
1: One of the following events has occurred:

- BTF=1 (Byte received or transmitted)
- ADSL=1 (Address matched in Slave mode while ACK=1)
- $\mathrm{SB}=1$ (Start condition generated in Master mode)
- AF=1 (No acknowledge received after byte transmission)
- STOPF=1 (Stop condition detected in Slave mode)
- ARLO=1 (Arbitration lost in Master mode)
- BERR=1 (Bus error, misplaced Start or Stop condition detected)
- Address byte successfully transmitted in Master mode.

Bit 6: ADD10 10 bit addressing in Master Mode
This bit is set by hardware when the master has sent the first byte in 10-bit address mode. It is cleared by software reading I2C_SR2 register followed by a write in the I2C_OUT register of the second address byte. It is also cleared by hardware when the peripheral is disabled ( $\mathrm{PE}=0$ ).

## 0: No ADD10 event occurred

1: The Master has sent the first address byte

## Bit 5: TRA Transmitter/Receiver

When BTF is set, TRA $=1$ if a data byte has been transmitted. It is cleared automatically when BTF is cleared. It is also cleared by hardware after detection of Stop condition (STOPF=1), loss of bus arbitration (ARLO=1) or when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : Data byte received (if BTF=1)
1: Data byte transmitted
Bit 4: BUSY Bus busy
This bit is set by hardware on detection of a Start condition and cleared by hardware on detection of a Stop condition. It indicates a communication in progress on the bus. This information is still updated when the interface is disabled ( $\mathrm{PE}=0$ ).

0 : No communication on the bus
1: Communication ongoing on the bus

Bit 3: BTF Byte transfer finished
This bit is set by hardware as soon as a byte is correctly received or transmitted with interrupt generation if ITE $=1$. It is cleared by software reading I2C_SR1 register followed by a read of I2C_IN or write of I2C_OUT registers. It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).

- Following a byte transmission, this bit is set after reception of the acknowledge clock pulse. In case an address byte is sent, this bit is set only after the EV6 event (see Figure 14.3). BTF is cleared by reading I2C_SR1 register followed by writing the next byte in I2C_OUT register.
- Following a byte reception, this bit is set after transmission of the acknowledge clock pulse if $A C K=1$. BTF is cleared by reading I2C_SR1 register followed by reading the byte from I2C_IN register.
The SCL line is held low while BTF=1.
0 : Byte transfer not done
1: Byte transfer succeeded

Bit 2: ADSL Address matched (Slave Mode)
This bit is set by hardware as soon as the slave address received matched with the OAR register content or a general call is recognized. An interrupt is generated if ITE=1. It is cleared by software reading I2C_SR1 register or by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
The SCL line is held low while ADSL=1.
0: Address mismatched or not received
1: Received address matched

Bit 1: M/SL Master/Slave
This bit is set by hardware as soon as the interface is in Master mode (writing START=1). It is cleared by hardware after detecting a Stop condition on the bus or a loss of arbitration (ARLO=1). It is also cleared when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : Slave mode
1: Master mode

## Bit 0: SB Start bit (Master Mode)

This bit is set by hardware as soon as the Start condition is generated (following a write

START=1). An interrupt is generated if ITE=1. It is cleared by software reading I2C_SR1 register followed by writing the address byte in I2C_OUT register. It is also cleared by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
0 : No Start condition
1: Start condition generated
$1^{2} \mathrm{C}$ Status Register 2 (I2C_SR2)
Input Register 8 (08h) Read only
Reset Value: 00000000 (00h)
7
0

| - | - | - | AF | STOPF | ARLO | BERR | GCAL |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7-5: Reserved.
Bit 4: AF Acknowledge failure.
This bit is set by hardware when an acknowledge is returned. An interrupt is generated if ITE=1. It is cleared by software reading the I2C_SR2 register or by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
The SCL line is not held low while $A F=1$.
0 : No acknowledge failure
1: Acknowledge failure

Bit 3: STOPF Stop detection (Slave mode).
This bit is set by hardware when a Stop condition is detected on the bus after an acknowledge (if ACK=1). An interrupt is generated if ITE=1. It is cleared by software reading I2C_SR2 register or by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
The SCL line is not held low while $S T O P F=1$.
0 : No Stop condition detected
1: Stop condition detected

Bit 2: ARLO Arbitration lost.
This bit is set by hardware when the interface loses the arbitration of the bus to another master. An interrupt is generated if ITE=1. It is cleared by software reading I2C_SR2 register or by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).

After an ARLO event the interface switches back automatically to Slave mode ( $\mathrm{M} / \mathrm{SL}=0$ ). The SCL line is not held low while ARLO=1.
0 : No arbitration lost detected
1: Arbitration lost detected
Bit 1: BERR Bus error.
This bit is set by hardware when the interface detects a misplaced Start or Stop condition. An interrupt is generated if $I T E=1$. It is cleared by software reading I2C_SR2 register or by hardware when the interface is disabled ( $\mathrm{PE}=0$ ).
The SCL line is not held low while BERR=1.
0: No misplaced Start or Stop condition
1: Misplaced Start or Stop condition
Bit 0: GCAL General Call (Slave mode).
This bit is set by hardware when a general call address is detected on the bus while ENGC=1. It is cleared by hardware detecting a Stop condition (STOPF=1) or when the interface is disabled ( $\mathrm{PE}=0$ ).
0: No general call address detected on bus
1: general call address detected on bus

### 14.5.3 $\mathrm{I}^{2} \mathrm{C}$ Interface Output Registers.

## $I^{2} \mathrm{C}$ Data Output Register (I2C_OUT)

Output Register 6 (06h) Read only
Reset Value: 00000000 (00h)
7

|  |  |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| I2CDO7 | I2CDO6 | I2CDO5 | I2CDO4 | I2CDO3 | I2CDO2 | I2CDO1 | I2CDO0 |

bit 7-0: I2CDO7-I2CDOO Data to be transmitted.

These bits contain the byte to be transmitted in the bus in Transmitter mode: Byte transmission start automatically when the software writes in the I2C_OUT register.

## 15 SERIAL PERIPHERAL INTERFACE (SPI)

### 15.1 Introduction

The Serial Peripheral Interface (SPI) allows fullduplex, synchronous, serial communication with external devices. An SPI system may consist of a master, one or more slaves, or a system, in which devices may be either masters or slaves.
SPI is normally used for communication between the ICU and external peripherals or another ICU.
Refer to the Pin Description section in this datasheet for the device-specific pin-out.

### 15.2 Main Features

■ Full duplex, three-wire synchronous transfers

- Master or slave operation
- Four master mode frequencies
- Maximum slave mode frequency $=$ CKM/4.
- Four programmable master bit rates
- Programmable clock polarity and phase
- End of transfer interrupt flag
- Write collision flag protection
- Master mode fault protection capability.


### 15.3 General description

SPI is connected to external devices through 4 alternate pins:

- MISO: Master In / Slave Out pin
- MOSI: Master Out / Slave In pin
- SCK: Serial Clock pin
- $\overline{\mathrm{SS}}$ : Slave select pin (if not done through software)

A basic example of interconnections between a single master and a single slave is illustrated in Figure 15.1
The MOSI pins are connected together as the MISO pins. In this manner, data is transferred serially between master and slave (most significant bit first).
When the master device transmits data to a slave device via the MOSI pin, the slave device responds by sending data to the master device via the MISO pin. This implies full duplex transmission with both data out and data in synchronized with the same clock signal (which is provided by the master device via the SCK pin).
The transmitted byte is replaced by the byte received and eliminates the need for separate transmit-empty and receiver-full bits. A status flag is used to indicate that the I/O operation is complete.
Four possible data/clock timing relationships may be chosen (see Figure 15.4), but master and slave must be programmed with the same timing mode.

### 15.4 Functional Description

Figure 15.2 shows the serial peripheral interface (SPI) block diagram.
This interface contains 3 dedicated registers:

- A Control Register (SPI_CR)
- A Status Register (SPI_STATUS_CR)
- A Data Register for transmission (SPI_OUT)
- A Data Register for reception (SPI_IN)


### 15.4.1 Master Configuration.

In a master configuration, the serial clock is generated on the SCK pin.

Figure 15.1 SPI Master Slave


Figure 15.2 Serial Peripheral Interface Block Diagram


## Procedure

- Select the SPR0, SPR1 and SPR2 bits to define the serial clock baud rate (see SPI_CR register).
- Select the CPOL and CPHA bits to define one of the four relationships between the data transfer and the serial clock (see Figure 15.4).
- The $\overline{\mathrm{SS}}$ pin must be connected to a high level signal during the complete byte transmit sequence.
- The MSTR and SPE bits must be set (they remain set only if the SS pin is connected to a high level signal).
In this configuration the MOSI pin is a data output and to the MISO pin is a data input.


## Transmit sequence

Transmit sequence begins when a byte is written in the SPI_OUT register.

The data byte is loaded in parallel into the 8 -bit shift register (from the internal bus) during a write cycle and then shifted out serially to the MOSI pin most significant bit first.
When data transfer is complete:

- The SPIF bit is set by hardware
- An interrupt is generated if the SPIE bit is set.

During the last clock cycle the SPIF bit is set, a copy of the data byte received in the shift register is moved to a buffer. When the SPI_IN register is read, the SPI peripheral returns this buffered value. Clearing the SPIF bit is performed by the following software sequence:

1. An access to the SPI_STATUS_CR register while the SPIF bit is set
2. A read to the SPI_IN register.

Note: While the SPIF bit is set, all writes to the SPI_OUT register are inhibited until the SPI_STATUS_CR register is read.

### 15.4.2 Slave Configuration.

In slave configuration, the serial clock is received on the SCK pin from the master device.
The value of the SPR0, SPR1 and SPR2 bits is not used for data transfer.

## Procedure

- For correct data transfer, the slave device must be in the same timing mode as the master device (CPOL and CPHA bits). See Figure 15.4.
- The $\overline{\mathrm{SS}}$ pin must be connected to a low level signal during the complete byte transmit sequence.
- Clear the MSTR bit and set the SPE bit to assign the pins to alternate function.
In this configuration the MOSI pin is a data input and the MISO pin is a data output.


## Transmit Sequence

The data byte is loaded into the 8 -bit shift register (from the internal bus) during a write cycle and then shifted out serially to the MISO pin most significant bit first.
The transmit sequence begins when the slave device receives the clock signal and the most significant bit of the data on its MOSI pin.
When data transfer is complete:

- The SPIF bit is set by hardware
- An interrupt is generated if SPIE bit is set.

During the last clock cycle the SPIF bit is set, a copy of the data byte received in the shift register is moved to a buffer. When the SPI_IN register is read, the SPI peripheral returns the buffer value.
The SPIF bit is cleared by the following software sequence:

1. An access to the SPI_STATUS_CR register while the SPIF bit is set.
2. A read to the SPI_IN register.

Note: While the SPIF bit is set, all writes to the SPI_OUT register are inhibited until the SPI_STATUS_CR register is read.
The SPIF bit can be cleared during a second transmission; however, it must be cleared before the second SPIF bit in order to prevent an overrun condition (see Section 15.4.6).
Depending on the CPHA bit, the $\overline{\mathrm{SS}}$ pin has to be set to write to the SPI_OUT register between each data byte transfer to avoid a write collision (see Section 15.4.4).

### 15.4.3 Data Transfer Format.

During an SPI transfer, data is simultaneously transmitted (shifted out serially) and received
(shifted in serially). The serial clock is used to synchronize data transfer during a sequence of eight clock pulses.
The $\overline{\mathrm{SS}}$ pin allows individual selection of a slave device; the other slave devices that are not selected do not interfere with SPI transfer.

## Clock Phase and Clock Polarity

Four possible timing relationships may be chosen by software, using the CPOL and CPHA bits.
The CPOL (clock polarity) bit controls the steady state value of the clock when data isn't being transferred. This bit affects both master and slave modes.
The combination between the CPOL and CPHA (clock phase) bits select the data capture clock edge.
Figure 15.4, shows an SPI transfer with the four combinations of the CPHA and CPOL bits. The diagram may be interpreted as a master or slave timing diagram where the SCK pin, the MISO pin, the MOSI pin are directly connected between the master and the slave device.
The $\overline{\mathrm{SS}}$ pin is the slave device select input and can be driven by the master device.
The master device applies data to its MOSI pinclock edge before the capture clock edge.

## CPHA bit is set

The second edge on the SCK pin (falling edge if the CPOL bit is reset, rising edge if the CPOL bit is set) is the MSBit capture strobe. Data is latched on the occurrence of the second clock transition.
A write collision should not occur even if the $\overline{\mathrm{SS}}$ pin stays low during a transfer of several bytes (see Figure 15.3).

## CPHA bit is reset

The first edge on the SCK pin (falling edge if CPOL bit is set, rising edge if CPOL bit is reset) is the MSBit capture strobe. Data is latched on the occurrence of the first clock transition.
The $\overline{\mathrm{SS}}$ pin must be toggled high and low between each byte transmitted (see Figure 15.3).
In order to protect the transmission from a write collision a low value on the SS pin of a slave device freezes the data in its SPI_OUT register and does not allow it to be altered. Therefore, the SS pin must be high to write a new data byte in the SPI_OUT without producing a write collision.

### 15.4.4 Write Collision Error.

A write collision occurs when the software tries to write to the SPI_OUT register while a data transfer
is taking place with an external device. When this occurs, the transfer continues uninterrupted; and the software writing will be unsuccessful.
Write collisions can occur both in master and slave mode.

Note: a "read collision" will never occur since the data byte received is placed in a buffer, in which access is always synchronous with the ICU operation.

## In Slave mode

When the CPHA bit is set:
The slave device will receive a clock (SCK) edge prior to the latch of the first data transfer. This first clock edge will freeze the data in the slave device SPI_OUT register and output the MSBit on to the external MISO pin of the slave device.
The $\overline{\mathrm{SS}}$ pin low state enables the slave device, but the output of the MSBit onto the MISO pin does not take place until the first data transfer clock edge occurs.
When the CPHA bit is reset:
Data is latched on the occurrence of the first clock transition. The slave device doesn't have a way of knowing when that transition will occur; therefore, the slave device collision occurs when software attempts to write the SPI_OUT register after its SS pin has been pulled low.
For this reason, the $\overline{S S}$ pin must be high, between each data byte transfer, in order to allow the CPU to write in the SPI_OUT register without generating a write collision.

## In Master mode

Collision in the master device is defined as a write of the SPI_OUT register, while the internal serial clock (SCK) is in the process of transfer.
The $\overline{\mathrm{SS}}$ pin signal must always be high on the master device.

Figure 15.3 CHPA/SS Timing Diagram

## WCOL bit

The WCOL bit in the SPI_STATUS_CR register is set if a write collision occurs.
No SPI interrupt is generated when the WCOL bit is set (the WCOL bit is a status flag only).
The WCOL bit is cleared by a software sequence (see Section 15.5).

### 15.4.5 Master Mode Fault.

Master mode fault occurs when the master device has its $\overline{\text { SS }}$ pin pulled low, then the MODF bit is set. Master mode fault affects the SPI peripheral in the following ways:

- The MODF bit is set and an SPI interrupt is generated if the SPIE bit is set.
- The SPE bit is reset. This blocks all output from the device and disables the SPI peripheral.
- The MSTR bit is reset, forcing the device into slave mode.
Clearing the MODF bit is done through a software sequence:

1. A read or write access to the SPI_STATUS_CR register while the MODF bit is set.
2. A write to the SPI_CR register.

Note: To avoid any multiple slave conflicts in the case of a system comprising several MCUs, the SS pin must be pulled high during the clearing sequence of the MODF bit. The SPE and MSTR bits may be restored to their original state during or after this clearing sequence.
Hardware does not allow the user to set the SPE and MSTR bits, while the MODF bit is set (except in the MODF bit clearing sequence).
In a slave device the MODF bit can't be set, but in a multi master configuration the device can be in slave mode with this MODF bit set.
The MODF bit indicates that there might have been a multi-master conflict for system control and allows a proper exit from system operation to a reset or default system state using an interrupt routine.


Figure 15.4 Data Clock Timing Diagram
CPOL $=1$

Figure 15.5 Clearing the WCOL bit (Write Collision Flag) Software Sequence


Clearing sequence before SPIF = 1 (during a data byte transfer)


Note: Writing in SPI_OUT register instead of reading in SPI_IN do not reset WCOL bit

### 15.4.6 Overrun Condition.

An overrun condition occurs when the master device has sent several data bytes and the slave device hasn't cleared the SPIF bit issued from the previous data byte transmitted.
In this case, the receiver buffer contains the byte sent after the SPIF bit was last cleared. A read to the SPI_IN register returns this byte. All other bytes are lost.
This condition is not detected by the SPI peripheral.

### 15.4.7 Single Master and Multimaster Configurations.

There are two types of SPI systems:

- Single Master System
- Multimaster System


## Single Master System

A typical single master system may be configured, using an ICU as the master and four ICUs as slaves (see Figure 15.6).
The master device selects the individual slave devices by using_four pins of a parallel port to control the four SS pins of the slave devices.

The $\overline{\mathrm{SS}}$ pins are pulled high during reset since the master device ports will be forced to be inputs at that time, thus disabling the slave devices.
Note: In order to prevent a bus conflict on the MISO line the master allows only one active slave device during a transmission.

For more security, the slave device may respond to the master with the data byte received. Then the master will receive the previous byte back from the slave device if all MISO and MOSI pins are connected and the slave has not written its SPI_OUT register.
Other transmission security methods can use ports for handshake lines or data bytes with command fields.

## Multi-master System

A multi-master system may also be configured by the user. Transfer of master control could be implemented using a handshake method through the I/O ports or by an exchange of code messages through the serial peripheral interface system.
The multi-master system is principally handled by the MSTR bit in the SPI_CR register and the MODF bit in the SPI_STATŪS_CR register.

Figure 15.6 Single Master Configuration


### 15.4.8 Interrupts

| Interrupt Event | Event <br> Flag | Enable <br> Control <br> Bit | Exit <br> from <br> Wait | Exit <br> from <br> Halt |
| :--- | :---: | :---: | :---: | :---: |
| SPI End of Transfer Event | SPIF | SPIE | Yes | No |
|  | Master Mode Fault Event |  |  | No |

Note: The SPI interrupt events are connected to the same interrupt vector (see Interrupts chapter). They generate an interrupt if the corresponding Enable Control Bit (SPIE) and the interrupt mask bit (MSKSPI) in the INT_MASK Configuration Register is set.

### 15.5 SPI Register Description

In the following sections describe the registers used by the SPI.

### 15.5.1 SPI Configuration Registers.

## SPI Control Register (SPI_CR)

Configuration Register 20 (014h) Read/Write
Reset Value: 00000000 (00h)

7

| SPIE | SPE | SPR2 | MSTR | CPOL | CPHA | SPR1 | SPR0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

Bit 7: SPIE Serial peripheral interrupt enable.
This bit is set and cleared by software.
0 : Interrupt is inhibited
1: An SPI interrupt is generated whenever SPIF=1 or MODF=1 in SPI_STATUS_CR

Bit 6: SPE Serial peripheral output enable.
This bit is set and cleared by software. It is also cleared by hardware when, in master mode, $\overline{\mathrm{SS}}=0$ (see Section 15.4.5 Master Mode Fault).
0: I/O port connected to pins
1: SPI alternate functions connected to pins
Note: The SPE bit is cleared by reset, so the SPI peripheral is not initially connected to the pins.

Bit 5: SPR2 Divider Enable.
This bit is set and cleared by software and it is cleared by reset. It is used with the SPR[1:0] bits to set the baud rate. Refer to Table 15.1.
0 : Divider by 2 enabled
1: Divider by 2 disabled
Note: This bit has no effect in slave mode.

Bit 4: MSTR Master/Slave mode select.
This bit is set and cleared by software. It is also cleared by hardware when, in master mode, $\overline{\mathrm{SS}}=0$ (see Section 15.4.5 Master Mode Fault).
0 : Slave mode is selected
1: Master mode is selected, the function of the SCK pin changes from an input to an output and the functions of the MISO and MOSI pins are reversed.

Bit 3: CPOL Clock polarity.
This bit is set and cleared by software. This bit determines the steady state of the serial Clock. The CPOL bit affects both the master and slave modes.
0 : The steady state is a low value at the SCK pin.
1: The steady state is a high value at the SCK pin.

Note: SPI must be disabled by resetting the SPE bit if CPOL is changed at the communication byte boundaries.

Bit 2: CPHA Clock phase.
This bit is set and cleared by software.
0 : The first clock transition is the first data capture edge.
1: The second clock transition is the first capture edge.

Bit 1-0: SPR1-SPR0 Serial peripheral rate.
These bits are set and cleared by software. Used with the SPR2 bit, they select one of six baud rates to be used as the serial clock when the device is a master (see Table 15.1). These 2 bits have no effect in slave mode.

Remark: It is recommended to write the SPI_CR register after the SPI_STATUS_CR register when working in master mode, vice versa when working in slave mode.

Table 15.1 Serial Peripheral Baud Rate

| Serial Clock | SPR2 | SPR1 | SPR0 |
| :---: | :---: | :---: | :---: |
| $\mathrm{f}_{\mathrm{CKM}} / 2$ | 1 | 0 | 0 |
| $\mathrm{f}_{\mathrm{CKM}} / 4$ | 0 | 0 | 0 |
| $\mathrm{f}_{\mathrm{CKM}} / 8$ | 0 | 0 | 1 |
| $\mathrm{f}_{\mathrm{CKM}} / 16$ | 1 | 1 | 0 |
| $\mathrm{f}_{\mathrm{CKM}} / 32$ | 0 | 1 | 0 |
| $\mathrm{f}_{\mathrm{CKM}} / 64$ | 0 | 1 | 1 |

## SPI Control-Status Register (SPI_STATUS_CR)

Configuration Register 21 (015h) Read/Write
Reset Value: 00000000 (00h)
7
0

| SPIF | WCOL | OR | MODF | - | SOD | SSM | SSI |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

Bit 7: SPIF Serial Peripheral data transfer flag. (read only)
This bit is set by hardware when a transfer has been completed. An interrupt is generated if SPIE=1 in the SPI_CR register. It is cleared by a software sequence (an access to the SPI_STATUS_CR register followed by a read or write to the SPI_IN/ SPI_OUT registers).
0: Data transfer is in progress or has been approved by a clearing sequence.
1: Data transfer between the device and an external device has been completed.
Note: While the SPIF bit is set, all writes to the SPI_OUT register are inhibited.

Bit 6: WCOL Write Collision status (read only).
This bit is set by hardware when a write to the SPI_OUT register is done during a transmit sequence. It is cleared by a software sequence (see Figure 15.5).
0 : No write collision occurred
1: A write collision has been detected

Bit 5: OR SPI overrun error (read only).
This bit is set by hardware when the byte currently being received in the shift register is ready to be transferred into the SPI IN register while SPIF $=1$ (See Section 15.4.6 Overrun Condition). It is cleared by a software sequence (read of the SPI_STATUS_CR register followed by a read in SPI_IN or write of the SPI_OUT register).
0 : No overrun error.
1: Overrun error detected.

Bit 4: MODF Mode Fault flag (read only).
This bit is set by hardware when the $\overline{\mathrm{SS}}$ pin is pulled low in master mode (see Section 15.4.5 Master Mode Fault). An SPI interrupt can be generated if SPIE=1 in the SPI_CR register. This bit is cleared by a software sequence (An access to the SPI_STATUS_CR register while MODF=1 followed by a write to the SPI_CR register).
0 : No master mode fault detected
1: A fault in master mode has been detected

Bit 3: Not used.

Bit 2: SOD SPI output disable
This bit is set and cleared by software. When set, it disables the alternate function of the SPI output (MOSI in master mode / MISO in slave mode)
0: SPI output not disable
1: SPI output disable.
Bit 1: SSM $\overline{\text { SS }}$ mode selection
This bit is set and cleared by software. When set, it disables the alternate function of the SPI Slave Select pin and use the SSI bit value instead of.
$0: \overline{S S}$ pin used by the SPI.
1: $\overline{S S}$ pin not used (I/O mode), SSI bit value is used.

Bit 0: SSI $\overline{\operatorname{SS}}$ internal mode
This bit is set and cleared by software. It replaces pin SS of the SPI when bit SSM is set to 1 . SSI bit is active low slave select signal when SSM is set to 1 .
0 : Slave selected
1 : Slave not selected.

Remark: It is recommended to write the SPI_CR register after the SPI_STATUS_CR register when working in master mode, vice versa when working in slave mode.

### 15.5.2 SPI Input Register.

## SPI Data Input Register (SPI_IN)

Input Register 5 (05h) Read only
Reset Value: 00000000 (00h)
7
0

| SPIDI7 | SPIDI6 | SPIDI5 | SPIDI4 | SPIDI3 | SPIDI2 | SPIDI1 | SPIDI0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

bit 7-0: SPIDI7-SPIDIO Received data.

The SPI_IN register is used to receive data on the serial bus.
Note: During the last clock cycle the SPIF bit is set, a copy of the data byte received in the shift register is moved to a buffer. When the user reads the serial peripheral data I/O register, the buffer is actually being read.

Warning: A read to the SPI_IN register returns the value located in the buffer and not the contents of the shift register (see Figure 15.2).

### 15.5.3 SPI Output Register.

## SPI Data Output Register (SPI_OUT)

Output Register 5 (05h) Write only
Reset Value: 00000000 (00h)

$$
\begin{array}{ll}
7 & 0
\end{array}
$$

| SPIDO7 | SPIDO6 | SPIDO5 | SPIDO4 | SPIDO3 | SPIDO2 | SPIDO1 | SPIDO0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |

bit 7-0: SPIDO7-SPIDO0 Data to be transmitted.
The SPI_OUT register is used to transmit data on the serial bus. In the master device only a write to this register will initiate transmission/reception of another byte.
Warning: A write to the SPI_OUT register places data directly into the shift register for transmission.

Full Product Information at http://mcu.st.com
Information furnished is believed to be accurate and reliable. However, STMicroelectronics assumes no responsibility for the consequences of use of such information nor for any infringement of patents or other rights of third parties which may result from its use. No license is granted by implication or otherwise under any patent or patent rights of STMicroelectronics. Specification mentioned in this publication are subject to change without notice. This publication supersedes and replaces all information previously supplied. STMicroelectronics products are not authorized for use as critical components in life support devices or systems without express written approval of STMicroelectronics.

The ST logo is a registered trademark of STMicroelectronics
© 2003 STMicroelectronics - Printed in Italy - All Rights Reserved

STMicroelectronics GROUP OF COMPANIES
Australia - Brazil - China - Canada - Finland - France - Germany - Hong Kong - India - Israel - Italy - Japan - Malaysia - Malta - Morocco - Singapore - Spain - Sweden - Switzerland - United Kingdom - U.S.A.
http://www.st.com

